HUGO GERNSBACK, *Editor*

RADIO CRAFT

TELEVISION OVER A LIGHT BEAM

SEE PAGE 22

In this issue—

VFO Transmitter or Exciter

Improving Midget Receivers

Moving-Coil Phono Pickup

JUL 1947

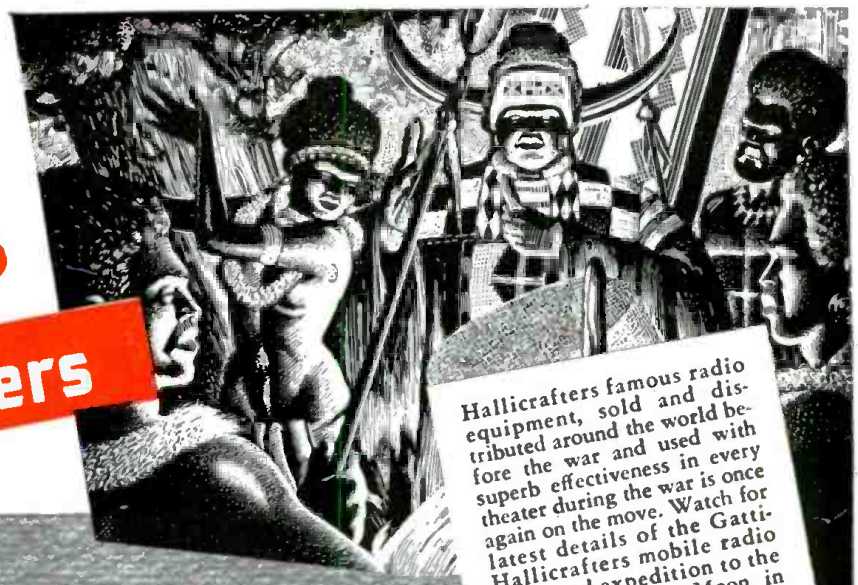
25¢

CANADA 30¢

RADIO-ELECTRONICS IN ALL ITS PHASES

Going places (AGAIN)

hallicrafters



Hallicrafters famous radio equipment, sold and distributed around the world before the war and used with superb effectiveness in every theater during the war is once again on the move. Watch for latest details of the Gatti-Hallicrafters mobile radio equipped expedition to the Mountains of the Moon in deepest Africa—a new and exciting test for the ingenuity of hams and the performance of Hallicrafters equipment.

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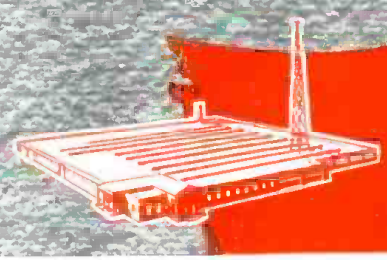

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RADIO SERVICE EDITION

JULY *Prepared by* SYLVANIA ELECTRIC PRODUCTS INC., Emporium, Pa. 1947

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1910 1947

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A dependable instrument of wide utility—sensitivity 1000 ohms per volt.
Ranges: Volts AC, DC, and Output Ranges, 0-10/50/100/500/1000; Ohms full scale, 500,000. Ohms center scale, 7200.

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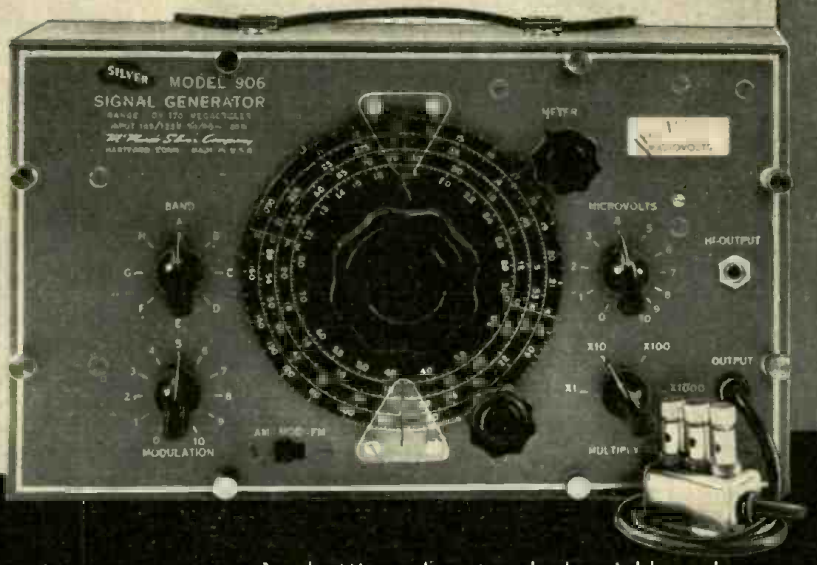

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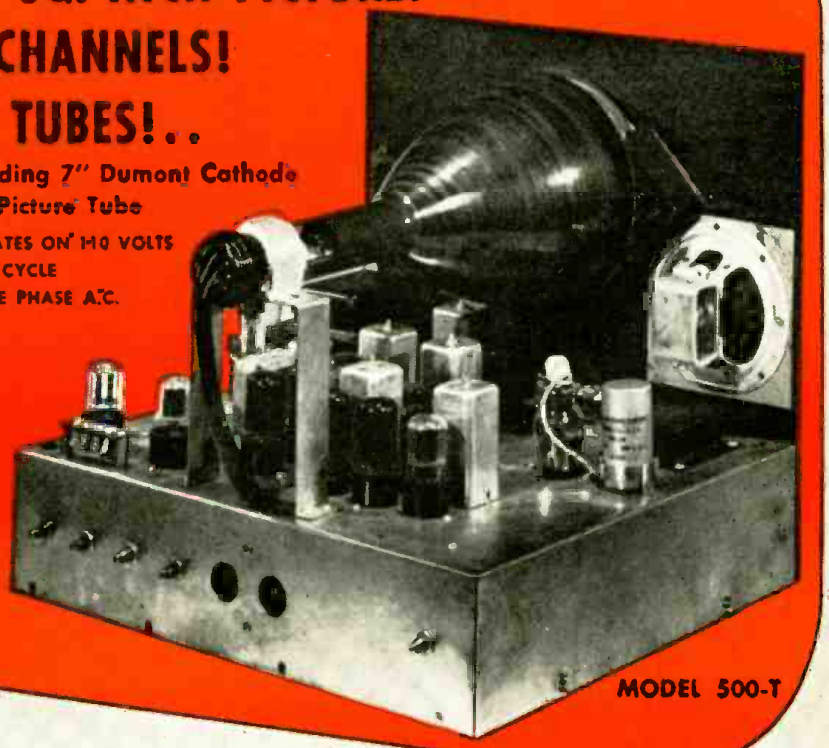

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JULY • 1947

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AND POPULAR ELECTRONICS
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RADIO & TELEVISION

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Ultrasonic circuits

On THE COVER:

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"My CREI training has helped me with radio theory so that I was prepared to undertake problems in design, construction installation and maintenance as they came along."
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—A Chief Engineer—4-22-47

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801 Oscillator and 801 Power Amplifiers, 2-46 Modulators and 1-46 Speech Amplifier 4 Xtal Frequencies and Master Oscillator on selector switch. 10 to 30 watts output. Tone Voice or C.W. Mod. Ideal for 80 meter band. Comes with 3 coils TU 17A 2000-3000 Kc. TU 16 3000-4500 Kc. TU 25 3500-5250 Kc. Black wrinkle case. Includes 2 separate cases to store extra coils. Frequencies chart and tubes included, packed in original cases, less crystals at this low price. Cat. No. MT-100 . . . Shipping weight 125 lbs. **$29.95**

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Mounted in aluminum shield can. 1500 KC, with air trimmer, impedence coupled type. **95c**

30 MC IF Transformer
In square aluminum can, silver slug tuned **29c**

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Thordarson Pri. 110 V 60 cy.—Sec. 6. 3V, 6A, CT. Cat. No. FT12. ..... **$1.49**

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SCR 625 used. **$49.95**
Brand new. 69.00

Ass't mica condensers, Cat. No. C-12—per 100. ..... $ 1.95
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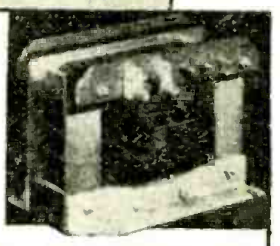
The famous boat anchor, widely used on the 144 MC band. Complete power supply 110V—AC. Less power transformer and tubes. **$14.95**
Shipping weight 100 lbs.

MODULATION TRANSFORMER 1KW

Shipping weight 52 lbs. **$14.95**



RCA modulation transformer is conservatively rated at 550 Watt audio to modulate that new KW rig. Really rugged construction with protective flashover gaps, which are adjustable. Terminals and gaps are mounted on a "Mycalex" terminal board. The laminations that make up this transformer are of high audio quality and are extremely thin, making it impossible for the core to "chatter or talk".
Audio Watts—550 Sec. @1-450 Mils Sec. @2-80 Mils Turns
Ratio—Pri Sec. @1-1:1 Pri Sec. @2-5:1 Pri Sec. @2 Tap-25:1.
Impedance Ratio—Pri @1-1:1 Sec. Pri Sec. @2-25:1 Pri Sec. @2 Tap-625:1.
DC Resistance—Pri 135 ohms Sec @1, 112 ohms; Sec. @2, 99 ohms.
Transformers insulation tested: Pri. 8000V.; Sec. @2-2000V. to the rest of the coils and core. Primary center tapped for Class "B" modulators. Secondary @2 will carry 80 Mils to modulate screens of beam power or screen grid tubes. Primary will match any Class "B" tubes up to 10,000 ohms plate to plate, such as 810's, 75T's, 805's, ZB120's, 203's, HY51Z's, 211's, 813's, 828's, 805's, 2037's.
Size 9½" wide, 7½" deep, 7¼" high. Heavy channel iron mounting brackets. Weight approx. 40 lbs.

BC 654
TRANSMITTER & RECEIVER

Frequencies range 3800-5800 KC.—calibration every 10 KC.—with crystal oscillator checked every 200 KC. Power output 17 watts, voice or CW. Complete with tubes and 200 KC. X-tal.
$14.95

BUTTERFLY Condensers

Oscillator butterfly assembly condenser 76 to 300 megacycles with acorn tube socket. Mounted on condenser. Catalog No. BC 3. **$1.95**

Type B — frequency range 300 to 1000 megacycles to be used with 368 AS doorknob tube. Cat No. BC2. **95c**

Sockets part of assembly

TUBES		TUBES	
813	$ 5.45	IT4	3S4
RK60	1.25	IS5	IR5
VT127	2.95	3Q4	6SL7
VR150	.69	6SN7	6SA7
829	2.45		
872	1.95	**59c each**	
211	1.45		
654	1.50	955	.65
656	.95	9004	.65

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THORDARSON T48003

2H—7H 550 MA swing choke. Size 4½ x 5½ x 5½. Square black crackle case. Cat. No. FC-205. **$5.95**

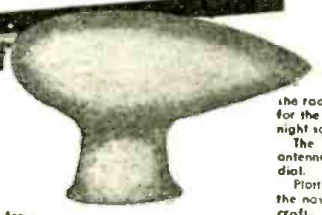
Thordarson 8HY 150M choke, Cat. No. FC201—95c
Thordarson 8HY 175 Mchoke, Cat. No. FC202—$1.49
Thordarson 12HY 25M choke, Cat. No. FC203—39c
Thordarson 8HY 350M choke, Cat. No. FC204—$4.95

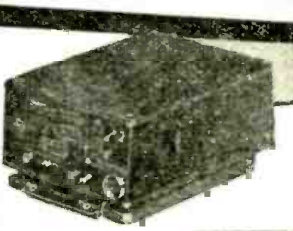
Receiver & Transmitter

SCR522, 100-156 MC. Used, in good condition. Complete with 18 tubes and crystals. **$29.95**

BRAND NEW SCR-269-F AUTOMATIC DIRECTION FINDER RADIO COMPASS

COMPLETE WITH COMPONENT PARTS **$75.00**





The radio compass SCR-269-F was designed to be the primary radio navigation compass for the United States Army and Navy Air Forces. Constant reception is possible day or night so that fixes can always be made to establish the plane's or ship's location.
The azimuth indicator is divided into 360 degrees and is connected to the loop antenna, therefore making it possible to navigate the ship in any direction as preset on the dial.
Plotting fixes is accomplished by selecting two or more stations and plotting these on the navigation map. The point of intersection of these lines, indicates the location of the craft.
This equipment comes complete with 17 tubes superheterodyne receiver which is tunable from 200-1750 KC in three bands. A complete instruction book for operation and maintenance accompanies this equipment.

—for—
- Commercial air lines
- Air charter planes
- Yachts
- Fishing vessels
- Home radios
- Ham operators
- Air freight planes

1 Radio Compass Receiver BC-433-F
1 Radio Control Box BC-434-F
1 Mounting FT-213-A
1 Mounting FT-224-F.
1 Loop LP-21-F (Includes Dehydrator)
1 Cord CD-365-A
1 Indicator I-81-F
1 Relay SW-172
1 Plug PL-112
1 Plug PL-118
1 Plug PL-122
1 Dehydrator Hose, Fitting & Clamps 10 foot lengths
1 Operating & Maintenance Handbook
1 Coupling MC-136
1 Tuning Shaft MC-124 (300")
1 Insulator IN-79
1 Insulator IN-81
1 Shaft Casing and Spline Drive
1 Shafting F/MC-124-(300")
5 Nut F/MC-124
5 Spline F/MC-124
5 Sleeve F/MC-124
1 Transformer C289AS-R16-F

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Hershel RADIO COMPANY
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20% DEPOSIT REQUIRED ON ALL C.O.D. ORDERS

Rectifier Transformer

117V AC input, 62V output at 3.5A. Rectify and use with filter and rheostat to run your 12 and 24 volt equipment $3.95

"Communications"

Has the Electronic Equipment You Need

GN-45 Hand Generators

Plate & Fil Supply Output 500VDC 140 ma and 6V 3A. 8"x8"x6" $4.95

Selections From Our Huge Microwave Stock

10 CM. PLUMBING

Sand Load—Wave guide section. App: 23" high with cooling fins (Dummy Ant.) ..$35.00
Wave Guide to Coax, with flange Gold plated 10 CM 25.00
Standing Wave Detector, rigid coax 58 ohm. 5.00
Coax rotary joint, with mounting plate 8.00
Wave guide (16 ft. lengths) per foot 10 CM 2.00
Heavy Flange for mounting wave guide 2.00

3 CM PLUMBING

Wave guide 90° bend E Plane, 18" long ..$ 4.00
Wave guide, 18" long S curve, cover to choke 3.50
Rotary coupler, wave guide in and out, choke to choke 6.00
Duplexer section using 1B24 ..... 10.00
Wave guide 5 ft. length, per ft. 1.95

1.25 CM. PLUMBING

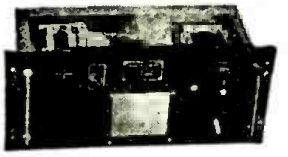
T section choke to cover .....$ 4.50
Flexible section choke to cover 3.00
Mitered elbow and S section cover to choke 3.50

MICROWAVE TEST PARTS

SPECIAL—3 cm. vernier drive dial, and resonant cavity Macguire Wavemeter #1539TFX, 21GA $20.00
4 inch directional coupler, 20Db drop 3 Cm—in and out, coax probe 5.50
8" long directional coupler, CU-90/UP-20Db drop has short right angle (10 cm) 5.50
1-222A Signal Generator—8-15 MC and 150-200MC 5 MC crystal osc. for calibration power input 115 volts, 60 cycles AC 47.50
W.E. Signal Generator 1-138 A (10 cm) indicator is a 0 to 200 microammeter. Value $400. Our price 75.00

SEND FOR OUR NEW MICROWAVE FLYER

TUNING UNITS
Ideal for E.C.O. Rig



From BC191 and 375—contains coils, chokes, dials, condensers; send your frequencies $2.75

TRANSMITTER BC 375

Air Corps version of the famous BC-191. Used; supplied with one tuning unit, meters, coils, condensers, etc. Special $9.95

SO RADAR

10 CM SURFACE SEARCH RADAR SETS SO-9 new with spares—complete installations. SO-13 used, in excellent condition.
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RADAR UNITS

SN Radar, complete, very compact, portable, new $650.00
RC 148 IFF, includes trans. rec. BC1267 and pwr. unit RA105-A. Operation: 117v/60c—157 to 187 MC 47.50
SA RADAR, PPI Unit, from SA-2 Radar RCA 7 in scope 165.00
SQ RADAR Complete Console, GE. B. PPI, or A. Presentation 250.00
SLa ppi console, WE Radar, good cond. 225.00
AN/APG-5 Trans.-rec. 2500MC Range radar uses lighthouse tubes 100.00
SCR 518 RADAR ALTIMETER—20000 ft. range 3" calibrated scale, 515MC, complete installation. New 279.00
RANGE UNIT SCR268—Radar. 115v/60c 15.00

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3131 (1 cm) Magnetron $20.00
3131 Fil. Transformer $6.50
2132 (JAN) 10 cm. 300 Kw peak pulse listed $200. OUR PRICE $25.00
720—BY MAGNETRON W.E. $25.00
ELECTRO - MAGNET for 720-BY....$13.50
QK59, QK60, QK61, QK62. Pkgd. Magnetrons 10 cm Band

$45.00 each

Magnets for Magnetrons $12.00
KLYSTRON oscillator tubes 2K25/723 ab designed for 3 cm operation. New. With complete data. Listed at $38.00, reduced to 7.75
McNally Klystron 707B (10 CM) with cavity 15.00
Tunable Cavity for 707B 3.50
Thermistor Beads (D-170396), for use with UHF and Micro-Wave Equipment (List $7.00). In separate sealed containers .95
Thermistor Buttons D 168391 ..... .95
Magnet for 3 centimeter magnetrons, incorporating a magnetron stabilizer cavity. Cinaudagraph Type UC 210 ..... 22.50
WE. Converter BC 437A 19.00
Parabolic Reflector—Ribbed—For operation in 3000-4000 MC Range. Made by Budd. Has Mounting Neck. New 85.00
Dipole—Can be used with above 5.00

DIRECTION FINDER
Designed for Marine Use

Navy DP12, covers broadcast and marine frequency bands, 100-1500 UC. Complete installation with accessories and spares. Operation: 115V AC or battery pack. Loop can be operated remotely.....$195.00

H. V. TRANSFORMERS

AMERTRAN PLATE TRANSFORMER. 115 v—60-cycle primary. 6200 volt-ct-700 mil secondary. Size 11"x14"x10". **$39.95**

AMERTRAN CHOKE—RMS test 15KV. 1 H. .8 amp. DC, DC resistance 7.5 ohms $8.95

Power and Filament
All Primaries: 117V 60 Cycles Secondaries

#5084—1000V CT and 8800V CT at 250 ma, and 6.3V at 1.5A $6.95
#5114—1500V CT at 200ma and 5V at A 6.95
#5054—980V at 450ma—$4.95. Per pair 9.00
#5109—6180V at 200ma 14.75
#5057—6.3 CT 1A, 5V CT 3A, 5V CT 3A 2.75
#5104—6.3V at 1A, 6.3V at 1A, 6.3V at 1A 2.45
#5126—5V CT 3A, 5V CT 3A, 5V CT at 6A 3.25

CHOKES

8.5hy, 125ma, 1780V test $1.45
Dual: 7hy, 75ma, 11hy 60ma, 1780V test 1.95
Very conservatively rated

Sonar Sound Detection Unit!

Ideal for detecting underwater sounds, such as fish swimming in schools, within a 15 mile area. Using a Rochelle salt crystal, which is about 1000 times more sensitive than quartz, as the active unit the sound is transmitted up a 60 ft. length of cable. It is completely enclosed in a solid rubber sheath whose physical size is: 16½"L. x 2½"dia. This sound detector was originally used in harbor defense. Coupled to an audio amplifier, this can be found to have many valuable applications.

Ask for SD-1 ..... **$6.95**

Similar to above but 3 times longer and with more crystals for greater sensitivity. **Model JR** $12.50

HEADPHONES

8,000 ohms impedance, complete with 6 ft. cord, plug, detachable cushions, leather headband. Brand new.....$1.95
Used, in good condition.... 1.00

RECEIVERS
NEW ARC-5 SUPERHETS....$6.00

Tubes (included) 3-12SK7; 1-12K8; 1-12SR7; 1-12AG. Ranges (Specify freq. desired) 1500-3000 Kc; 3-6 Mc; 6-9.1 Mc. Power: 24-28 VDC for only, new $6.00
Extra — To Go With ARC-5 Dynamotor $1.95
Control Box $1.00
Control drive cables 1.50
Single mounting rack 1.00
Double mounting rack 1.65
Shock mounts ..... .95

All parts interchangeable with SCR-274-N

RCA 12-TUBE SUPERHET

Frequency: 100 to 1500KC. Used but in excellent condition. Power supply 115V AC 24" L x 18" W x 13" H $29.95

That Condenser You Need
Oil Type

1 mf 300 vdc	$ .20
2 mf 300 vdc	.25
4 mf 300 vdc	.30
4 mf 400 vdc	.50
5-5 mf 400 vdc	1.05
1 mf 500 vdc G.E.	.25
2 mf 550 vdc	.25
.25 mf 600 vdc	.20
.85 mf 600 vdc	.25
1 mf 600 vdc	.30
8-8-4 mf 650 vdc	1.45
.1-.1 mf 7000 vdc G.E. Dry	2.00
2 mf 800 vdc	.35
4 mf 600 vdc	.60
10 mf 600 vdc	.85
1 mf 1000 vdc	.75
.1 mf 1000 vdc	.25
2 mf 1000 vdc	.89
4 mf 1000 vdc	1.00
10 mf 1000 vdc	1.40
1 mf 1500 vdc	.95
.4 mf 1500 vdc	.15
2 mf 660 ac/1000	.85
4 mf 1500 vdc	1.20
1 mf 2000 vdc	1.00
1 mf 3000 vdc	4.95
1 mf 7500 vdc	12.50
.25 mf 20,000 vdc	17.50
10-10-10 mf Synchro cap 90v/60c	2.50

SANGAMO CAPACITORS

.03 .006, 10KV $17.50
.0001, 10KV 15.50

We have a wide selection of 120 watt resistors. Send your requirements.

ALL MERCHANDISE GUARANTEED. MAIL ORDERS PROMPTLY FILLED. ALL PRICES F.O.B. NEW YORK CITY. SEND MONEY ORDER OR CHECK. SHIPPING CHARGES SENT C.O.D.

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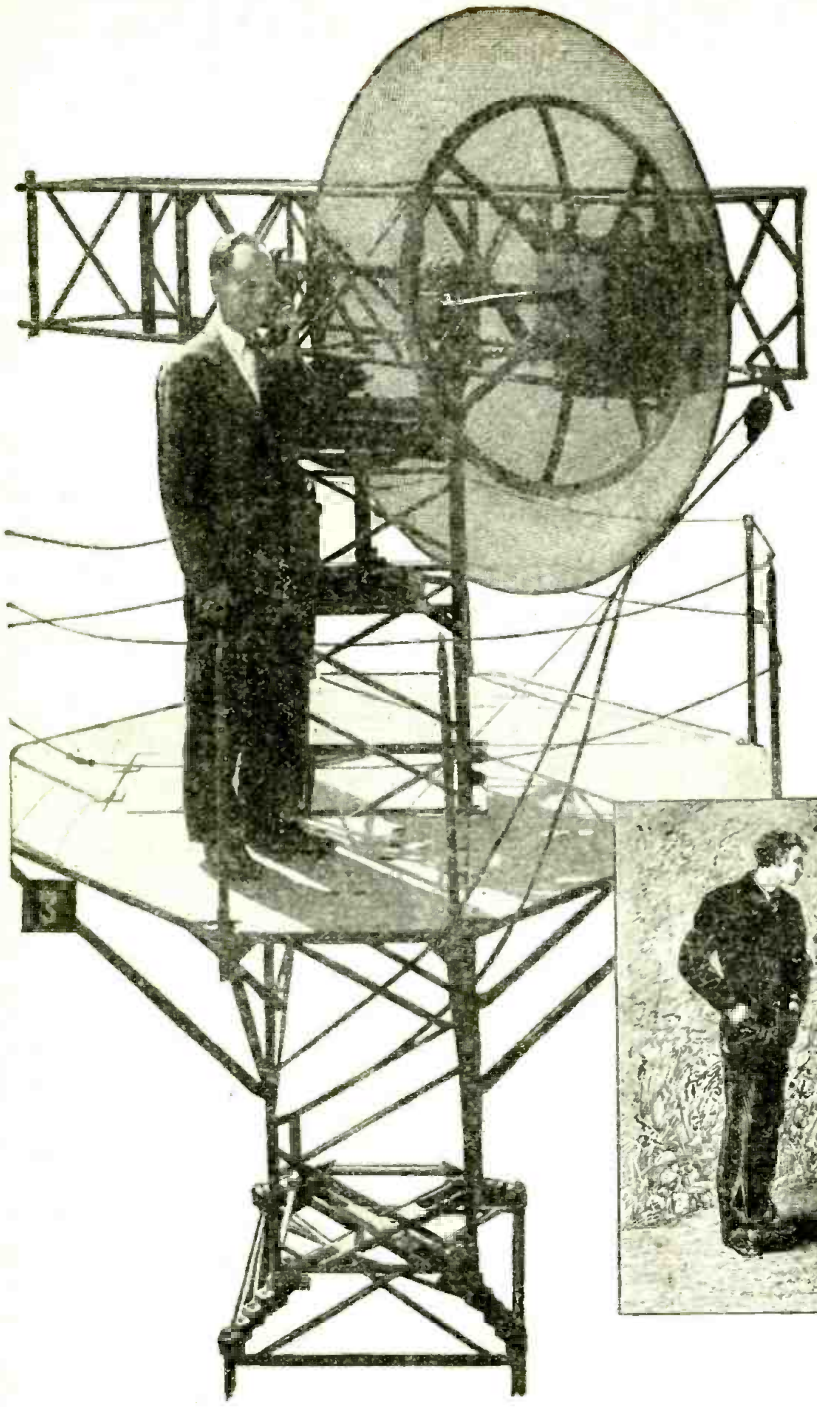
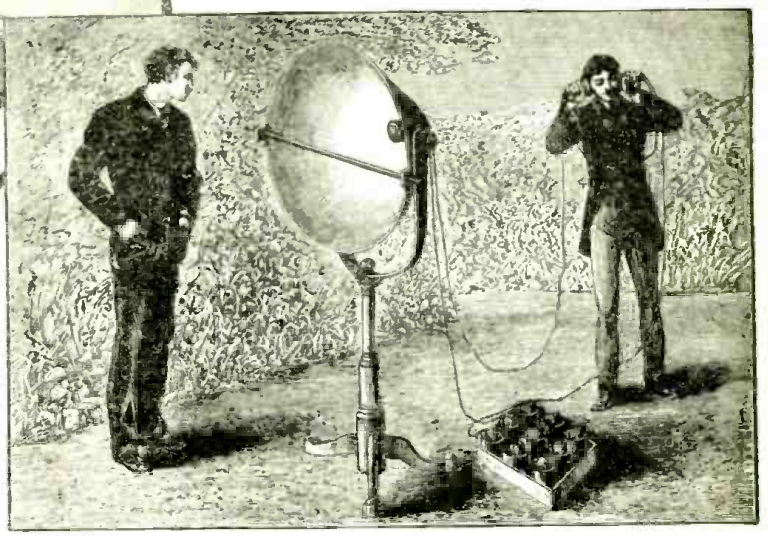
131-C Liberty St., WH 4-7658, New York City 7, N. Y.



Words that rode on a beam of light



If Alexander Graham Bell could look at the microwave antenna in the illustration, how quickly his mind would go back to his own experiments, 67 years ago!

For in 1880 the inventor of the telephone had another new idea. Speech could be carried by electric wires, as Bell had demonstrated to the world. Could it be carried also by a *light* beam?

He got together apparatus—a telephone transmitter, a parabolic reflector, a selenium cell connected to headphones—and "threw" a voice across several hundred yards by waves of visible light, electromagnetic waves of high frequency.

Bell's early experiment with the parabolic antenna and the use of light beams as carriers was for many years only a scientific novelty. His idea was far ahead of its time.

Sixty years later communication by means of a beam of radiation was achieved in a new form—beamed microwave radio. It was developed by Bell Telephone Laboratories for military communication and found important use in the European theater. In the Bell System it is giving service between places on the mainland and nearby islands and soon such beams will be put to work in the radio relay.

In retrospect, Bell's experiment illustrates once again the inquiring spirit of the Bell System.

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EXPLORING AND INVENTING, DEVISING AND PERFECTING FOR CONTINUED IMPROVEMENTS AND ECONOMIES IN TELEPHONE SERVICE

"BRAND" vs. "ORPHAN" RADIOS

The Trend of Price Structure in Radio Receivers

By HUGO GERNSBACK

IN the postwar readjustment period through which the country is now passing the radio industry is caught in the same predicament as are all other major industries in the United States.

Due to the high price scales engendered by World War II in the period between 1941 and 1945, the cost of commodities has practically doubled in the United States.

The inheritance of the war price structure is still with us and it will take considerable readjustment before a lower price trend can again be achieved.

In the meanwhile, the many strikes which have upset the economy of the country have taken their toll with the result that earning power has been curtailed while the savings of the workers have approached the vanishing point. The consequent result is that the purchasing power of a large cross section of the country has been seriously impaired. The high prices on almost all commodities make it impossible for the greater proportion of the population to buy the merchandise which it needs, simply because the spiraling price structure outruns the earning power of most of the population today.

As in previous periods of inflated prices, the increasing resistance of the buying public has reached its zenith, so that only necessities are bought now. This resistance will continue until the adjustment period has run its course.

It would also seem that for the time being general wage increases have largely ceased. We will probably soon see stabilization of wages and income of the greater percentage of the population in this country, and a downward adjustment of our price economy, thus making further immediate wage increases unnecessary.

All of this is not new—it has been said time and again by many economists in the country—and it would seem reasonable to expect a slow downward trend of prices from now on. There may be exceptions in certain luxury and allied industries, but the main industries know now that the public will not come back into the market to buy merchandise on a large scale until it is priced within the reach of the majority in this country.

The radio industry is in no different position than other major industries and the buying resistance of the public necessarily includes radio as well.

The radio industry, too, is in the unhappy position where their cost of raw materials, parts, etc., and their wage cost is at an all-time high, making it at the moment almost impossible to reduce prices without bankrupting many firms. All materials, parts—with practically no exceptions—that go into the making of radio sets are at present—or have been until very recently—at their peak. The same is true of wages, coupled with the fact that the daily output of the workers nowadays is generally below what was par before the war—thereby distorting and aggravating the condition still more.

The radio industry is further bedeviled by the additional fact that many of the newer and smaller firms could not withstand the buying resistance of the public and had to throw upon the market, in some of our large centers, a certain amount of distress merchandise. This aggravated the picture because when large advertisements appeared in the newspapers offering radio sets and other radio merchandise at greatly reduced prices, the public jumped to the conclusion that all reputable set makers would have to follow suit and reduce their prices drastically, too. This, however, did not prove to be the case, because many of the responsible old line manufacturers prefer to fill warehouses with radio receivers rather than see them sacrificed below their cost.

That the large radio set manufacturers mean what they say, and that they do not contemplate immediate large scale price cuts, is best reflected in a statement from one of the foremost radio manufacturers, who issued the following statement just as we go to press:

"Although no price changes are contemplated at this time, should any downward revisions in price occur, the trade is guaranteed to be rebated for the difference between old and such new prices as may be established."

According to this manufacturer, the guarantee covers inventory purchased (by dealers) between the dates of April 15, 1947, and December 31, 1947.

Dealers and radio stores in many U. S. centers, it appears, have recently refused to buy standard brand radio receivers because they were afraid to stock up on leading brand radios which they felt would soon be offered at reduced prices.

It is necessary, however, for the dealer and storekeeper to stay in business, therefore he must buy some merchandise that he can turn over quickly. So we now have again—as even before the war—the spectacle of a number of "no-name" brands, that is, unknown radios which are flooding the market in earnest. Such receivers are often advertised for less than one-half of the price of established makes of similar sets. That does not mean that they compare in quality or performance with standard brands. Often many of these unknown receivers are thrown together by so-called "bedroom manufacturers," of whom there are many hundreds in this country.

These firms buy war surplus and other odd parts wherever they can be found at a price. The cabinets usually are of unseasoned, poor-grade wood and will not stand up for long, nor will the sets themselves. There is very little real factory inspection or testing and as the parts are not uniform and often include discards and "seconds," the receivers themselves are no bargain no matter at what price they are offered.

(Continued on page 71)

RADIO-ELECTRONICS

Items Interesting to

RADAR has been recognized by the Civilian Aeronautics Administration, it was revealed last month. In a move which came as a distinct surprise even to its own field personnel, the Administration abandoned its previous policy of considering ground-controlled approach (GCA) radar as a purely "supplemental" aid in bringing aircraft safely to earth under instrument-weather flying conditions and issued instructions that it may be used as a primary airport-approach system.

This step reverses the CAA's steadfast earlier assertion that its new GCA installations at La Guardia Field, Washington and Chicago would be utilized only to monitor aircraft let-downs through soupy weather on the CAA-sponsored runway localizer and glide-path radio beams of the instrument landing systems (ILS) already installed at dozens of air terminals throughout the country.

TWO AND ONE-HALF WATTS is the power of an FM broadcast transmitter installed last month at Syracuse University. Proposed, designed and installed by General Electric Co. as a suitable unit for intra-mural broadcasting, it was the object of a special visit of the FCC. The Commissioners are very much interested in the possibilities of low-wattage university and school FM transmitters.

The establishment of such stations on a wide scale would encourage adult as well as student educational programs and at the same time help train thousands of students to help meet the demand developing for commercial FM station personnel. With this low-powered equipment designed to assist educational institutions with limited budgets, it will be possible for many schools to get on the air quickly, G-E officials believe.

Under average conditions, the range from the antenna point of the transmitter provides an excellent signal in all parts of the campus and surrounding student living centers.

FM broadcast stations would continue and extend the present system of "college broadcasting" now carried on at several universities with carrier current sent over electric light lines, or in some instances, over heating-pipe systems.

TELECOMMUNICATIONS proposals by the United States delegation to the International Telecommunications Conference now meeting at Atlantic City, New Jersey, include a Central Frequency Registration Board and permanent International Radio Consulting Committee. A third proposal is that emergency administrative radio conferences be called from time to time to deal with problems requiring immediate solution.

The Atlantic City conference, which opened May 12, has the important work of revising the Cairo Radio Regulations, which have guided international radio since 1938. Fifty-nine nations and 8 international organizations were represented by more than 700 delegates at the opening session.

SUPERSONIC LAUNDRIES may be the next development in the application of this branch of electronics to industry. The possibility of using supersonics to wash clothes was revealed last month by Sir Edward Appleton, radar pioneer and secretary of the British Department of Scientific and Industrial Research.

Sir Edward explained that dirt is held to fabric by electrical attraction. Soap and other solutions, called detergents, are now used to break this electric bond.

If current research is successful, sound waves may do part of the job in the future. Supersonic vibrations are being used to shake out the dirt particles from clothes. Sound waves would also emulsify the dirt in the cleaning solution to keep it from getting back on the clothes.

The idea of using sound waves to shake dirt off fabrics was developed from the wartime Asdic submarine detector.

COMMUNICATIONS EXPANSION in the vehicular field is so great as to require a special FCC hearing, which has been set for September 8.

The common carrier type of mobile service is expanding at "a very rapid" rate, according to the FCC. Common carrier highway service is proposed for 79 cities on the mainland of the United States and 2 in Hawaii. More than 3,000 mobile units have been authorized in this class. These, together with associated fixed stations, represent an investment of $4,500,000.

Urban automobile units have been authorized to the number of 5,600 mobile units. The total investment in this field is estimated at about $6,000,000. Expansion is delayed only by the inability of radio manufacturers to furnish the equipment as rapidly as desired.

The telephone companies propose service which will extend wire telephone communication to and from land, sea or air vehicles. Three types of common carrier service are in prospect: communication between any regular telephone and any mobile unit; special two-way dispatch service between a central office and specified mobile units; and a one-way signalling service to mobile units.

RADIO LISTENERS want high fidelity, Dr. Harry F. Olson, head of the acoustics research laboratory, RCA Laboratories, reported last month. Recent surveys have shown that persons listening to music reproduced through standard amplifiers and loudspeakers have not expressed preference for full-range fidelity. Therefore certain engineers hold that full-fidelity is unnecessary and merely adds to cost of amplifier equipment. Others hold that the results were not due to a preference for restricted-range music, but to imperfections in reproduction which were reduced with reduction of audio frequency range.

Dr. Olson made his tests with a live orchestra, eliminating all factors which could be introduced by amplifiers or speakers. Tests showed that 75 percent of listeners between 30 and 40 years of age preferred full frequency range. Among listeners between 14 and 20, only 59 percent expressed an appreciation for the unrestricted tonal range in this classification of music.

"The listeners in the latter age group," Dr. Olson stated, "are probably influenced by listening to radios, phonographs and juke boxes rather than orchestras and are, therefore, conditioned to a restricted frequency range."

The 2½-watt transmitter used to broadcast to the student body of Syracuse University.

MONTHLY REVIEW—

the Radio Technician

Radio Items of the Month

Smallest 3-way portable is claimed by Sentinel, who offer a 4 x 4 x 8-inch model. The little set is a regular 5-tuber and its performance is comparable to that of larger radios, according to the manufacturer.

Radio life-saving is not all confined to the distress frequencies, says Arthur Magee, New Jersey commissioner of motor vehicles. The 15-minute broadcast series, *Highways to Safety*, released by NBC, coincided with the lowest fatal accident record for 14 years. The radio broadcast "has undoubtedly contributed materially to this record," says Mr. Magee.

David Sarnoff, president of the Radio Corporation of America, was elected president of the Army Signal Association at its recent Fort Monmouth convention.

Milk and beer are now being pasteurized by a continuous radio heating process. Milk treated electronically, says George Brown, RCA electronic heating authority, keeps longer than ordinary pasteurized milk.

Russia plans to put 28 new and powerful radio stations into operation by 1950, the Soviet Communications Minister reported last month. He announced that Russia is now broadcasting in 30 foreign languages, as well as 70 languages used by native inhabitants of the Soviet Union.

Overproduction of small table models and erratic parts production have been the cause of dealers' and manufacturers' greatest difficulties, delegates and visitors at the recent Radio Parts Show were told.

Estimates of FM receiver production for 1947 were revised downward last month by the RMA. Earlier estimates of 2,500,000 for the year were cut to "between a minimum of 1,800,000 and a maximum of 2,100,000 receivers."

Columbia's color television research is being "drastically curtailed" according to company officials. The company will concentrate its efforts for the time being on actuality broadcasts such as sports and special events.

Heavy water, used in atomic research, and so important a short time ago that military expeditions were organized to destroy places where it was prepared, can now be bought by research workers from the Atomic Energy Commission for $15 an ounce—less than the price of high-class perfume.

GEORGE P. ADAIR resigned last month as chief engineer of the FCC, after an association with the Commission which goes back to 1931. He will open offices as a radio engineering consultant.

Accepting his resignation with regret, the FCC announced the appointment of George Sterling, wartime head of the Commission's radio intelligence division, as Mr. Adair's successor.

SIGNAL STORAGE used during the war by the Germans enabled them to transmit a dozen or more messages in a fraction of a second, the U. S. Department of Commerce reported last month.

Submarines in dangerous waters would surface and send all their messages almost instantaneously, cutting down the possibility of being intercepted and located by Allied search equipment. The signals were received, amplified and projected on the screen of a cathode-ray tube of such persistence that *the signals could be stored for three weeks*, if necessary. Thus the receiving operator could decode them at his convenience.

The cathode-ray tube beam in this equipment scans a plate covered with a layer of microscopic quartz particles embedded in a photoelectric base. The particles of quartz hold a charge which varies with the intensity of the scanning beam. To transcribe the stored record, a beam of "black light" (either infra-red or ultra-violet, to avoid visible interference) is directed at the photoelectric plate, each point of which emits electrons in proportion to the charge at that point.

An electromagnetic lens focuses the streams of electrons on a fluorescent screen. The screen produces a visible image, which will last for 15 minutes, provided the plate continues to receive ultra-violet or infra-red light. However, the image can be wiped off the screen at any time by turning the light off and scanning the plate with an electron beam of uniform intensity.

PRICE CUTS on radio receivers have been announced by two leading companies. Emerson has announced reductions ranging from $3 to $20 on 9 table radio and radio-phonograph combinations. Admiral has slashed prices from 20 to 25 percent on its new 1947 line of receivers.

The trend was also noted by Majestic Brands, Inc., in a message to dealers which assured them that they would be rebated should the prices be reduced on any Majestic models authorized or established by the company or by Majestic Radio and Television Corporation.

In other quarters denials that radio prices are headed lower were heard, though it is generally admitted that prices may fall on the present stock of small table radios, which were produced in large quantities in 1946. Console models are largely unaffected by the price cuts, which have hit most directly on table and portable models.

MAJOR GENERAL H. C. INGLES, who served as Chief Signal Officer of the United States Army from July, 1943, to March, 1947, has been elected President and a Director of RCA Institutes, Inc., it was announced last month by David Sarnoff, President of the Radio Corporation of America.

As Chief Signal Officer, General Ingles was responsible for the Army's world-wide communication system, the enormous supply program of communication and electronic equipment for the Army, as well as the Signal Corps' research and development program.

His achievements in these fields of military activity are revealed in the citation accompanying the award of the Oak Leaf Cluster.

The citation reads, in part:

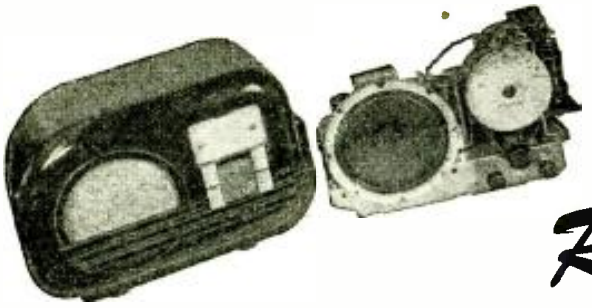
"Technological advancements made under his direction included the application of communications equipment in the field of psychological warfare, the development of radar to a degree which made it one of the most formidable of weapons, the use of radio relay systems for bridging inaccessible terrain, the use of panoramic reception for visual monitoring over wide frequency spectrums, the adoption and extensive use of teletypewriter and radiophoto transmission equipment and the use of highly secret cryptographic systems."

PRODUCTION of FM-AM receivers rose notably in the first part of the second quarter of 1947, the Radio Manufacturers stated last month.

The April output of FM-AM receivers was 47 percent over that of March, and production of all types of receivers in April was higher than during the first quarter of the year.

Estimated April production of all types of receivers by Association members was 1,548,540, compared to 1,337,269 in March. Total for the 5 weeks from March 31 to May 2 was 1,759,723, the highest 5-week production in all the history of radio.

Television receiver production for April was 7,886, an increase of 1,247 over the figures for March.



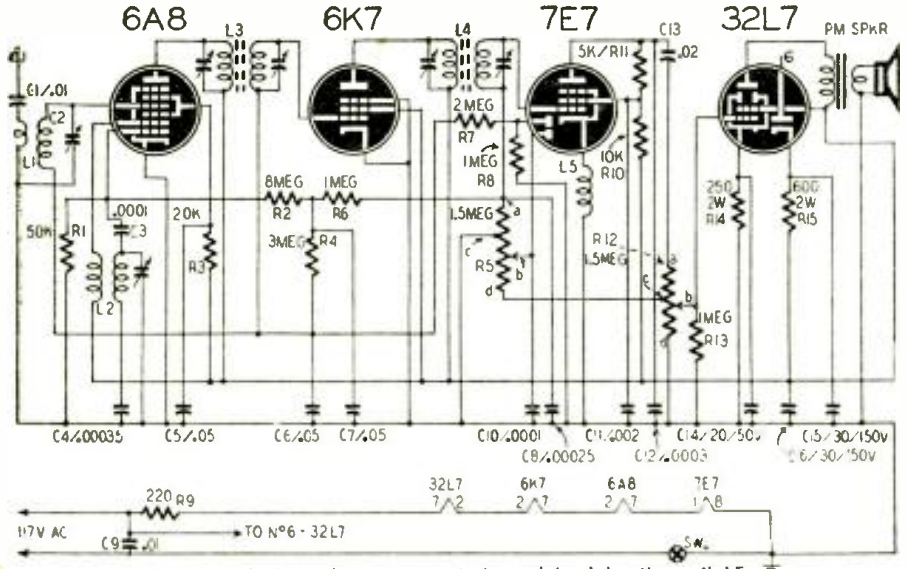
This receiver embodies a new and rarely-used principle in its method of detection.

Reflexed Four-tuber

By W. T. CONNATSER

YOU will enjoy building and be proud to own this 4-tube superheterodyne a.c.-d.c. broadcast receiver. A minimum of other parts are necessary for its construction, and the total cost should not be too high. Or an existing set may be readily modified to include the excellent second detector feature of this set.

The set is built around the demodulator (second detector) circuit invented by Frederick C. Everett (Pat. 2,361,616). This patent was reviewed by RADIO-CRAFT in the March, 1945, issue.



The queer-looking detector circuit is explained by the coil L5.

A complete set is here shown by way of illustration, and the hookup and functions of the 7E7 and directly related components are explained in some detail.

Obviously the 6A8 converter, the 6K7 i.f. amplifier, and the beam-power amplifier and B-voltage rectifier may be replaced by any combination of tubes that will serve the purposes. As to the 7E7, it may be replaced by any triode or pentode with the necessary diodes. The inventor of this detector circuit expresses a preference for a medium-mu triode of the 6SR7 type.

In the set here shown a high-impedance-primary antenna coil L1 is used, which permits the use of almost any length of antenna. Of course a single coil or a loop may be of use instead.

The oscillator coil L2 is standard for the 6A8 type tube, and is for 456 kc.

The padding condenser C4 generally will be about 350 μμf for a 456 kc i.f. It is best to use an adjustable padding condenser. The tuning condenser C2 is a 360-μμf, 2-gang unit.

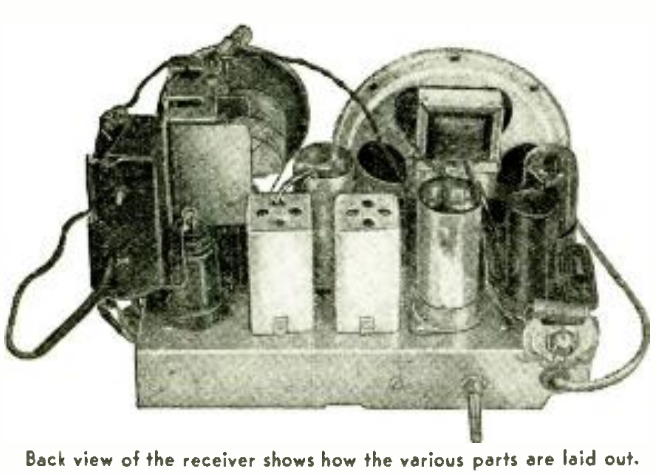
The hookup of the 32L7 beam-power amplifier and B-voltage rectifier is conventional for this tube.

Negative bias for the 6A8, 6K7, and 7E7 tubes is obtained from the oscillator grid of the 6A8 tube (pin 5). This negative voltage is dropped through R2 to bias the 7E7 tube, and is further dropped through R4 to the a.v.c. circuit to bias the 6A8 and 6K7 tubes.

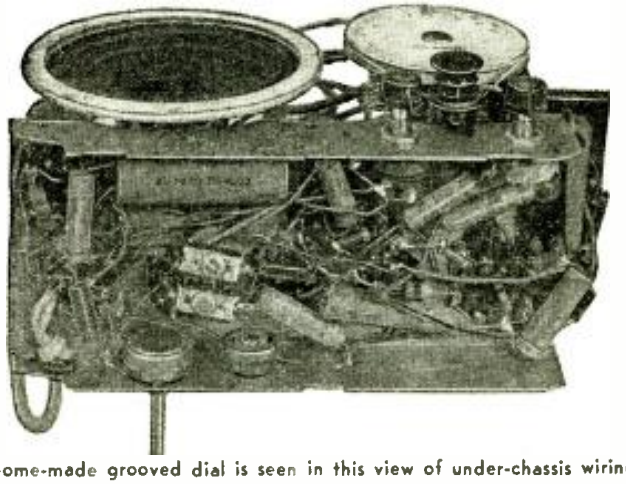
The a.v.c. circuit is made up of R7 and R8, and C6 and C7, and is of superior type. This network maintains volume level as between stations of varying power, and is designed to filter out all pulsating currents that might prevent good reception. Furthermore, it effectively isolates the 6A8 and 6K7 tubes from the 7E7 tube.

The resistors R10 and R11 in the plate and screen-grid circuits of the 7E7 tube, because of the low B-supply available in this set, must be the lowest values that will not short-circuit the audio signal or adversely affect the tonal quality of reception when the reflex feature of the set is used. When no reflex is to be used at all, then the plate of the 7E7 tube is connected directly to the B-plus supply and by-passed to ground by a large condenser, and an appropriate resistor and by-pass con-

(Continued on page 66)



Back view of the receiver shows how the various parts are laid out.



Home-made grooved dial is seen in this view of under-chassis wiring.

FRENCH RADIO-MODEL AUTO

By SIMON COUDRIER

RADIO CONTROL is coming sharply into the public eye. We hear much in the press of radio-controlled airplanes, as at Bikini. Directed by radio at a distance of several miles, they were able to penetrate the radioactive clouds created by the explosion and furnish important photographs of yet little-known phenomena.

The factories in which the atom bombs were built also used electronic remote controls which permitted operating dangerously radioactive equipment from a distance of several hundred meters.

During the war of 1939-1945, remote-controlled radio tanks and planes appeared in action. Pocket tanks used by the Germans at Anzio served as mobile mines and moved at a speed of more than 15 miles per hour. Radio-controlled planes and rockets were used by both sides in the latter stages of the war.

Finally, the Americans developed jet-propelled aerial torpedoes. These were released from airplanes, whose occupants could direct them to their target by modifying their trajectory with radio-controlled rudders.

Already in France certain laboratories are studying pilot models of remote-controlled vehicles and planes. and several amateurs have constructed models of boats, gliders. and automobiles which are controlled by radio.

A photograph shows a radio-guided automobile built by a 28-year-old French constructor, M. Roveyaz. Of reduced dimensions, it can be driven in any desired direction, at a speed which can be controlled remotely by the operator with the radio transmitter shown in the photo on page 74.

This development is all the more interesting because of the small size of the model. The reader will understand that the difficulties increase in inverse ratio to the size and weight. With such an automobile, weighing 5½ pounds and measuring 20 inches, it is hardly possible to send it out to the country to forage for food, while the owner, defying the thunders of the Minister of Supplies, remains comfortably seated in an armchair. and controls the speed and direction of his machine by pushing several buttons on the small box! But anything that is possible for a model of reduced dimensions is an assured fact for an automobile of normal dimensions, except of course the question of visibility. This makes it necessary to perform all maneuvers within the field of vision (up to 600 meters in our case) and often makes it preferable to employ a model of a ship maneuvering on a large body of water (or the edge of the sea) or a remote-controlled plane. There can be no question of installing a televisor on a small model.

Construction of the model

Before studying M. Roveyaz' model, let us remember the exact definition of remote control. It is putting into operation at a distance, at the will of the operator, electrical apparatus, without having recourse to wire.

The transmitter works on a wave length between 4 and 6 meters. Experience has shown that medium power (between 15 and 30 watts) is sufficient for reliable radio control up to a distance of at least two miles.

The oscillator section is of the classic type. Fig. 1 is the schematic. It uses two 6L6's, as oscillator and power amplifier, the transformer TR permitting connection of a modulator. Construction of this equipment presents no difficulty, but it demands material of the highest quality.

Another system employed by M. Roveyaz consists in using a 1-tube transmitter, operating on different frequencies, spaced several centimeters apart, and modulated by applying an audio-frequency signal to the grid. Each wave length then corresponds to the control of one operation.

The skilled constructor no doubt will have his own ideas as to the coils and capacitors, but the following may serve as a guide.

L1—6 turns No. 14 wire on 1-inch form spaced to occupy ¾ inch. Cathode tap 1¼ turns from ground end.

L2—5 turns No. 14 wire on 1-inch form spaced to occupy 1¼ inch. Air-wound.

(Continued on page 74)

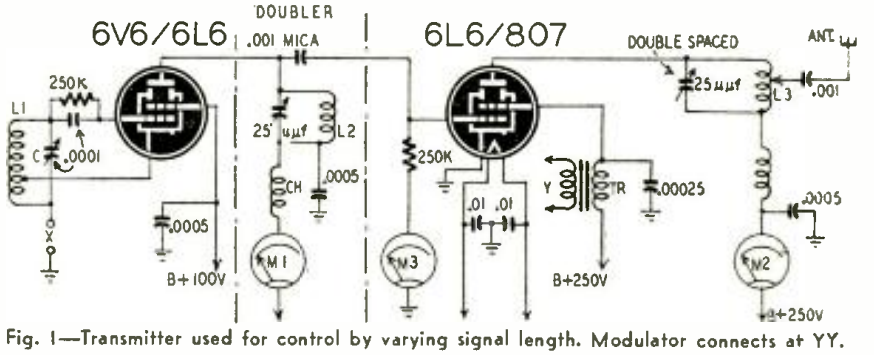
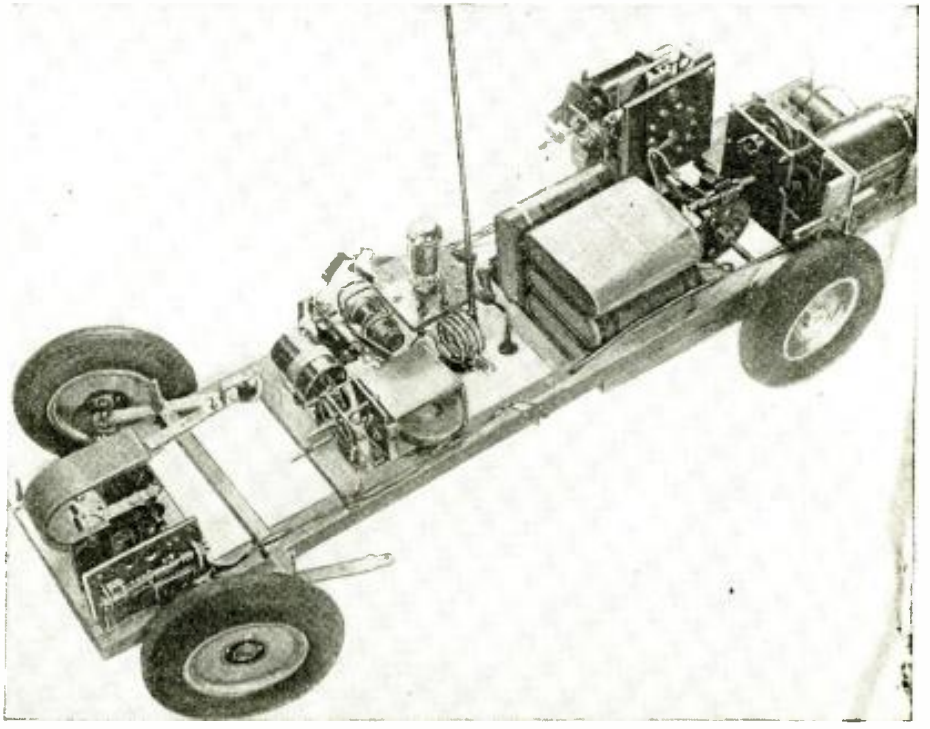


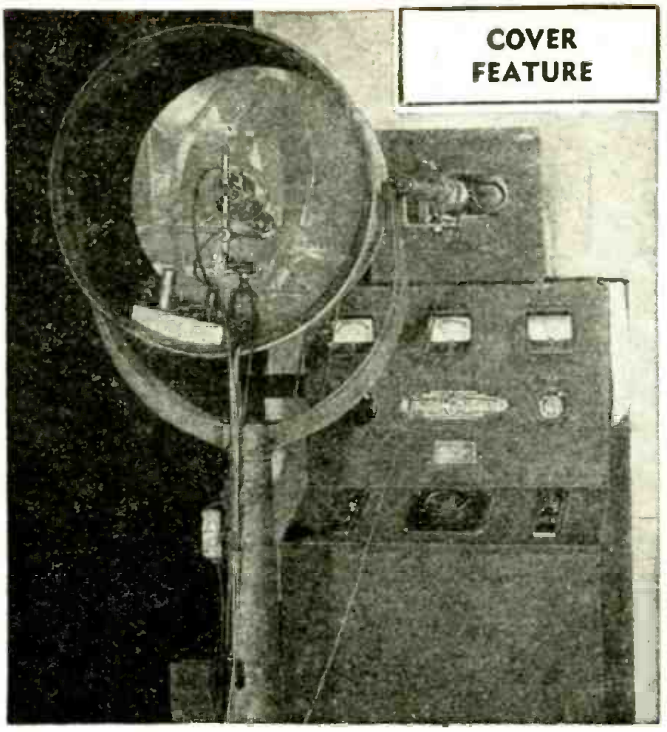
Fig. 1—Transmitter used for control by varying signal length. Modulator connects at YY.

Control is effected at X by a system which, by contactor, key, or automatic telephone dial, gives a signal whose duration is more or less variable, depending on the operation to be performed.



Model automobile chassis used by M. Roveyaz, with the receiver which controls its operation.

COVER FEATURE



The transmitting end. Equipment is mounted in a parabolic reflector, like that of the receiving unit, which is shown on the cover.

Television Over A Light Beam

Light offers several advantages for directional television and radio relays

THE history of radio communication is a history of ever-higher communication frequencies. The recently announced color television service, for instance, will work in a band near 500 megacycles. But the apparatus pictured on our cover operates at *600 million megacycles!* Yes, the only possible explanation is, of course, that these ultima-frequency waves are ordinary light!

Light offers several advantages for short-distance relaying of television programs, and possibly for other forms of dispersing information ordinarily carried by radio. Among the most important for television is the elmination of "ghosts," which are one of the worst bugbears at lower frequencies. Another advantage is privacy. Because of the extremely directional qualities of a ray of light, the beam can be focused exactly on the receiving station, with no possibility of undetected interception. This would adapt it especially to such applications as the transmission of television programs from a central light transmitter to local motion picture theaters for showing of pictures on ordinary theater screens. This even could be done in such complete secrecy that the program could not be snatched off the air by unauthorized receivers.

Although first demonstrated last November before the American Institute of Electrical Engineers and more recently before the Federal Communications Commission, photovision as a basic idea harks back to 1934 when Patent No. 1,984,673 was issued to Allen B. DuMont of Upper Montclair, N. J. That invention relates to electro-optical systems and particularly to a direct-vision television system quite independent of the usual electrical transmission channels. A feature of the invention is the employment of a high-powered light source which is capable of being viewed over very long distances. The light source is modulated in accordance with television signals. Originally Dr. DuMont proposed to use a high-powered light source such as a water-cooled neon lamp, or a lamp such as used for aerial beacons, and which are visible over distances of 15 to 25 miles.

With the advent of electronic scanning, Dr. DuMont revised his invention to transmit electrical signal-element equivalents in the modulated light beam. Such signal elements now comprise not only the modulation for lights and shadows of the image, but also the synchronizing pulses for the positioning of the pictorial lights and shadows on the usual cathode-ray screen, with the full wealth of detail which electronic scanning provides.

Special cathode-ray tube

Since the practicality of *photovision* depends on a powerful yet highly responsive light source, it was necessary to develop an entirely new type of cathode-ray tube with a fixed, intensity-modulated beam that can be varied up to 5,000,000 times per second. In the earlier demonstration equipment, this tube, which is shown at right, produces a dull, light-green spot less than 1 inch in diameter on the fluorescent screen. It seems uncanny that such a source of illumination can transmit television pictures over considerable distances despite full daylight.

The phosphor used for the screen is one of the new materials under development, with a delay time less than one-tenth that of calcium tungstate (P5), heretofore the fastest standard phosphor. The screen has an efficiency approximately 60 percent that of the highly brilliant P1 phosphor. However, other phosphors capable of producing far greater intensities—yet with the necessary extreme response speed—are under development in the DuMont Laboratories for the full commercialization of photovision. In addition to greatly increased illumination levels for the transmitter light source, more refined optical systems such as critical reflectors or veritable searchlights, as well as focusing lenses, are now under consideration.

Receiving equipment

The receiving end of photovision is a simple photoelectric cell on which the intercepted light beam is focused, by suitable reflectors or lenses or mirrors. This cell converts the modulated light beam into corresponding electrical values which reproduce the transmitted image on the usual cathode-ray tube screen. Thus the receiving circuit is vastly simplified by the elimination of all r.f. and i.f. stages because the output of the photomultiplier cell or tube is sufficient to modulate a picture tube directly.

The photocell is shown in its position inside the large parabolic mirror housing. The strange distorted reflection in the mirror will be seen to be the image of the cell's shield can, with its corrugated top and lens opening on the inner side.

The special cathode-ray transmitting tube.

The equipment shown on the cover and in the photographs represents a workshop setup, and the informal position of the photocell is due to the fact that it is most convenient not to have a permanent mounting during experimental work. Amplifier equipment, mounted for convenience in the transmitting parabola housing, of course would be mounted below or behind the parabola in permanent equipment.

Meanwhile the sound component of

(Continued on page 69)

NARROW-BAND FM FOR HAM RADIOS

By NORMAN L. CHALFIN

MANY amateur operators are now transmitting narrow-band FM. These signals are received on a standard AM communications receiver by "side-slope" detection. The receiver is detuned slightly, so that the frequency modulation results in changes in the received amplitude of signal. A positive frequency shift causes a rise in amplitude and a negative frequency shift causes a fall. This pickup method does not permit making a center-frequency adjustment or measurement. To tune the signal "on the nose" a discriminator is required. Discriminators can be built into or attached externally to standard communications receivers. For the more ambitious operator these can be wired in with a change-over switch.

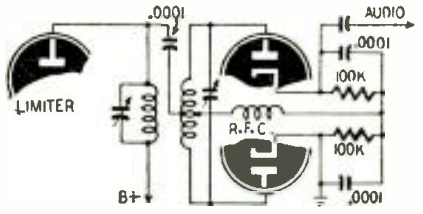


Fig. 1—Standard Foster-Seeley discriminator.

The circuit diagram of a standard Foster-Seeley discriminator for FM detection is shown in Fig. 1. It can be made from a center-tapped-secondary i.f. transformer. Several of these full-wave detector i.f. transformers are on the market. In place of the duo-diode a pair of germanium crystal rectifiers may be used. The Sylvania 1N35 is a commercially available matched pair of crystals. These are quite expensive. Two 1N34's can be used as well and are available at lower cost. The 1N21 works equally well.

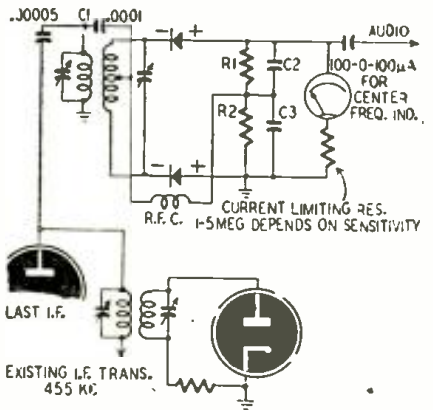


Fig. 2—Discriminator circuit with crystals.

The discriminator with crystals is shown in Fig. 2. The major advantage of the crystals is that they may be wired right into the transformer can along with all other components as shown in the photograph.

Attachment to receiver

One of these discriminator units may be wired into a commercial receiver in one of several ways. The unit may be wired to a capacitor coupled to the plate of the i.f. amplifier tube just preceding the detector as shown in Fig. 2. A double-pole, double-throw switch is installed, with the arm connected to the audio input so that it can be switched from the AM to the FM detector outputs, as shown in Fig. 3. The AM detector wiring should be left intact so that the a.v.c. is retained. The second pole of the switch can be used to ground the cathode of the last i.f. tube. This will give some limiter action. Also some limiter action was obtained with the switching circuit of Fig. 3 on a commercial receiver when the last i.f. amplifier tube was changed from a remote cutoff type to one with a sharp cutoff. The replacement for a 6SK7 would be a 6SJ7. For the 6SG7 a 6SH7 may be substituted. Any of the above changes will necessitate realigning the i.f. transformer to correct for the detuning caused by the new circuits. A more ambitious project would be to add a limiter stage and make the add-on unit a 2-tube affair. Or, if a crystal diode discriminator is used, there would be but one tube plus the diodes. A circuit arrangement for this is shown in Fig. 4. The arrangement is left to the ingenuity of the individual user, because the physical position will vary with the sets into which they may be connected.

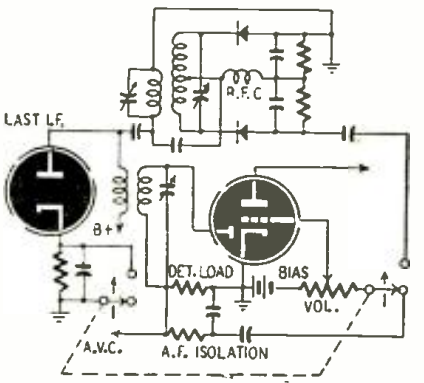


Fig. 3—Application to a standard receiver.

Tuning indicator

A center-frequency indicator can be connected to the detector system, as shown in Fig. 2. This would be a zero-center meter with about 100 or 200 microamperes on either side. When the carrier center frequency is tuned to the intermediate frequency the meter indicates zero. The limiting resistor is inserted to prevent banging the sides of the meter with high-level signals.

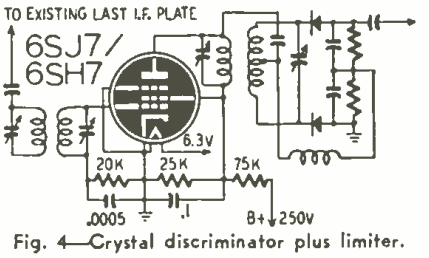


Fig. 4—Crystal discriminator plus limiter.

If a center-tapped i.f. transformer is not available a single-ended transformer may be used, as shown in the basic circuit of Fig. 5. This arrangement obtains its center-tapping through two resistances and application of the r.f. energy from the preceding tube plate to both ends of the secondary winding. This method of discriminator connection is employed where slug tuning is desired. The parts values, other than those shown in previous figures, appear in Fig. 6. Where a band-pass type of transformer is available, its tertiary winding can be employed for the discriminator connection, as in Fig. 7.

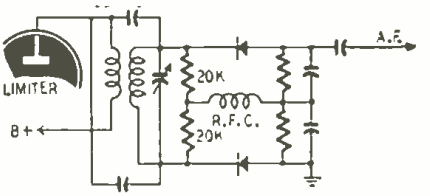


Fig. 5—How to use a single-ended secondary.

(Continued on page 72)

COLOR TELEVISION

By
H. W. SECOR

PART II

ONE of the most interesting color television devices recently demonstrated to the FCC is the *Trichromoscope* picture tube. This tube, as Fig. 1 and Photo A show, has three electron guns built into it, one gun for each color to be reproduced. This tube may solve many problems for the simultaneous as well as the sequential devotees. For one thing it does away with the fussy adjusting and focusing of three separate cathode-ray tubes.

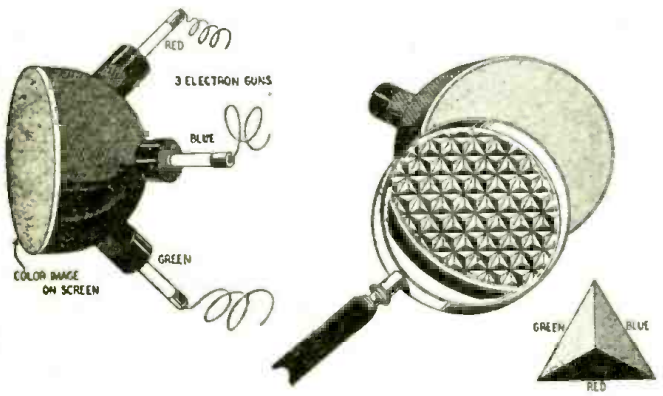


Fig. 1—The Trichromoscope. Magnified section is interior of screen.

Each color signal is injected into its respective gun in the new Du Mont tube. The three electron beams from the guns are focused on a special prismatic screen at the large end of the tube. The screen has myriads of small facets, shaped like pyramids, pressed on its inner surface. Each of the three sides of a pyramid is treated with a phosphor coating that glows with a different color (red, blue, or green) when the beam from its associated gun strikes it. When the various prismatic surfaces are illuminated by the modulated electron beams, a single image in natural color results. Pictures of great brilliance and contrast are produced by this tube, says Dr. Du Mont.

A unique method of producing color in movies from black and white images (on positive film) was recently demonstrated before the FCC in Washington by the inventor of the system, Richard Thomas. Its adaptation to television has not been demonstrated to date.

As Fig. 2-a shows, three black-and-white images are printed on a standard movie film, each image being graded for *color tone* by passing it through accurately balanced, red, blue, and green filters. The three images in each frame are picked up by a Thomas-color *optical color analyzer* unit as shown. For producing color movies, this lens and filter unit restores color to the three images and blends them into a *single* color image, which is projected onto the movie screen.

The optical color analyzer unit comprises a series of lenses, prisms, and color filters, which split the image into three new black-and-white images, corresponding in color value to red, blue, and green. A single lens in the optical color analyzer picks up the image or scene and splits it into three color tone images, which emerge from the unit through three separate lenses, as the diagram shows. See photo B.

At the projection end the action is reversed. The position of the color filter unit is such that the three color tone images picked up are resolved by colored filters into one image in full restored color. This image is projected onto a screen.

In its proposed adaptation to television, the optical color filter unit breaks up the image into three color images—red, blue, and green. These three color images are projected onto the mosaic screen (photoelectric) of an orthicon tube (See Fig. 2-b). Here the electron scanning beam within the tube scans the three images as *one* picture and sends the resulting signals (or modulated current) to the television transmitter.

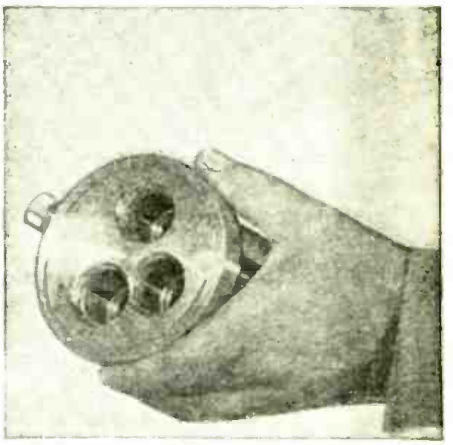


Photo B—The Thomas optical color analyzer.

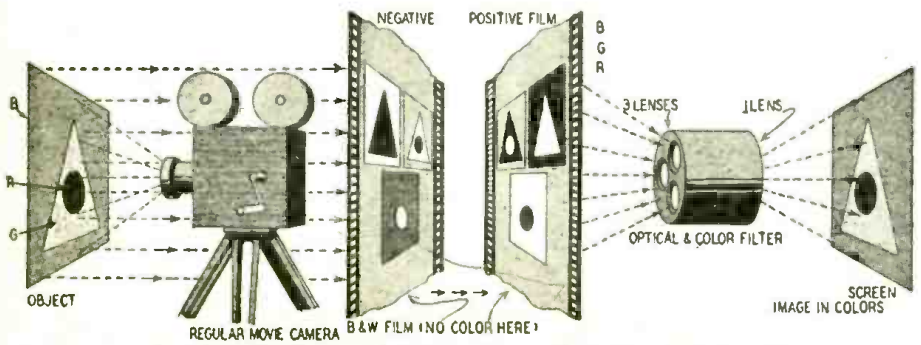


Fig. 2-a—The Thomascolor process, as applied to moving pictures.

The picture is transmitted the same as any ordinary black-and-white television image. At the receiver a kinescope tube reconstructs the image (really three *color tone* or monochrome images; they have no color yet). A second Thomascolor optical and color analyzer unit *(in reverse)* is placed in front of the kinescope screen. As the three images corresponding to red, blue, and green are passed through the analyzer, color filters restore color to each. A single color image emerges from the color unit and is projected onto the screen.

This system should be particularly well adapted to the simultaneous method of television. As the three separation images are spatially identical both in film and on the orthicon tube, they can be scanned as one, permitting transmission on *one* carrier. This means a great saving in the frequency band width required and merits the close study of television engineers.

It should be noted that one of the Thomascolor features is the fact that a movie (in black and white) can be taken of fires and other news events (which could not be approached by cumbersome television trucks). After rapid development of the film, it can be rushed to a television station and broadcast.

Sleeper color system

Two patents have been issued to George E. Sleeper (Nos. 2,389,645 and 2,389,646) on a simplified system of *color television*. Fig. 3 shows the principle of the Sleeper system, which employs black and white transmission up until the time the image is finally thrown onto the receiver screen. This system has not been demonstrated, to our knowledge.

The object is picked up by the *quadchromatic* lens, which projects four images of different color onto the photoelectric screen in the camera tube. These images are scanned in a group as *one* picture, in the same manner as a black-and-white image. The picture (with its four *color tone* components) is transmitted as ONE *black-and-white* image, thus requiring but *one* carrier.

At the receiver a single cathode-ray tube reproduces in *black and white* (with graduated grays, whites, and blacks—tones corresponding to the four respective color values) four separate images, similar to those observed in the camera tube at the transmitter. These images at the receiver lack color so far, but they do possess the correct detail and intensities corresponding to the four colors. Color is now restored to each image (red, blue, yellow, and green) by collecting the images through a color filter and lens system. This unit is arranged in *reverse* order to the one at the transmitter, and projects the four blended color images onto the screen as a single image in full color.

The quadchromatic color filter unit contains lenses for picking up the images (or image, depending upon which function it is to perform—i.e., convert one image into four or the reverse), plus suitable prisms and color filters, as shown in Fig. 4. A simplified sketch of such an optical system, using four separate lenses and four color filters, Fig. 5, will aid in making the process clearer. As in the Thomas color system, no color is actually transmitted, only the shaded or toned images corresponding to each color.

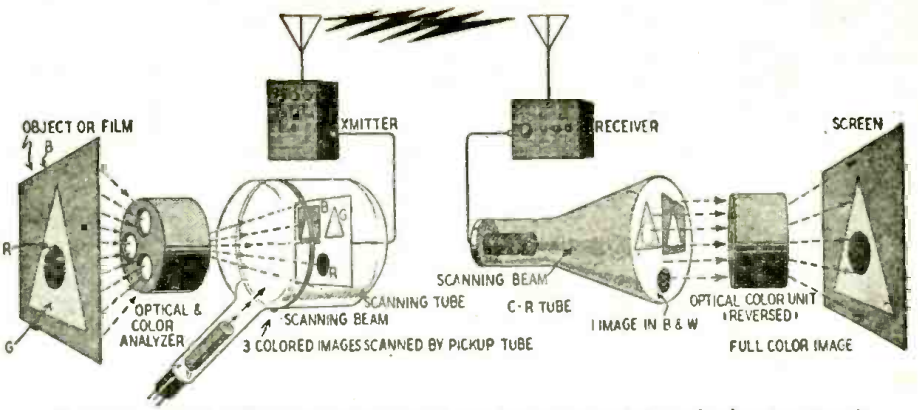


Fig. 2-b—How the Thomascolor process is applied to transmission of television signals.

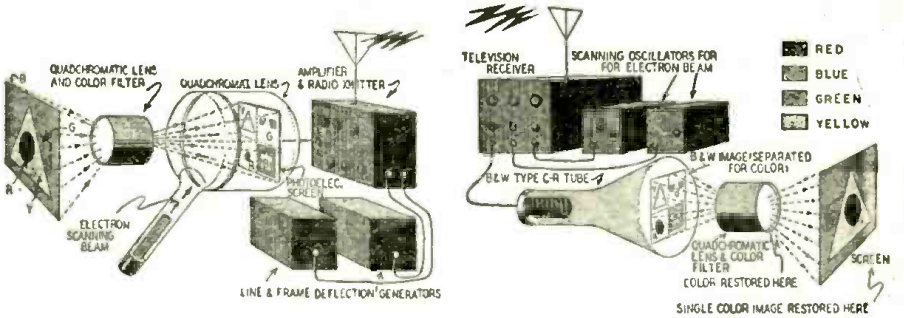


Fig. 3—The proposed Sleeper color television system separates and re-unites four colors.

How the systems compare

Considerable criticism has been leveled at the mechanical (revolving) color filter used in the sequential apparatus. Dr. Goldmark in his report to the FCC illustrates and describes a full electronic sequential system which employs no mechanical color filters.

The 525-line standard proposed in the CBS report to the FCC has been attacked by some experts, who advocate a greater lineage. Here is some interesting data on this controversial phase of color television.

The 525-line image approaches 16-mm movie performance, with regard to the resolution or detail of the image. A noticeable increase in definition is obtainable only at the expense of appreciably greater video band width. As Dr. Goldmark points out, the visual acuity or perception of detail is fairly well saturated at 525 lines. At least 50 percent more lines would have to be added to realize a noticeable improvement in definition. If 750 lines are used, as some advocate, the necessary band width is doubled and each station's channel will be 27 mc wide. The number of channels available in the high-frequency television band will be reduced to 16. If 750 lines and 60 color frames per second are used instead of 48, there will be required a video band width of 25 mc and a chan-

(Continued on page 59)

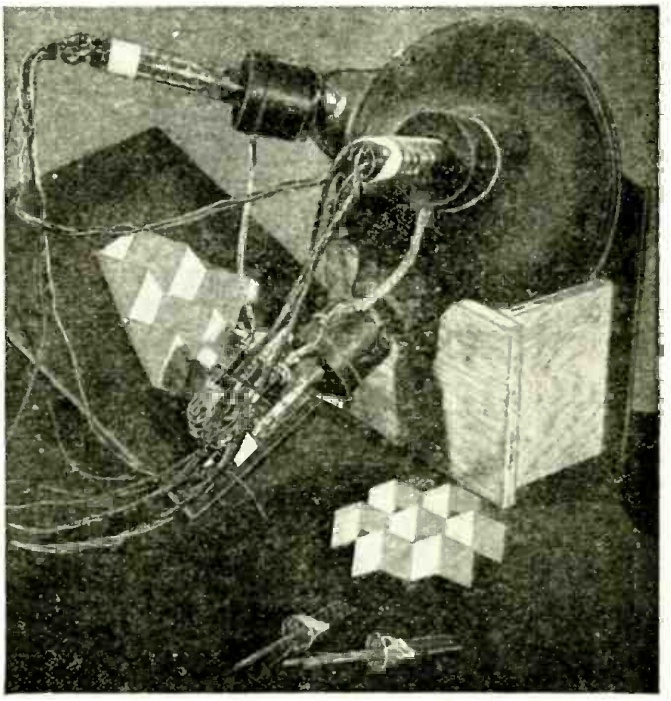
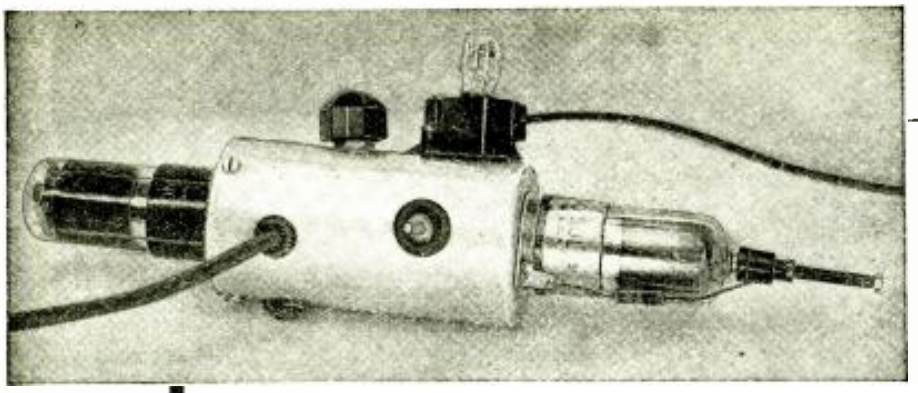


Photo A—Typical Trichromoscope in the experimental laboratory. Cardboard models of the inside screen surface are to be seen nearby.



By M. E. BLAISDELL

"Submarine" Signal Tracer

A PROPERLY-DESIGNED signal tracer can be one of the most valued instruments on the serviceman's workbench. It must be able to indicate the presence or absence of a signal from the antenna to the loudspeaker of any conventional receiver and at the same time locate distortion, hum, and indicate stage gain.

Side view of tracer shows a 1-inch screw extending from the side. This is insulated from the can and is in series with the ¼-watt neon lamp in the conning tower. It is used as an indicator for common ground of an a.c.-d.c. radio receiver. If the neon tube lights when this probe is touched to the chassis or common ground, it indicates that the power plug of either the tracer plug or the set under test must be removed and reversed before attaching the common ground clip to the set.

The small piece of mica cemented to the nut on the probe tip is used on intermittent sets under test. This prevents that sudden shock which inevitably starts the receiver working normally again. When using this mica tip, the serviceman will notice that it works best on positive potentials such as plate and screen terminals — removing the mica tip and applying the probe tip direct on negative terminals such as the grid and a.v.c. grid return path, etc.

Stage gain can be noticed by increasing or decreasing the volume control on the tracer.

A 12J7 is used as the actual probe. The plate, screen, and suppressor are tied together. The tube then functions as a triode, producing little or no distortion to be amplified by the 70L7 pentode. The rectifier section of the 70L7 furnishes the B-power while the A-supply is furnished from the a.c. mains through a 250-ohm line-cord resistor. According to Ohm's law this resistor is calculated to be 233 ohms. However the extra 17 ohms will take care of initial line voltage fluctuations or surges. The body of the tracer is an ordinary vibrator can with the 5-prong wafer type socket removed and an 8-prong wafer type socket inserted in its place for the 70L7 tube. A hole is punched at the other end of the can to accommodate the other 8-prong socket. Each socket is wired and all resistors and condensers especially selected to fit the space, which is small. However, if one uses small 150- or 200-volt condensers, he will find ample room for all components. The photo and figure illustrate the simplicity of construction. As all components are shielded, one does not have to be too careful in instrument design. One word of caution, however: be sure to tape up any connections that are exposed and liable to short circuit.

Make a careful check after the instrument is assembled and before inserting the power line plug. A continuity check of the line cord resistor and each tube in turn to common ground is especially advised. Use ohmmeter test for electrolytics on high resistance by inserting test prods to pin 3 and 4 and ground (counterclockwise from key) outside of set, first removing the power tube. If needle jumps full scale and finally settles past the 1-megohm mark, and filament continuity is O.K., then the instrument is ready to plug in. Check voltage at plate phone jack. It should be around 60 volts—load resistors to minimize distortion and hum are the answer to low voltage at the plate. Every precaution was given to the design for a smooth, even ripple and as close to pure d.c. as possible without seriously affecting the over-all gain of the tracer.

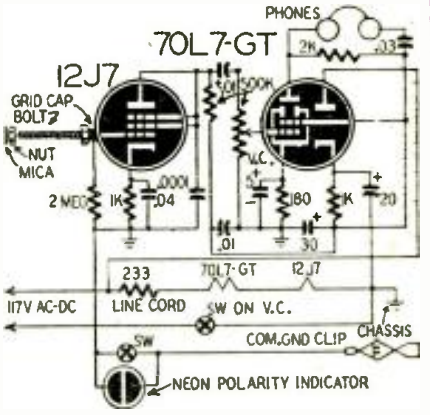
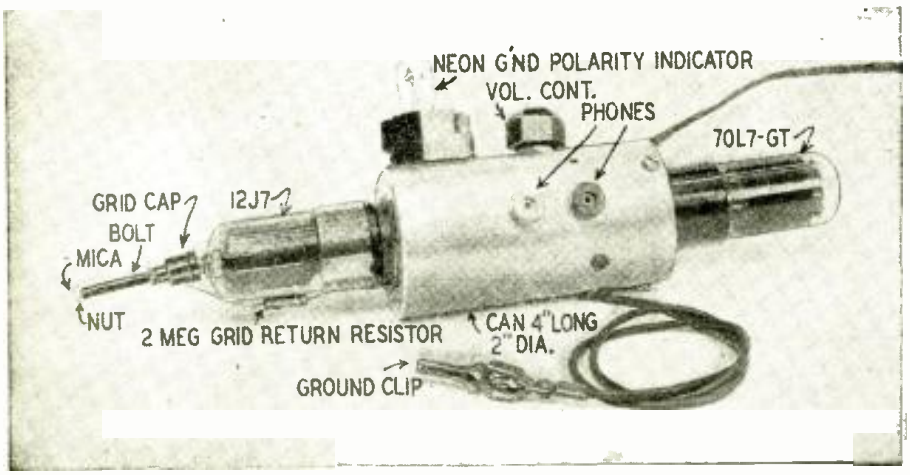


Fig. 1—Hookup of the 2-tube signal tracer.

Normal tracing procedure is to start at the antenna post of the receiver and work from grid to plate, stage-by-stage, until the defective stage is located. With the probe on the antenna post, one or more stations will probably be heard. At the grid of the first stage, it should be possible to select a station with the tuning condenser. If not, look for open or shorted elements in the circuit. As the probe is moved toward the speaker, reduce the volume control on the tracer to keep the signal constant. The amount of reduction necessary is a measure of stage gain.



Full details of the "one-man submarine" can be seen in this photo.

An Electronic Photometer

Some interesting information on photometry is also included in this article.

By J. G. REED

LORD KELVIN very aptly stated, when it is possible to measure a quantity in any experiment, it can be stated safely that there is an understanding of the problem. Photometry is a science in which the subject matter, light, appears at first sight to be an intangible something as elusive as the will-o'-the-wisp, and not likely to yield easily to quantitative measurement. Simple oil-spot photometers are very useful in *comparing* light intensities, and it was such an elementary device, intended for the measurement of light delivered by a photographic enlarger, that first received the author's attention. A series of developmental experiments culminated in the production of the *phototube photometer* described in this article.

With this knowledge it is an easy matter to make a simple oil-spot photometer that will make photo enlargement a much easier process. While not eliminating experience and empirical judgement, work will be placed on Kelvin's truly scientific basis of measurement of a factor—light—which heretofore has been a photographic ingredient over which writers have skipped in haste.

Take a piece of good drawing paper, and with a pin dipped in light machine oil, place a spot on the paper. This spot looks darker than the surrounding paper if light is on the same side as the observer, and changes to a bright spot if the light source is on the opposite side. This is because the treated paper is relatively transparent at this spot. In the first instance less light is reflected, and in the second, more is transmitted by the spot than by the surrounding paper.

Mount the photometer disc on a simple stand—or for a preliminary test it may be held in the hand—and with a table lamp on either side, move the disc back and forth. With two lamps of equal candlepower, and the disc not central between them, if the observer is on the side nearest lamp A of Fig. 1 the oil spot will appear as a dark patch. If the observer's position is changed to the side of lamp B the oil spot will be bright. Now if the photometer disc is slowly moved away from lamp A there will be a critical point where the transmission of light through the oil spot from one lamp exactly balances that lost by nonreflection of the other light, and the oil spot will disappear. Depending on the relative opaqueness and transparency of the untreated and treated paper, the point of balance will be quite marked.

If a small housing is put at the rear of the photometer disc, and lamp B is replaced by a radio dial lamp or clear Christmas-tree light controlled by a variable resistor, we have the basis of a simple photometer.

With precautions taken to counteract the color change of the spot as the brilliancy of the lamp is varied, a very useful darkroom instrument may be constructed. Fig. 2 is a suggested construction for this simple photometer. Dimensions may be varied to suit available material.

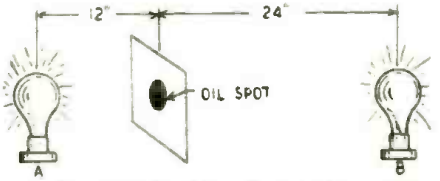


Fig. 1—Principle of the oil-spot photometer.

For long life and constancy of calibration the photometer lamp should be underrun. If a 6-volt, 3 candlepower auto lamp is used, the maximum voltage should not exceed 5 volts, while for the standard 14-volt Christmas-tree light, 10 volts will be found sufficient.

A synthetic oil spot

The detailed drawing, Fig. 3, indicates the composition of the translucent area, which in a permanent instrument must be something more substantial than an oil spot. Use a small square of photographic ground glass for the cover glass. Immediately below have a layer of thin white card or heavy typewriter paper. The paper should have a hole in it about one-eighth of an inch in diameter. Below that place a layer of green-blue tinted cellophane. This tinting is chosen for two reasons; one to neutralize the reddish light from the lamp when at low brilliancy, the other to provide a light to which the eye has maximum sensitivity at low illumination levels. Have the hole through which the light from the lamp passes about five-sixteenths of an inch in diameter so that it will form a halo around the central light spot through the white typewriter paper. This will be found to improve the balance of the spot to reflected and transmitted light conditions.

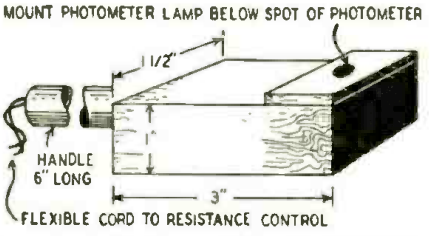


Fig. 2—Practical type of optical photometer.

Cement the combination at the edges and to the body of the photometer with a light brushing of model airplane dope or coil cement, clamping it until dry to make a firm job.

The external control resistor should have a value about twice that of the hot resistance of the lamp, and should be wirewound. For a 6-volt, 3-watt lamp the resistor should have a value of 24 ohms approximately. One of the old-style 30-ohm filament control resistors as used in battery receivers will be found just right.

Having determined experimentally an acceptable exposure time, future exposures are simplified to a procedure in which the diaphragm of the enlarger lens is controlled to give just that light necessary to balance the photometer spot at a standardized setting of the resistor.

An electronic photometer

Being one of those individuals who secure as much, if not more, pleasure in building gadgets for the darkroom as from the photographs produced, the problem of calibrating the simple photometer so that it would read the actual candlepower of the incident light gave the author considerable food for thought. The outcome was the construction of a *phototube photometer* that completely eclipsed the simple oil-spot instrument for which it was originally intended merely as a calibration adjunct. The circuit of this *de luxe* photometer is that given in Fig. 4. It consists of a 3-tube power unit plus a phototube exploring head, the latter being connected to its amplifier through a shielded flexible cable. Light falling on the sensitive surface of the phototube releases electrons which generate a negative potential on the grid of tube V1 of the electronic indicator. Reliance is not placed on any calibration of vacuum tubes as in a tube voltmeter. The two tubes operate merely as couplers to the balancing meter, which in turn serves as a portion of the voltmeter to measure

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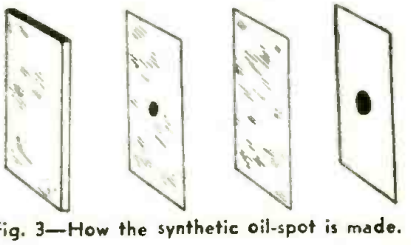
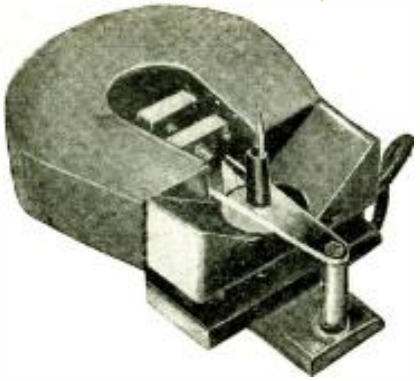


Fig. 3—How the synthetic oil-spot is made.



The skilled sound experimenter can construct this high-quality phonograph pickup himself, and obtain results comparable with those produced by commercial pickups.

Dynamic Pickup

By P. H. RUSSELL

THE moving-coil pickup, if properly designed and made, will take from the record exactly what is there. If the record is first-class and a true reproduction of what the recording microphone received, this pickup will pass to the amplifier a true electrical image of the groove on the record. If the record is poor and distorted, this pickup will show it as such. If the surfaces are bad, scratch will come through. It will reproduce exactly what is given to it. Oscillograph pictures of pure sine waves from constant-frequency records show pure sine wave output from its vibrating parts. It is, as nearly as possible, without character of its own, and is like a perfect mirror, reflecting truly the image presented to it.

The moving-coil pickup has only one very trifling weakness. Its output is low, therefore demanding great amplification.

The characteristics of a moving-coil pickup are essentially neutral—it has no characteristics. At its best, the frequency response is practically flat from 12,000 cycles down to 25 cycles per second. These frequencies are outside, above, and below any which are usually recorded. Its actual response runs down to 5 cycles per second, and up to 15,000 to 16,000. Thus it can take care of any improvements in recording which may be expected. It needs practically no damping. Indeed, for any but loud passages, damping can be omitted entirely. Hence the needle can float from side to side almost without force. These factors combine to enable this pickup to operate with as little as ¼ ounce weight on the record. It thus follows that record wear is negligible, and surface noise is reduced to a minimum. Finally it is stable, not readily damaged, uncritical in the positioning of its parts, and the very light damping can be permanent, never needing attention or replacement. Needle changing becomes an occasional necessity instead of being a constant nuisance. Best of all, this pickup brings out all that there is in any record. Since using a moving coil, this writer has repeatedly heard from his records subtle effects and combinations of instruments that never before were heard when the finest crystal available was in use. The sounds heard can almost be described as stereophonic, as there is such a strong feeling of separation of the instruments. True stereophony cannot be had from a single speaker, but this is the nearest approach that can be conceived.

A pickup of this type can be made by any careful worker on any ordinarily equipped bench. The design shown gives results which not only are far ahead of any crystal so far heard, but compare well with the very fine (and expensive) pickup of commercial make which the writer has had in use for several months. It was made on the writer's home bench with only the ordinary very simple tools that every home mechanic possesses except for two. These two can be dispensed with, although they are so cheap and useful that they warrant spending the dollar or two they cost. They are, first, a set of 5 or 6 very small and fine needle files such as are used by jewelers, and second, a jeweler's saw frame and a dozen or so of the fine wire saws to go with it. The cost of all will be $1.50 to $2.00 only, and these small files and saws have innumerable uses. (N.B. Mount the saws in the frame "pull-cut." Used "push-cut" they break as fast as you can put them in the frame.)

A warning here. Small tools cutting small work cut very fast. Work gently, and the material will yield. Work heavily, and it will tear and spoil.

Heart of the pickup

Figs. 1-a to 1-h show the first part to be made, the moving-coil former and chuck assembly. In working this out, care and the best accuracy you can apply are necessary. It is the heart of the instrument, and the more accurately and better balanced you make it, the better will be the results. It is made up of three pieces of metal lightly soldered together.

The completed pickup. Tone-arm weight is compensated by a spring.

First take a piece of brass (shim stock from a garage is excellent) about 0.004-inch or 0.005-inch thick. ½-inch wide, and ⅞-inch or so long. Soften this with heat to remove all tendency to spring. Roll the brass around a piece of ¼-inch round rod, cutting the meeting ends and filing them with the flat needle file until they butt exactly. Work the

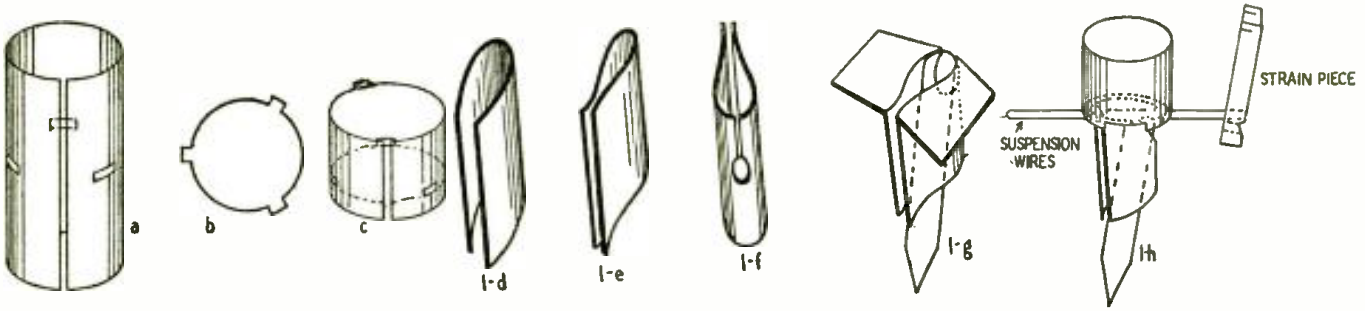


Fig. 1—Components of the moving element, shown in breakdown. Complete references and details of construction are given in the text.

brass into place with the pliers until a perfect cylinder results which fits the ¼-inch rod exactly. Tie a piece of thread around the cylinder temporarily, and at equal intervals, at equal distances of 5/32-inch from one end, cut three very small slots through the wall as shown on Fig 1-a. Use the rod on which you rolled the cylinder as a mandrel for this and for future operations when it is useful.

Take another piece of the same brass, softened as before, and well flattened. Cut a disc ¼-inch in diameter to fit inside the cylinder you have made, and leave three lugs or ears, as shown in Fig. 1-b, which will correspond with the three slots you cut in the cylinder wall. With care partly unroll the cylinder, insert the disc, and roll and tie the cylinder again. Insert the mandrel and trim it up if necessary. Bend up the lugs until they are flush, remove the mandrel, and solder as lightly as you can, seeing that the solder runs well into the joints. Acid flux is best for this work, and a good hot soldering tip helps, too. Now replace the mandrel, setting it snugly against the diaphragm, or disc, you have just fixed, and trim the cylinder down to 5/32-inch on one side of the diaphragm and 1/32-inch on the other. This is shown in Fig. 1-c. The coil form is now complete, but the needle chuck must be attached to it. For this pickup we do not use the old-fashioned screw fastening. It is too heavy. The screw alone, apart from its mount, would weigh as much or more than the combined coil, chuck, and suspension. Instead we use a self-locking chuck. That shown was devised by the writer especially for this pickup.

Needle and chuck

Thin needles are lighter, and so are better. The movement is so light and free that heavy needles are a great disadvantage. The best alloy or chromium needles should be used. You may like a sapphire point better. It is very durable and, with the light weight of this pickup, very easy on records. So, if you decide on a sapphire, acquire one of the standard commercial needles with a round shank and get a jeweler or watchmaker to straighten it and cut it down on his lathe to 37/1000-inch thick and ⅜-inch to 7/16-inch long from point to the opposite end of the shank. A shoulder can be left near the point so that the setting of the jewel will not be disturbed.

Any needle will do, however, so long as you decide on the thickness of the needle you will use. This must be settled, since the very lightweight chuck to be used is built to take only needles of a specified thickness.

Having settled this point, take a piece of brass, or better, phosphor bronze (from an old switch for example, which is what the writer used) about 0.012-inch or 0.014-inch thick, ½-inch long, and wide enough to fold around the thickness of needle you have chosen, with a bit over for a "pinch." Select a piece of waste wire of the proper thickness and fold the bronze lengthwise around the wire. Pinch the surplus in the vise. Take the round needle file and make an oval slot in the back, or folded part of the chuck so that one end of the slot is ¼-inch from one end of the chuck. Cut a slit in the back from the slot to the other end of the chuck. These three steps are shown in Figs. 1-d to 1-f. Fold open *exactly evenly* the ends made by cutting the slot and the slit, making the bend at the slit end of the oval slot. Cut off these ends so that they will fit exactly into the 1/32-inch cup side of the former, exactly centering the chuck. Flatten the turned-down ends so that they face the diaphragm neatly. See Fig. 1-g. Cut tiny slits in the sides of the shallow end of the form below the diaphragm. Trim off the pinched metal. Then set the chuck in place in the form and turn down the metal where you cut the shallow side until the ears of the chuck are locked. Again using acid flux, solder the chuck in place. The appearance then will be somewhat as in Fig. 1-h, which shows the needle as it will sit in place and also the suspension wires which will be attached.

In mounting the chuck, arrange the ears with a slight bend so that when mounted the pinch side will make an angle of 5 to 7 degrees with the coil form. This simple chuck holds a needle thus: The needle enters the chuck freely. When the weight of the pickup rests on it, the needle rests on the diaphragm of the coil form, in the oval slot which was left, and against the pinch of the chuck, being lightly jammed so that no lateral movement of the needle in the chuck is possible. The drag of the record as it turns locks the needle firmly. Yet it can be slipped out easily for changing. The force of the magnet in the pickup keeps the needle from falling out of the chuck when it is removed from the record, and when it is replaced the weight of the pickup at once sets the needle back in the right place, which is the only place at which it can rest. Figs. 1-g and 1-h show the chuck with the needle as it will rest when in use. Fig. 1-h also shows the strain and suspension wires which will be discussed later.

Pole pieces and mountings

The next step will be to make the pole pieces. No dimensions can be given for these as they will depend on the magnet which will be used. Even the shape shown in Fig. 2 may be varied, but that indicated suits very well the usual horseshoe magnet. The essential thing is a circular gap between the pole pieces in which the coil will be set finally, with room to vibrate in the magnetic field between the poles. The gap should be small enough so that as great a concen-

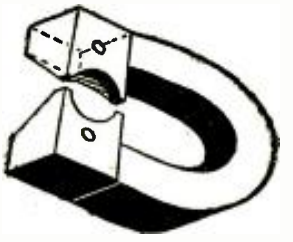
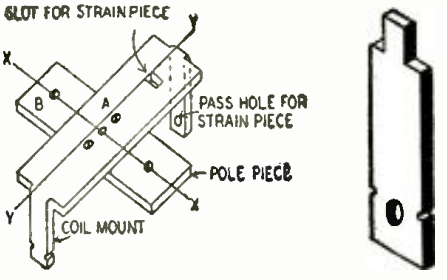


Fig. 2—Pole pieces should resemble these.

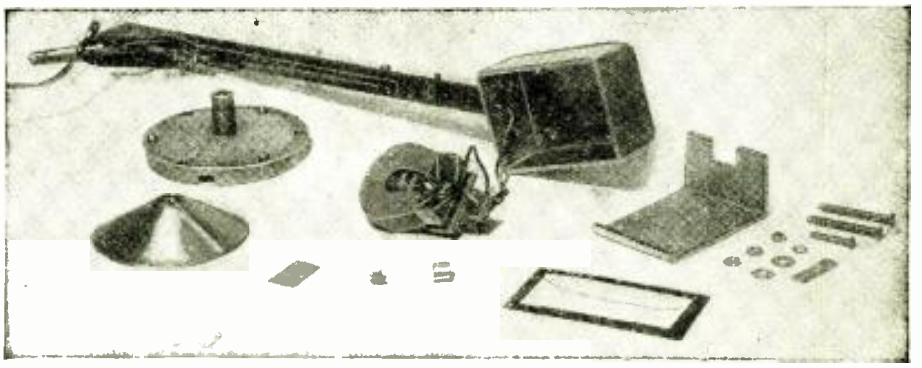
tration of magnetic lines of force as possible can be directed into the coil, and no larger than enough that the coil will not touch the pole pieces in vibrating. For our ¼-inch form, ⅜-inch would be a trifle large. 5/16-inch does very well. Fig. 2 shows how they are made in the pickup designed here, but other patterns may be worked out. The faces of the pole pieces which meet the magnet should be ground flat on the whetstone so that close all-over contact is made. Neglect of this will result in loss of field strength. The inner faces of the pole pieces should be chamfered as shown so as to concentrate the lines of force in the gap.



Figs. 3 and 4—The mounting and strain piece.

The pole pieces now should be mounted. For this, cut a piece of 1/16-inch brass into appropriate size and shape to hold the pole pieces. Drill the mounting strip and drill and tap the pole pieces to correspond. Set in 1/16-inch or 3/32-inch spaces between the pole pieces and the mount. Figs. 3 and 5 will indicate the idea here.

Next will come the mount for the



Exploded view of pickup. A piece of the metal ribbon is shown on bordered paper, foreground.

coil (A in Fig. 3). This will be another piece of 1/16-inch brass fastened centrally. It should be cut about 2¼-inch larger than the width of the pole pieces, leaving enough material for final trimming. Its exact length is not critical. It should be drilled with two holes and the pole piece mount (B in Fig. 3)

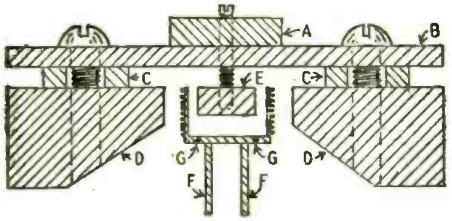


Fig. 5—Cross-section of completed assembly at line X-X. A—coil mount; B—pole piece mount; C-C—magnet spacers; D—pole pieces; E—magnetic core screw assembly; F—needle chuck and coil; G-G—Coil suspension wires.

drilled and tapped to correspond. Round-head screws should be used, and the holes in the coil mount should be about 1/16-inch large to allow for adjustment in case the coil, when finally mounted, is slightly off center. (Work in the home or small shop cannot be as accurate as that in a well-tooled machine shop.) Cut ⅜-inch from the center, toward one end, and bend the mount down at right angles. Cut the width of this turned down part down to ⅛-inch with the file or jeweler's saw. This turn down should be about ½-inch long. About ¾-inch from the center turn down the other end and narrow only enough to permit it to clear the inner walls of the horseshoe magnet. This turn down should also be ½-inch long. In the top, and about 3/16-inch or so from the end just turned down, drill and file a rectangular slot ⅛-inch or so long, across the width of the coil mount, and a trifle wider than the thickness of the brass you are using.

Cut another piece of brass 3/16- or ¼-inch wide, with a square-cut lug or projection on one end to fit loosely into the slot cut in the coil mount. This is shown in Fig. 4, the strain piece. Near the bottom of this strain piece drill and tap a hole for a small screw, and drill and file in a vertical direction a hole in the end of the coil mount on which the last work was done, as shown in Figs. 3 and 6, which show these parts clearly.

This work completes the mounts, except that for the coil itself. For this a loop of fine but strong wire is used. After a number of experiments, final settlement was made on a piece of a nonmagnetic hairspring. A bent or broken hairspring should be obtainable from any watchmaker or repairer. Equally well would do the bronze wire used for the pin-springs in Yale locks. The choice of material does not matter so long as it is thin, nonmagnetic, and strong enough to take some tension without stretch.

Assembling the pickup

Mounting the coil needs care. First cut four tiny slits in the walls of the lower, shallow end of the form, just up to the level of the diaphragm. These should be as nearly the size of the wire you are using as possible, and must be so cut that when the wire is laid in the slots it makes two parallel lines which are equidistant from, and parallel with, that diameter of the form on which lies the plane of the slope of the chuck. Figs. 1-h, 5 and 6 show how these wires are to be set.

Make a bend in the wire to fit around the end of the coil mount opposite the strain piece and bend the extreme tips of the ends of the wire so that they will fit around it when it is set about ⅛-inch from the coil mount. Solder securely to the strain piece and cut off any waste ends of wire. Slip the wire over the mount and tighten the strain piece gently, letting the wire find its place on the mount, parallel with its top side. Assemble with the pole piece mount with pole pieces in place. Slip the form into place, centering it in the gap in the pole pieces. Set the wire in the slots in the form. Slip some pieces of card between the form and the pole piece mount, just enough to hold the form against the wires. Then, see that all is secure and centered and solder the wires lightly in place in the slots. By this method of assembly in place, the desirable exact placing and centering of the coil can easily be arranged.

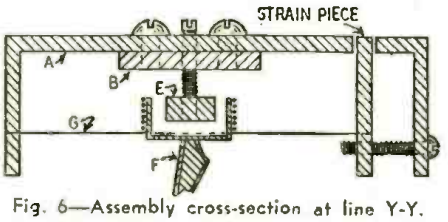


Fig. 6—Assembly cross-section at line Y-Y.

Disassemble the parts and with a sharp knife and fine file clean up the solder, removing all surplus to leave the form as light as possible. Gum a slip of cigarette paper around the form and wind on evenly 20 to 25 turns of the finest enamelled copper or aluminum wire you can obtain. The number of turns is not critical. Paint the wire with collodion (10 cents worth from the drugstore) or with a varnish made from a spot of Duco household cement and ether or acetone. Solder one end of the wire to the form itself, or to any part of the mount. This will be the low-voltage end of the coil. Leave the other end open for the moment.

From an iron rod or bolt 3/16-inch thick cut a cylinder ⅛-inch long. Drill and tap in the exact center for a small iron screw. Do not run the threads clear through. Take the corresponding screw and cut off its head. Drill and tap the pole piece mount and coil mount so that this hole is exactly centered over the middle of the coil form. Attach the iron cylinder to the screw tightly (that is why the threads were not cut all through) and set it in place in the mount so that the cylinder will just fit inside the cup of the form with clearance all around and below. Fasten with a lock-nut above. This part is E in Figs. 5 and 6.

Attach to the mounts at any convenient place a slip of insulating material to which the high-voltage end of the coil can be attached.

Now everything should be ready for final assembly. To avoid spurious vibrations on loud passages, slight damping is necessary. For this purpose a very light grease is good. This can be prepared by mixing vaseline with machine oil so that the mixture will nearly but not quite—run. It is almost oil rather than grease when the texture is right—a sort of oily jelly. Fill the cup of the form with the mixture and spread it around the outside of the coil—enough to fill the space between the coil and the pole pieces. This damping will not deteriorate and will never need attention or replacement; but be sure that the mixture is not in the least stiff or the coil will not respond.

Attach the high voltage end of the coil, set the magnet against the pole pieces, and everything is ready for setting in a head case and tone arm. In assembling, tighten the coil mounting wires so that they are just taut and no more. Overstrain is needless; it may break the delicate wires and in any case puts unnecessary drag on the needle and record. When taut, put a speck of rosin on the screw at the strain piece and melt into the threads with a match held at the head end of the screw. The heat will run, melting the rosin, and lightly, but sufficiently, locking the strain piece. The case and arm can be worked out in any way that appeals to the maker, remembering that the pivoting in both directions must be very free. The final weight of ½-ounce or so on the record is so little that any binding will cause the needle to skip grooves.

In working out the tone arm, provide for a spring or weight counterbalance. The former is theoretically better, though not so easy to arrange. Fig. 7 shows one system of spring counterbalancing which works well.

The pickup head should be offset from the tone arm. Perfect tracking in a radial tone arm is not possible, but an offset angle of 23 degrees with a 10-inch

(Continued on page 62)

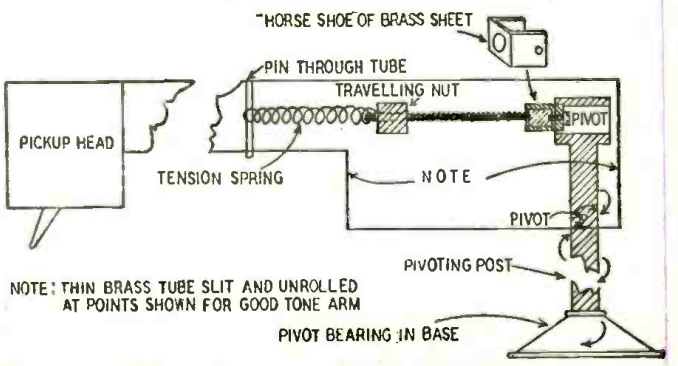
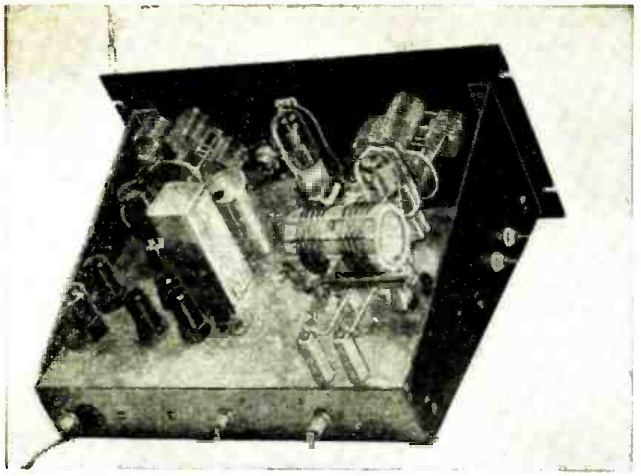
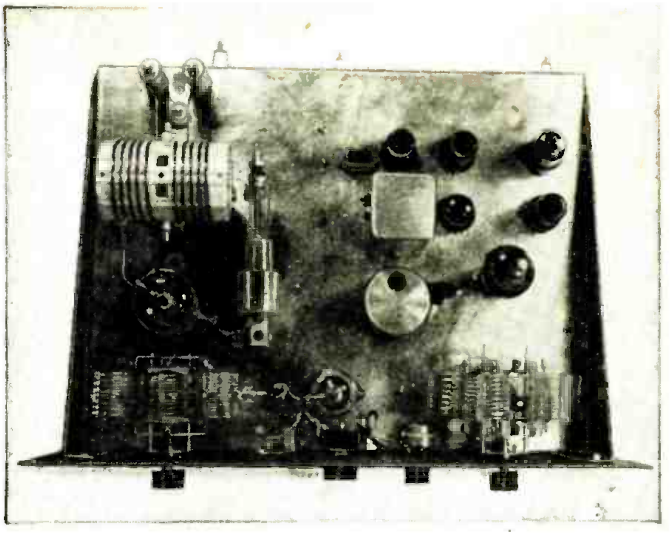


Fig. 7—A spring can be used as shown to counterbalance the head.





Rear and top views of modulator and oscillator sections. Tubes of the FM section shown in Fig. 1, are those nearest rear of chassis.

250-Watt FM-AM Transmitter

PART I--The FM Modulator and Stabilizer

By HARRY D. HOOTON, W3KPX

A 250-WATT amateur-band transmitter recently developed by the author is strictly postwar in its design and application. This transmitter has a medium-power, band-switching r.f. exciter and frequency multiplier stage using a Taylor TB-35 beam-power tetrode tube, and an HK-54 final r.f. amplifier. Either crystal or v.f.o. control may be used as desired, and there are provisions for either amplitude or frequency modulation.

The electrical circuit consists of a 6F6-G crystal-controlled or variable-frequency oscillator, a TB-35 buffer and frequency multiplier, and an HK-54 final r.f. amplifier. These will be discussed in Part II of this article. Shown in the schematic (Fig. 1) is a 6SN7-GT FM speech amplifier, a 6SJ7 reactance tube, a 6H6 discriminator rectifier, a 6SA7 mixer, and a 6C5 crystal oscillator. These tubes are a part of the FM circuit.

The 6SN7-GT, 6SJ7, 6H6, 6SA7, and 6C5 tubes and their associated circuits are used for FM only. The 6SN7-GT is a dual-triode and is used as a 2-stage resistance-capacitance-coupled a.f. amplifier. The 250,000-ohm potentiometer in the grid circuit of the 6SN7-GT second section should be referred to as a *deviation* control rather than a *gain* control. The adjustment of this potentiometer determines the FM band width transmitted. Its indicating scale should be calibrated in terms of the deviation in kilocycles from the main carrier or unmodulated resting frequency.

The 6SJ7 is a reactance-tube modulator. In the circuit arrangement shown in Fig. 1, the 6SJ7 plate-cathode circuit is connected across the 6F6-G FM oscillator grid tank circuit. The plate of the 6SJ7 is connected, through a 0.005-μf coupling condenser, to the hot side of the 6F6-G grid coil. Another 0.005-μf coupling condenser transfers r.f. from the hot side of the oscillator tank coil through a 50,000-ohm resistor to the control grid of the 6SJ7. A current then flows through the 50,000-ohm resistor and the capacitor formed by the input capacitance and stray capacitance of the 6SJ7 grid circuit, producing a voltage drop across the capacitor. The value of the resistor is such that the r.f. volt-

(Continued on page 68)

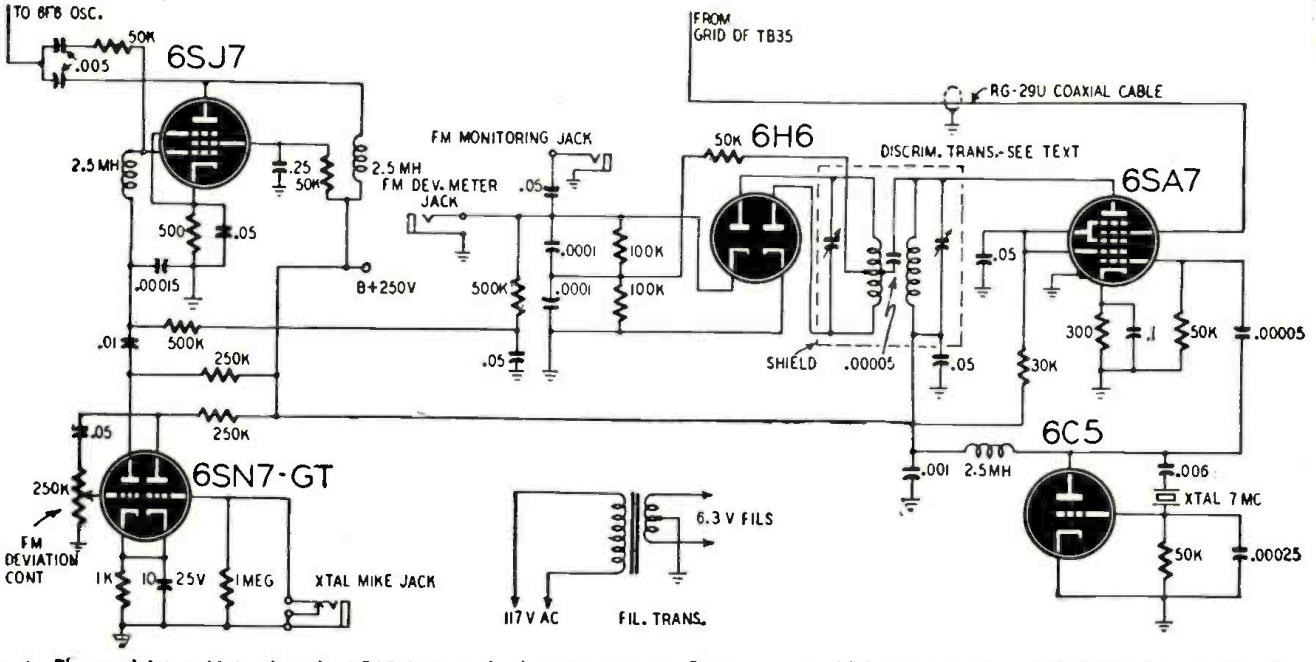
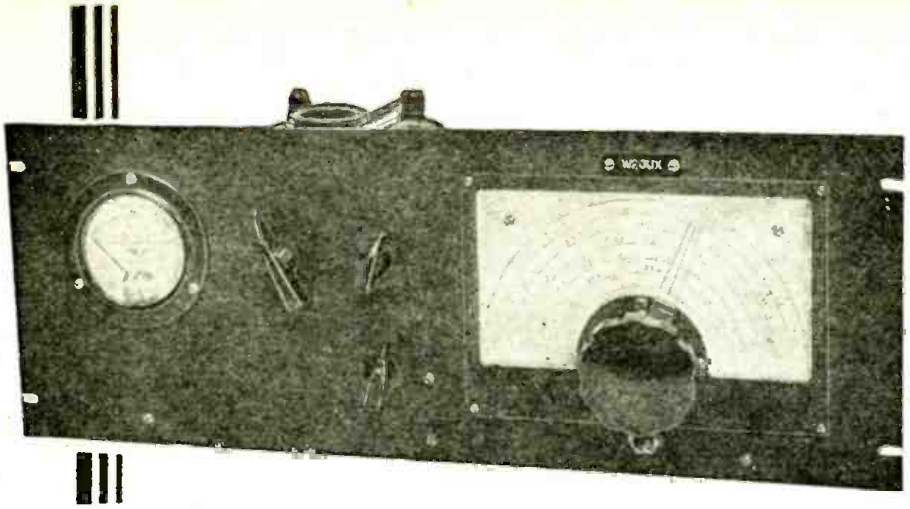


Fig. 1—The modulator. Note that the 6SA7 is wrongly drawn as a 6A7. Suppressor should be to cathode and TB-35 lead to pin No. 8.



V.F.O. EXCITER OR TRANSMITTER

Part II—Constructing the transmitter

By I. QUEEN, W2OUX

A COMPACT and easy to operate rig, this unit is highly effective as a complete transmitter or as an exciter for a multiband unit. The heart of the transmitter is the variable frequency oscillator exciter and many interesting features have been incorporated. Among them are:

V.f.o. circuit of high stability.

Fundamental (80-meter) calibration covering approximately 19 linear inches.

Good spread of other bands.

Choice of oscillator or final keying.

More than 50 watts output with only 400 plate volts.

Panel meter calibrated in watts input to final.

Only two major controls.

Network matching any length antenna.

Output continuously variable over wide range.

Use as complete v.f.o. transmitter for 80 and 40 or as exciter for other bands.

Break-in operation not affecting oscillation.

Alternative crystal control.

Many amateurs already have a multiband transmitter and require only a stable v.f.o. exciter unit. This portion of the transmitter is at left in Fig. 1. It features two stages, a 6SJ7 electron-coupled oscillator and a 6F6 buffer. Output is more than sufficient to fully excite a 6L6, 807, or similar type amplifier, doubler, or final stage. The signal quality is equal to that of a crystal and drift is negligible after a minute or two. The high-C oscillator circuit uses a minimum of 850 μμf capacitance (C2) made up of a 500-μμf and a 350-μμf condenser. These, together with the tuning coil, are enclosed in a shield can in one corner of the oscillator compartment (see photograph). The oscillator operates in the 160-meter band, for greater stability.

The coil is of the permeance - tuned type, 2 inches long and ⅜ inch in diameter. It was rewound with No. 28 wire to reach the 160-meter band with the iron core about halfway out. This required 40 turns, and the cathode tap is made 15 turns from the low end. After the exciter is completed and ready for calibration, the iron-core screw is adjusted to just reach 4 mc with the tuning condenser at minimum. This core adjustment is convenient for minor changes such as recalibration should such ever be necessary. Once set, it is locked by a nut tightened on it. The oscillator is tuned with a 140-μμf midget variable capacitor, anchored to the chassis and controlled by a National ACN dial.

In designing the electron-coupled oscillator, it was decided to concentrate all efforts on one tuned circuit for all bands which might be used rather than fall back on plug-in coils, switching arrangements or multiple-tuned circuits. On a percentage basis, the 80-meter is the widest. For example, of the entire 3.5- to 4.0-mc range, only the portion 3.5 to 3.65 doubles into the 40-meter band. Therefore if the 80-meter scale fills the dial, the 40 calibration is very crowded and occupies less than one-third of the scale! The situation is no better on other harmonic bands. The problem may be solved by using two sweeps of the dial to cover the 80-meter band.

As shown in the photograph, the outer scale covers from 3.5 to 3.73 kc, which multiplies conveniently. The second scale covers from 3.71 to 4.0 mc. To do this a fixed capacitance C3 is switched into or out of the oscillator circuit by the toggle switch under the dial, depending upon whether the lower or upper half of the band is to be worked. A generous overlap is provided. This arrangement works out "fb." The excellent band spread makes it easy to zero-beat a signal and to estimate to about a kilocycle. With the smooth ACN vernier dial, the exciter becomes an accurate "signal generator" for amateur bands.

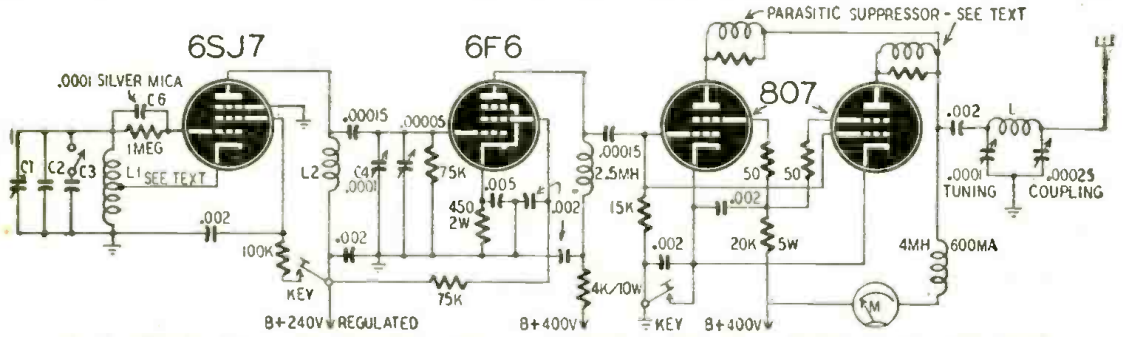


Fig. 1—Circuit of the exciter-transmitter. The e.c.o. is tuned to 160 meters, doubling in the plate circuit.

Adjustment of the high-frequency half of the band (from 4 mc) has been

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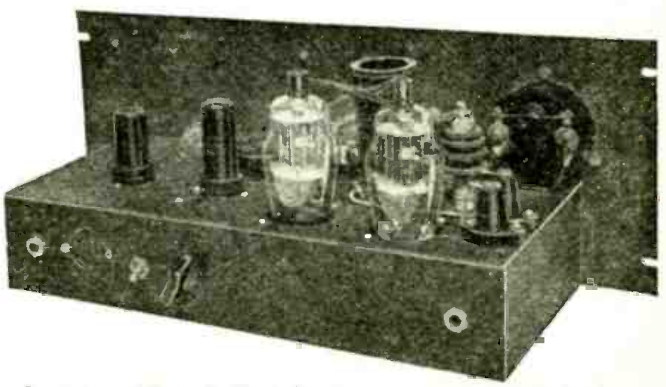

Rear view of the unit. Note shunt across meter and jacks for keys.

THE CRYSTAL RADIO MAKES A COME-BACK

By HARRY WINFIELD

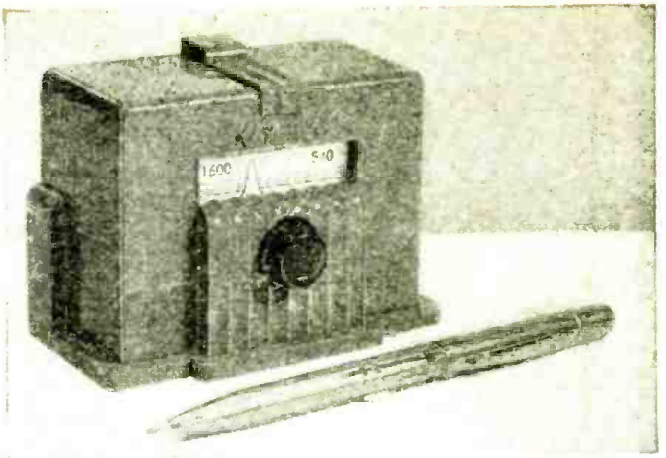

Photo B—The Revell radio is a highly-finished miniature receiver.

NOW that the fussy adjustment required with the old crystal detector has been done away with by the fixed crystal cartridge, the crystal receiver has had a rebirth. No batteries are required, and where a fairly long aerial can be used, the crystal receiver is quite satisfactory for picking up broadcast programs on a pair of headphones.

An aerial at least 75 to 100 feet long (longer if possible) and a good ground connection to a water pipe or other grounded metal system are essential. In some cases programs can be heard by simply connecting the aerial and ground posts of the receiver to a steam radiator and a metal bed or some other metallic mass. This is especially true near powerful broadcast stations. Of course the signals are much louder if an aerial is used, and increase directly with its length and height.

The crystal set does not amplify; therefore every bit of the original signal must be preserved and transferred to the crystal detector. One of the secrets of success with crystal sets is to use a pair of *highly sensitive* headphones.

The aerial wire can be of any convenient size, from No. 12 to No. 18 gauge copper. It should be as long and high as possible, and well insulated at each end. A length of insulated wire (such as bell wire) may be run along the baseboard or over the beams of an attic, to make an indoor aerial which sometimes gives fair results. A ground clamp should be used to make *firm* connection to a water or steam pipe. In the country, when camping for instance, the radio may be grounded by connecting the ground wire to a piece of metal thrown into a stream. If a wire fence is handy, it may be tried as an aerial. Another way to pick up signals is to connect the aerial post of the set to one side of a condenser, the opposite terminal of the condenser being joined to one side of a telephone circuit.

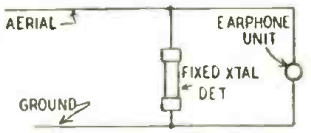

Fig. 1—Mystery radio's ultra-simple circuit.

Some of the new crystal receivers are illustrated on this page.

The mystery set

The simplest radio receiver imaginable combines a headphone with a crystal detector bridged across it. Such a combination is used in the "mystery" receiver illustrated in Photo A. The diagram of such a receiver (incidentally one of the first crystal sets ever used by the early experimenters) is given in Fig. 1.

The headphone in this receiver is a hearing-aid type and is so small that it fits into the ear. The fixed crystal detector fits inside the housing; all that remains are the 2 insulated wires with spring clips at their terminals. One clip may be connected to a water pipe and the other to an aerial wire or substitute metallic mass.

This set is useful as an emergency receiver or for other purposes, but has no tuner for selecting desired stations. The station intercepted will depend upon the natural frequency of the antenna system and the relative strength of local broadcast stations. If the aerial wire is long, other stations may be tuned in by trying different size condensers in series with it. A variable tuning coil (one fitted with a slider) could be connected in series with the aerial and would provide some degree of tuning. A variable condenser in series with the aerial may also be tried.

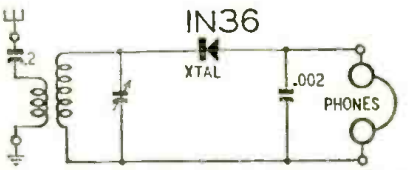

Fig. 2—The Revell radio circuit is standard.

Revell radar radio

A more ambitious crystal set is illustrated in Photo B. The diagram for this *tunable* receiver appears in Fig. 2. The drawing, Fig. 3, is a sectional view of the ingenious variable condenser used

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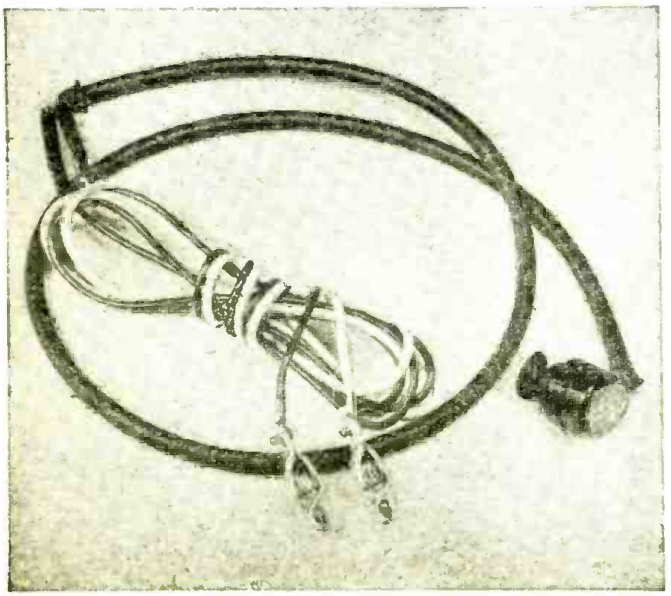
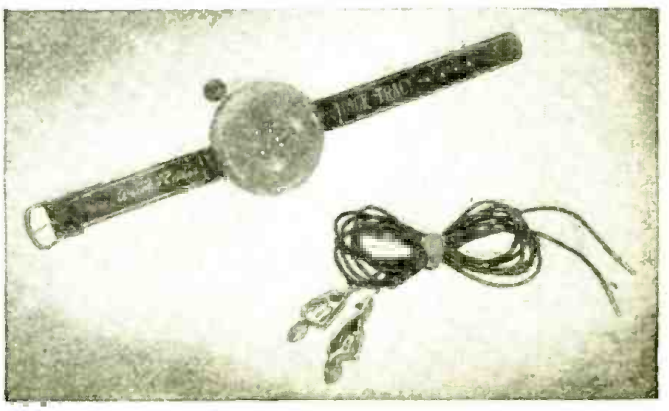
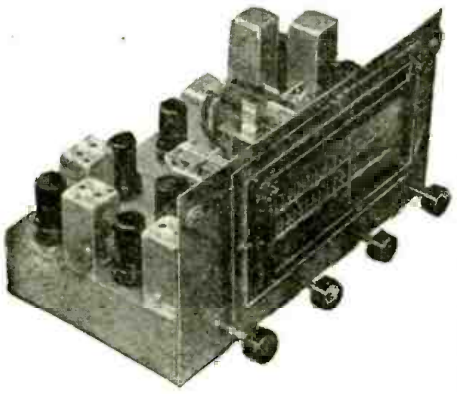


Photo A, left—Taybern Equipment's Mystery radio plugs in the ear.
Photo C, below—Atomic radio, made by Da-Myco, is tuned by the stem.



RADIO SERVICE SET DATA



Browning FM-AM Tuner RJ-12

THE Browning Model RJ-12 FM-AM tuner is one of the first of its kind to appear on the postwar radio market. It is supplied by the manufacturer as a chassis 7⅜ inches high, 13½ inches wide, and 9 inches deep, designed to be fitted into bookcases, shelves, drawers and other built-in installations in the home or office. Its rack-mounting counterpart, the Model RJ-14, has a panel for mounting in a standard relay rack for laboratory or special installations.

The chassis consists of separate superheterodyne FM and AM circuits. The AM section tunes the broadcast band from 540 to 1650 kc and the FM section from 88 to 108 mc. The detection systems are connected alternately to the volume control through a section of the band switch and will give good results with an a.f. amplifier that will work with 0.1 volt input or more.

An audio amplifier, loudspeaker, antenna, and a power supply delivering 250 volts d.c. at 65 ma and 6.3 volts a.c. at 4 amperes are required to complete the installation. All voltage and control leads are brought into a terminal strip at the bottom of the chassis. If the voltages cannot be supplied by the amplifier, a Browning Model PF-12 power supply, Fig. 2, can be used. A high-fidelity amplifier such as the Model RJ-12 amplifier, Fig. 3, is recommended for use with the tuners.

The tuner has a 4 x 8 inch slide-rule dial with separate 6¼-inch calibrated scales for FM and AM bands. Four walnut-colored bakelite control knobs are located symmetrically across the bottom of the chassis.

Although the RJ-12 is essentially two separate receivers, the circuits are arranged so that only four controls are used. The on-off switch, on the left side of the chassis, is connected to the terminal board so that it can make or break the a.c. power line to a power supply and amplifier that may be located some distance from the tuner. The sec-

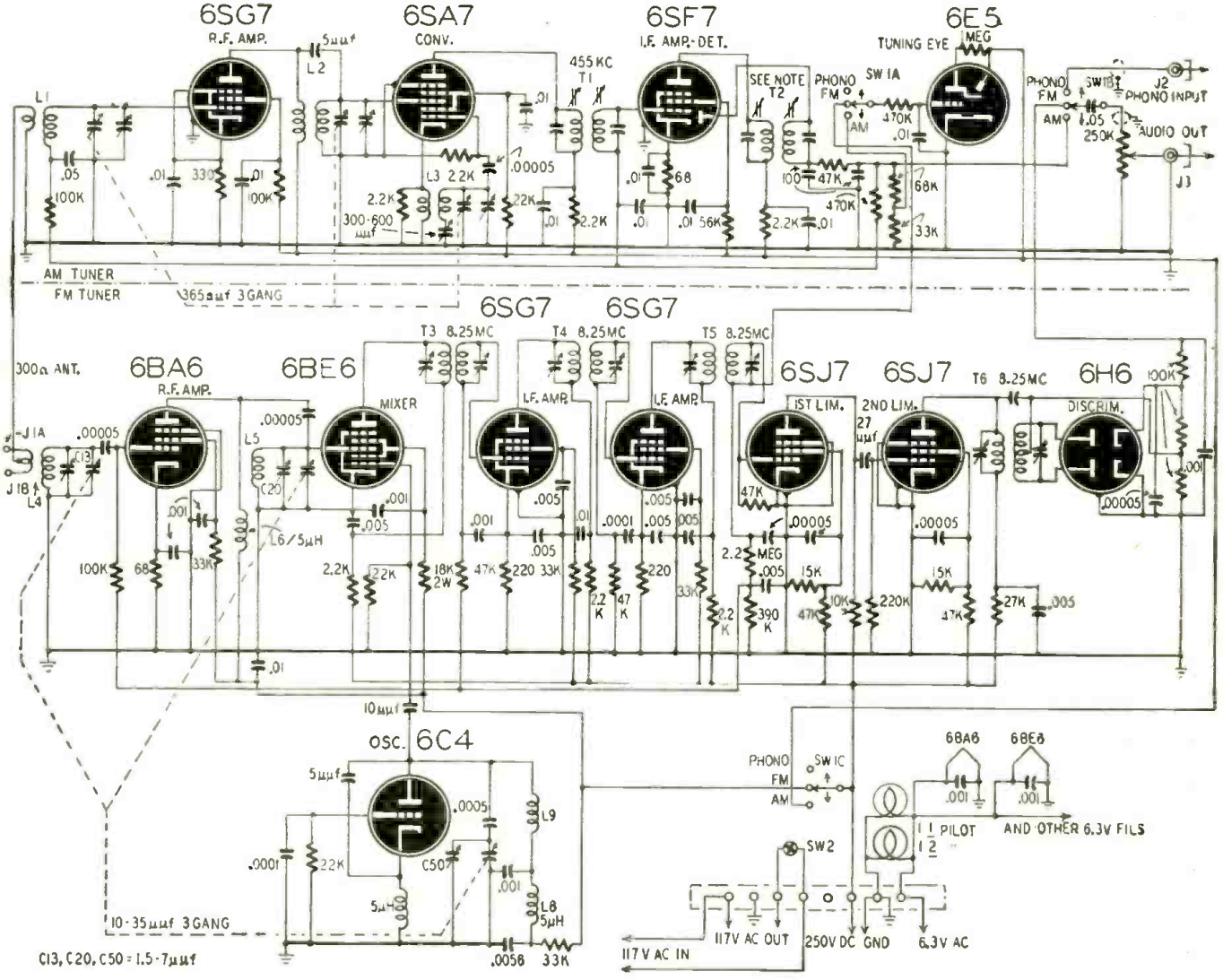


Fig. 1—This tuner is 2 receivers, using separate components throughout. Only the power supply and the volume control are common.

RADIO-CRAFT discontinues with this issue the Radio Data sheets. Printing 12 (or even 144) circuits a year has become meaningless. New models have been turned out literally by the hundreds. Means of securing information about them by servicemen is excellently organized. The space will in the future be devoted to articles about new and *different* receivers and other radio equipment, new servicing methods (FM, etc.) and topical servicing subjects.

ond knob is the tuning control. A slow-tuning drive mechanism, requiring 10½ turns for complete band coverage on either band, is ganged to the tuning condensers of the FM and AM sections so that both are tuned together.

The band switch is the third control from the left. Its setting is indicated by markings on a scale visible through a small aperture in the center of the dial. In the FM position B-plus voltages are removed from the AM section. In the AM position, B-plus voltage is removed from the 6C4 high-frequency oscillator. In the phono position, plate voltages are removed from both tuners. Other sections of this switch connect the 6E5 tuning indicator and volume control to the section in use. The volume control—the right-hand knob—is connected through the switch to a phono input jack on the rear skirt of the chassis so that it can be used to control the volume of any phono attachment that may be used with the amplifier.

The circuit and tubes

The AM section consists of a 3-tube superheterodyne circuit using a 6SG7 r.f. amplifier, 6SA7 mixer-oscillator and a 6SF7 455-kc i.f. amplifier, diode detector, and a.v.c. voltage source for the i.f. and r.f. grids. The circuit is highly sensitive—the manufacturer claims a sensitivity of 1 microvolt or better over the entire band. The selectivity leaves little to be desired, even on the crowded high-frequency end of the band. The response curve seems to be sufficiently wide to pass the side bands of the average broadcast station and still provide adequate station separation.

The FM circuit is designed to work from a suitable antenna coupled to the input of the tuner through a 300-ohm line. The entire antenna system, dipoles and transmission line, is the antenna for the AM section. The front end of the FM tuner uses three of the new miniature tubes. They are: 6BA6 r.f. amplifier, 6BE6 mixer, and 6C4 separate oscillator tuning 8.25 mc higher than the signal frequency. The 6BE6 is followed by a 2-stage i.f. amplifier using 6SG7's, according to the schematic. The model tested used two 7AG7's in this position. The i.f. transformers are overcoupled to provide a band-pass of approximately 150 kc. They are followed by a 2-stage cascade limiter using 6SJ7's. These circuits are adjusted so that complete limiting takes place with a signal input of less than 15 microvolts. The limiter stages work into a Foster-Seeley discriminator using a 6H6. High-frequency de-emphasis is used on the output of this circuit to compensate for pre-emphasis at the transmitter.

Tested in a Manhattan office building with an indoor antenna, the FM tuner proved to have the expected wide-range audio response typical of FM. Since all FM stations heard were of good intensity it was impossible to make a sensitivity check on weaker signals. On AM, high-frequency audio response was definitely lower than on FM, as expected, because of normal side-band cutting. The AM sensitivity seems to be better than average for a set of this type.

When the tuning indicator is used on the AM channel, minimum shadow indicates correct tuning. On the FM band, the indicator beam overlaps and exact adjustment must be made by tuning for minimum background noise.

Servicing angles

The AM intermediate frequency is 456 kc, and alignment is standard. To align the FM section, a frequency-modulated signal generator and an oscilloscope is required for best results. The generator should have a sweep well in excess of 200 kc each side of the 8.25 mc i.f. General principles of visual FM servicing were discussed in the March, 1946, issue of RADIO-CRAFT. Note that this receiver is overcoupled to give a double-humped curve with a slight dip at the center frequency.

An AM signal of exactly 8.25 mc may well be applied to the same grid as the FM signal, to furnish a center marker. The discriminator is aligned in standard fashion.

An AM generator and a high-resistance voltmeter in the grid return of the first limiter is a less satisfactory alignment equipment, but can be used if better apparatus is not available. The method is to adjust to maximum meter reading (working back from the last i.f. transformer) at 8.25 mc. The generator is then varied in steps of about 20 kc each side of the center frequency, adjusting for a symmetrical curve, humped equally on each side of the center frequency.

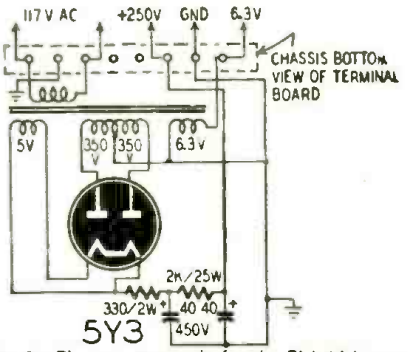
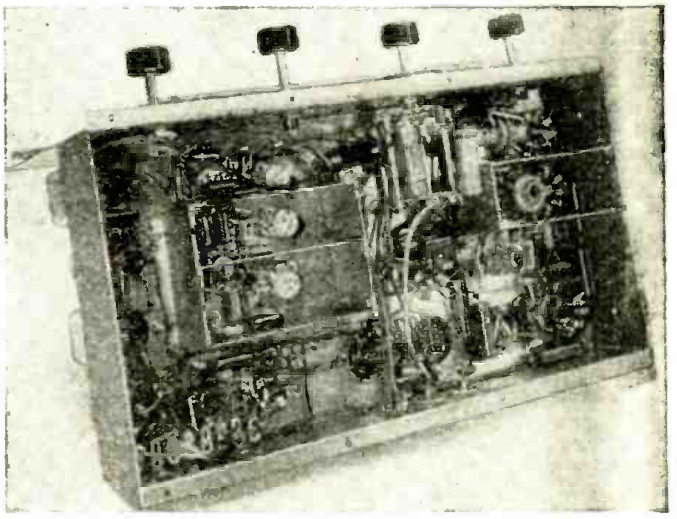


Fig. 2—The power supply for the FM-AM tuner.


Under side of the tuner. The shields in center enclose FM coils.

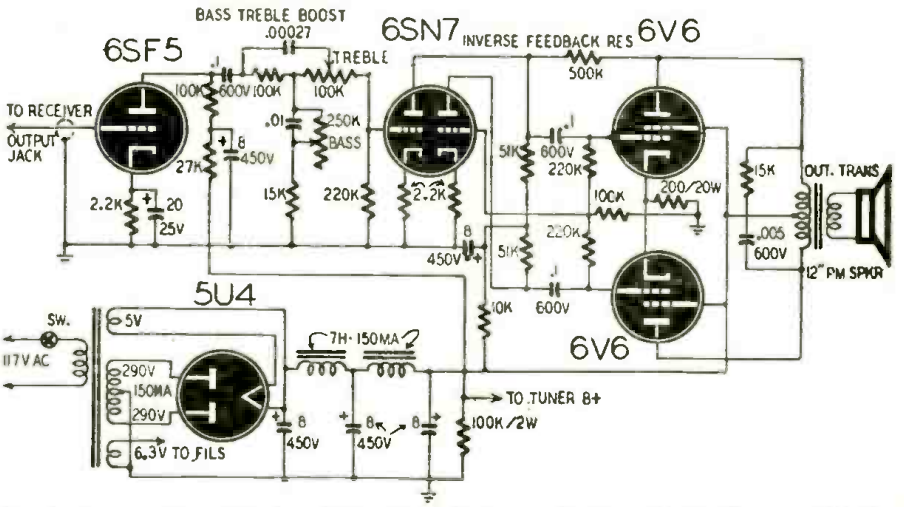
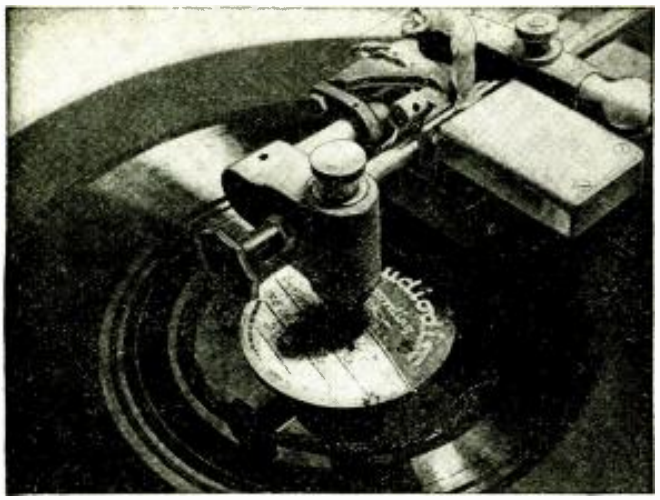
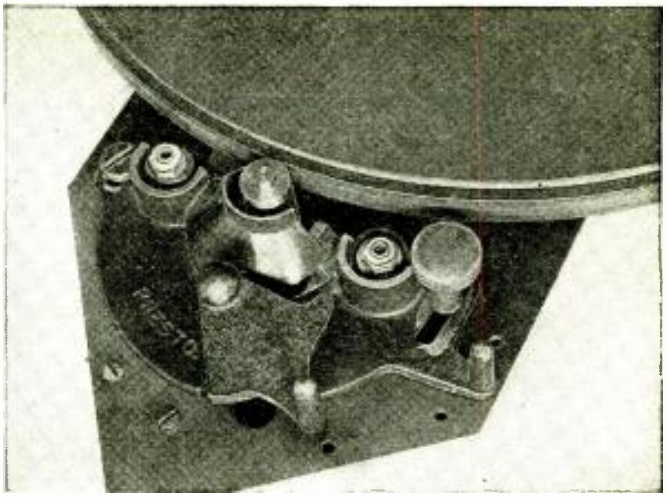


Fig. 3—Any good amplifier may be used with the tuner. This circuit is the Browning RJ-12.



Recorder working well, chip thrown clear and moving toward center.



Drive roller and rim must be kept free from dirt, damage or wear.

A Small Recording Studio

Part V — The commonest recording troubles

By J. C. HOADLEY

THE recorder who understands the fundamentals of his art will recognize most recording troubles as they appear. As a refresher for the skilled technician, and as a help to those who have not reached the professional stage, the following check list is appended. It contains the commonest causes of poor recordings, as well as the simplest methods of correcting them.

1. *Vibration being transferred from the drive motor to record.* This is evidenced by the appearance of spokelike or wavy patterns on the record, which herald rumble or wow or both.

It may be that the motor has excessive vibration because of improper centering of the armature, worn bearings, poor balancing of the armature, or an off-center drive pulley on the motor shaft. If the motor is mounted in rubber, it may be touching some part of the motor board, or the rubber may be hardened by age, oil, or heat. The motor may be adjusted too tightly against the rubber idler wheels so they do not reduce the vibration transmitted to the turntable.

The motor board may not be securely screwed down to the cabinet in which it is mounted. The plate on which the recording mechanism is mounted should be secured to a solid base, so that it cannot vibrate. Any vibration picked up by the motor will appear in the playback as a 60-cycle hum, which may vary in intensity throughout one revolution of the turntable.

Other record patterns can be caused by misalignment of any part of the overhead mechanism so that binding occurs. This can be so serious as to slow down the turntable for a small part of one revolution of the turntable. This will result in a very low frequency rumble. It is too low to be heard as such but will cause a modulation of the higher frequencies, producing a noticeable wow. The ear is particularly sensitive to a change in pitch of a given tone, so even a very small variation in speed of the turntable during one revolution will result in a particularly displeasing effect.

Another noticeable pattern is an apparent difference in the spaces between the grooves, caused by inexpensive, poorly made, or defective lead screw or half nut follower, or both. There should be no play in the cutter carriage when it is lowered onto the lead screw. *It is wise to wiggle the carriage sideways to be sure the half nut is seated in the lead screw, as it is very disconcerting if it seats itself in the middle of a recording and ruins the disc.*

2. *High scratch level when recording is played back.* The most obvious cause is a dull cutting stylus, although it can be caused by a stylus set at the wrong cutting angle, by a poor quality blank, too hard material on the surface of the blank, or too deep cutting. Surface noise results also from a defective playback needle or one which does not match the groove radius. Dust is a common cause of surface noise, and casually wiping a disc builds up a static charge on its surface which makes the dust adhere more tenaciously. Discs should be kept in their envelopes when not being played.

A chipped sapphire playback stylus will ruin a disc with one playing, as the edges of the chipped spot are as sharp as a razor, due to the crystalline structure of a sapphire. A so-called permanent needle should never be removed from and later replaced in a playback pickup which has a needle chuck. It is almost impossible to put the needle back in the same position. If it is rotated only slightly, it will present a sharp cutting edge to the record and ruin it.

3. *Playback pickup jumps grooves or skids across the surface of a recording.* The most prevalent cause of this ill is too shallow grooves. Playback pickups vary considerably in their ability to track a record. This ability is inherent in the design of the pickup and includes the compliance of the moving assembly, diameter of the stylus, point pressure, needle point impedance, design of the arm, and quality and freedom of the arm bearings.

Inexpensive pickups of the low-pressure type have difficulty in tracking high-amplitude low-frequency passages. This characteristic should be taken into account when a disc is made for a customer. It is wise to cut a record a little deeper and refrain from extreme low-frequency boosting when making a recording for someone who is going to play the disc back on a home radio-phonograph combination.

Groove jumping is caused also by too thin walls between grooves, particularly when played back with a lightweight pickup.

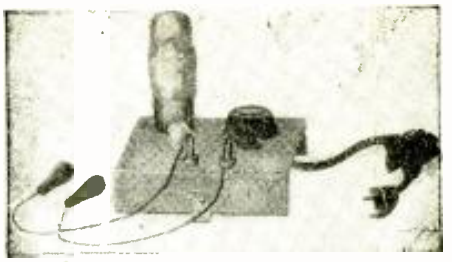
4. *Distortion which was not present in the material recorded.* Distortion may be caused by overloading the cutter head. This is particularly likely to happen when boosting the high or low frequencies. The recording level indicator cannot follow the peaks, which may be several times as high as the average recording level. It is wise to record at a level well under the maximum the head and grooves will stand.

Distortion can be caused by a recording amplifier which is defective or of insufficient power capability to have low distortion on the peaks. It is wise to have at least three times as much power available as the maximum level the head will stand.

Too light a cut will cause distortion because the needle cannot accurately

(Continued on page 57)

Rapid Checker For Capacity-Continuity



By ALFRED SHORTCUT

FOR the past five years a mechanic has been bringing me his condensers and coils for testing. Each time I have put the condensers on my Solar CC-160 and checked for leakage, then capacity. With my Weston 772 I have then determined if the coils were open. Since my friend is a washing-machine mechanic, specifically a Maytag mechanic, all the condensers are the same capacity and all coils the same value. Invariably, when this mechanic finds a motor that will not start he finds the cause is a coil (secondary open) or a leaky condenser.

Over a period of years of this testing, I have learned to make the tests quickly, and since the mechanic is my friend, there has been no charge. However, over this period of years, it has represented wasted time. The other day he asked, "Why can't you make a little gadget I can test these with?" I started thinking, and the unit which is the basis of this article was worked out.

There are hundreds of cases where a simple tester is needed to show leakage and if a condenser or coil is open this tester will serve that purpose.

While capacity of electrolytic condensers decreases as the electrolyte dries out, this is not the case with paper condensers. They usually have their rated capacity or are open. This characteristic makes them easier to test, especially in my specific case. For reasons of economy the instrument had to be a.c.-d.c., and for both economy and simplicity, it had to dispense with meters. With this in mind the circuit of Fig. 1 was dreamed up.

A receiving triode is used as a rectifier since the B current is negligible. Any type will do. A Type 37 was chosen because there were plenty of 5-prong sockets in the junk box. Because of the low current drain a simple R-C filter composed of R2 and C1 is sufficient.

The leakage-checking section is the familiar relaxation oscillator. With S DOWN, filtered d.c. is fed to the terminals. If a condenser is connected to the Cx terminals, any leakage in the condenser will allow a current to flow and charge C2. When C2 has charged to a voltage sufficient to ionize the neon bulb V2, it will flash and discharge C2 and the process will repeat itself. How often this occurs depends on the size of C2, the ionization voltage of V2, and the leakage resistance of the condenser under test. V2 and C2 are chosen so that the light will flash only at long intervals if the leakage of Cx is negligible.

The capacity-analyzing circuit of the tester is the really unique part. Since it is necessary to know only that the condenser has capacity and not necessary to know how much capacity, as explained above, the circuit was made very simple.

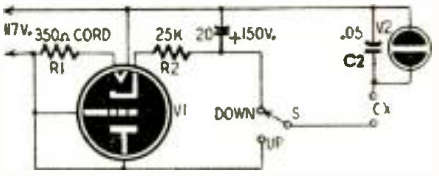


Fig. 1—Circuit diagram of the quick checker.

When S is UP, a.c. is fed to the Cx terminals. Since a good condenser will pass a.c., Cx and C2 act as a voltage divider. If Cx is large enough then the neon bulb lights, indicating Cx has capacity. If Cx is open or very low in capacity, there will not be enough voltage to cause V2 to light.

In using the checker, always check for leakage with S DOWN first. The neon bulb should blink only at very long intervals. If it blinks frequently or glows, *discard the condenser*. If this test is O.K. put S in the UP position. The neon bulb should glow.

The instrument can also be used as a continuity checker.

In my particular case, S is closed and the Cx terminals connected to the magneto coil to test it. Good coils cause the neon bulb to glow brightly. Defective coils cause a faint glow or blinking of the light, while coils entirely open produce no glow.

Construction of the instrument is very simple, and the figure and photograph are self-explanatory. For safety's sake, the circuit is insulated from the chassis, and the test clips insulated with rubber boots.

Parasitic Oscillations

MANY amateur rigs and even commercial transmitters are subject to parasitic oscillations unless proper precautions are taken. A parasitic is due to feedback from circuits or circuit elements of the transmitter. It can result in overheated tubes and components, poor efficiency, raspy signals, and undesired radiation. If the circuit responsible is concentrated (such as a choke coil) the parasitic frequency is much lower than the operating frequency. If it is distributed (as in conducting leads) the parasitic frequency is much higher.

A parasitic is heard as a rough, unstable signal when tuned in on a receiver. Its approximate frequency can be determined by tuning to the lowest frequency at which it appears, but this shaky note often comes and goes rapidly as the result of vibration or hand capacitance. A more definite indicator is a neon bulb held near the plate or grid lead of the suspected stage, with excitation off. It will glow yellow if the frequency is relatively low and will show red or purple on very high frequency. An incandescent bulb with the usual single-turn loop also can be used to detect the r.f. In many cases a milliammeter connected in the grid or plate head will show undesired oscillations.

Eliminating the parasitic is generally a cut-and-try procedure. Certain rules apply in every case. During attempts to reduce the oscillation the bulb or meter should be left connected so that improvement can be seen immediately if it occurs.

If the parasitic is known to be a low-frequency one it is due to the design of the set (type of components used, their position or their arrangement in the circuit).

It is preferable not to use choke coil coupling in both input and output. If it is necessary these chokes should not be alike. Tapped coils as in Fig. 1 also give trouble. By-pass condensers can resonate with chokes or coils to generate the low frequencies. Note in Fig. 2 that RFC1 and RFC2 make a t.p.t.g. circuit when tuned by circuit capacities and their own distributed capacitance. A good system is to short (or remove the tap) from one coil at a time and note the difference. The position of the coil sometimes makes a difference in the parasitic.

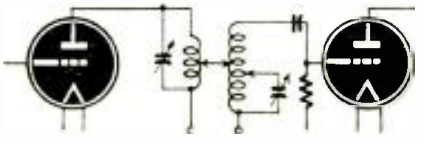


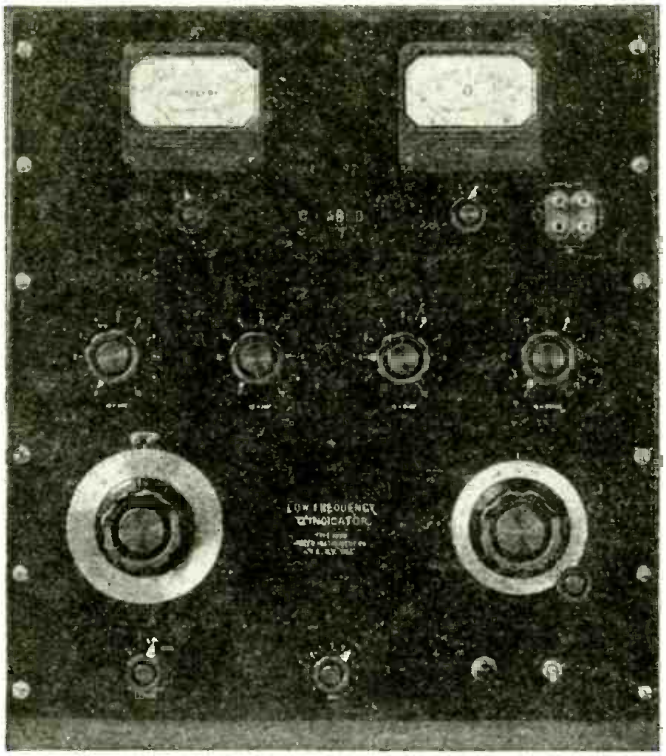
Fig. 1—Tapped coils are a source of trouble.

If there are no choke coils or tapped coils in the circuit the parasitic is likely to be a high-frequency oscillation. This can occur in any type transmitter since it is due to resonating leads. By-pass condensers should be connected directly at tube elements or power supply terminals. Even an inch or two of wire can set up a stubborn parasitic. Grid

(Continued on page 56)

"Q"

Some facts on a much used, but not always well understood term, the "figure of merit" of a coil or other inductive element.



Courtesy Freed Instrument Co.

A commercial type of Q meter used for measuring large inductances.

By RAYMOND G. JOHNSON

THE symbol Q appears often in literature referring to the design of coils and tuned circuits. This quantity Q is defined by the ratio

$$\frac{\text{reactance}}{\text{resistance}}$$

Its importance is evident when it is considered that reactance and resistance determine circuit selectivity and sensitivity. Accurate determination of Q is usually required in laboratories and designers' workshops, but servicemen, technicians, and amateurs are more concerned with relative Q values.

It is usual to assume that the coil is responsible for all tuned-circuit loss because a good condenser has very little resistance. In other words, coil Q can be considered to equal circuit Q. The total resistance of a coil is due to: (1) d.c. resistance; (2) eddy currents and skin effect; (3) dielectric loss. Generally, high-frequency circuits have greater Q because of greater reactance. However, since total resistance also increases with frequency, the change is not as great as might be supposed at first. Actually, over a small frequency range, Q remains practically constant for any given circuit.

Fig. 1 shows a parallel resonant circuit which has been simplified for analysis by drawing the coil resistance R as a separate component. This leaves the coil as a pure inductance L. When tuned to resonance, the coil and condenser reactances are opposite and practically equal, so the r.f. tank current (I) is equal to E/R. The voltage across the resistance is IR and across the coil is IX where X is the reactance found from the formula $2\pi fL$. Since $Q = X/R$, a simple development of the equation shows that $X = QR$, from which it is apparent that *the voltage across the coil is Q times as large as the total circuit voltage.* Note that THE TOTAL IMPRESSED OR INDUCED CIRCUIT VOLTAGE IS MAGNIFIED Q TIMES BY THE RESONANT CIRCUIT.

Because of the higher reactance/resistance ratio, a high-Q circuit tunes more sharply and produces a higher peak voltage at resonance than a circuit with a low Q. Therefore, the tank coils of sharply tuned voltage amplifiers should be designed for high Q. This high Q is gained by reducing the resistance for a given reactance or hy increasing the reactance without proportional increase of resistance.

Low coil resistance is ohtained by using good insulation and dielectric materials such as polystyrene, quartz, or isolantite. At v.h.f., the conductors should be made of metal tubing or ribbon, preferably silver-plated, to minimize skin effect. If shielding is used around a coil, it should not be too close because of eddy-current loss.

For a given size and shape of coil, the reactance may be increased by using an iron core. The core is made of finely divided alloy which greatly increases the effective permeance without proportional increase of resistance due to eddy currents. The net reactance may also be increased by minimizing the distributed capacitance between turns of the coil. This can be done by space-winding, which is especially important at higher frequencies.

To calculate a coil's Q, its reactance and resistance must be known. The first may be determined from charts or tables giving inductance from coil dimensions. Unfortunately, the second represents a very difficult problem because of the many variables. It is much easier to measure the Q of a coil without knowing its resistance. The Q meter is widely used in laboratories and manufacturing plants. Fig. 2 shows its basic schematic. The instrument requires an r.f. generator, variable condenser, and 2 voltmeters. One voltmeter must measure values in the neighborhood of 0.02 volt, and the other is designed for the range of about 5 volts. As shown in the diagram, the low-reading meter is usually an r.f. current meter in series with a known resistance.

To operate the Q meter, the coil to be measured is connected across L and the generator is adjusted to the desired frequency. The r.f. output is increased until the voltage across R is some convenient value, say 0.02 volt. As a practical example, R may be 0.04 ohm and the ammeter adjusted for 0.5 ampere. The condenser is then adjusted for resonance and the voltage is noted on the vacuum-tube voltmeter. If it happens to be 2 volts, the Q of the coil would be $2/0.02 = 100$. This follows

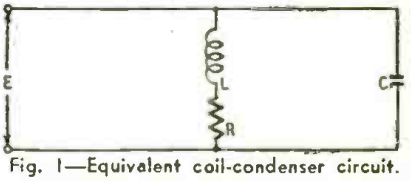


Fig. 1—Equivalent coil-condenser circuit.

from the magnification property of the resonant circuit. Note that R must be low enough not to interfere seriously with the remainder of the circuit but must be large in comparison to the coil resistance.

In replacing coils in radio receivers it is important to substitute one with approximately the same Q. A lower value

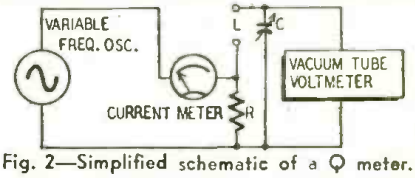


Fig. 2—Simplified schematic of a Q meter.

reduces sensitivity, while a higher value may cause instability or oscillation. For this purpose, a comparative Q meter is sufficient. The meters of Fig. 2 need not be calibrated in exact values. The v.t.v.m. is used merely to indicate relative Q.

ANTENNA PRINCIPLES

PART VIII---Metallic lens and electromagnetic horn antennas

By JORDAN McQUAY

A NEWCOMER to the microwave field of transmission and reception is the *metallic lens system* developed recently by the Bell Telephone Laboratories. Operating at wave lengths of less than 5 centimeters, the system (Photo A) consists of a large, square metal lens. Action is analogous to that of light waves: a glass lens slows down the light passing through it, and the thicker the glass the greater the effect. A double-convex lens slows light rays nearest its optical center more than those nearer the periphery. In this manner, the light rays can be brought into focus at some point beyond the lens.

Similarly, microwaves *gain speed* in passing through wave guides or between resonant metal plates. And a metal lens system—consisting of a geometrical arrangement of such plates—can be used to equalize the wave front of microwaves, thus concentrating the energy in a given direction.

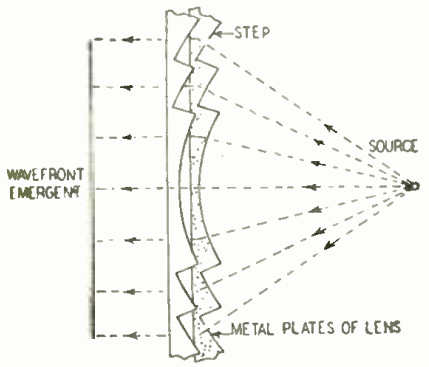


Fig. 1—Cross-section of metal-lens antenna.

When microwaves from a point source (Fig. 1) pass through the wave-guide-like sections of the lens, they are speeded up because of the dimensions and arrangement of the multiplate metal lens. These shaped plates speed up the edges of the beam relative to the center so that the emerging wave front is flattened—since the path between metal plates is longer at the sides than at the center.

Avoidance of long paths between plates through "stepped" construction, besides saving material and weight, results in uniform transmission over a very wide frequency band. This is an important advantage in broad-band microwave communication systems.

The arrays of metal plates have been made in a variety of sizes, from those only a few inches square for millimeter transmission and reception to the largest array for use with 5- and 6-centimeter waves. With the latter array it's possible to obtain a radiation pattern less than one-tenth of a degree in width, an important advance in microwave directivity.

Metallic lens systems shortly will be used in intercity links for the simultaneous transmission of television, telephone, telegraph, facsimile, teletype, and other services—all possible with microwaves only a few centimeters in length.

Electromagnetic horns

Another unidirectional u.h.f. radiator is the *electromagnetic horn*, of which there are two principal types: the sectoral horn and the conical horn (Fig. 2).

Shape of these devices is such that microwave energy is concentrated in a particular direction. The field pattern is influenced by the shape of the horn, mouth and throat dimensions, and the *mode* of waves being radiated. The horn is normally fed by a wave guide system. It may be considered as a sort of transformer, which matches the wave guide to free space and thus prevents standing waves within the wave guide feed system.

For proper matching, the width of the mouth should be about 10 wave lengths. Height of the mouth, not so critical a dimension, can be between 4 and 7 wave lengths. The throat of the horn normally will be the same size as the cross-sectional dimensions of the waveguide feed system. Electromagnetic horns may be used over a moderately wide band of operating frequencies.

Specific design data is beyond the scope of this article, primarily because of complications introduced through the use of wave guides. Actual or practical dimensions will vary according to the *mode of such* waves, and according to phase distribution at the mouth of the horn. Waves must be in phase for maximum forward radiation. The amount of directivity is determined by the amount of flare, or flare angle, between the throat and mouth of the horn.

The same horn can be used for both transmitting and receiving. But waves must be of correct mode and polarization for a particular horn. Horns are more often used for transmitting than receiving.

All types are constructed of thin sheet metal. All inner surfaces and joints must be smooth and well polished.

The *sectoral horn* (A in Fig. 2) is commonly used because it can be excited by any of several different wave modes, the nature of each mode influencing the design of the actual horn. This type of horn is also used to feed or radiate energy into parabolic reflectors. The sectoral horn is flared in only one dimension. The optimum flare angle is between 40 and 60 degrees when the length of the horn (distance from throat to mouth) is about the same as the width of the mouth. Most of these values can be determined experimentally, without involved calculations or without knowledge of the exact mode being transmitted or received.

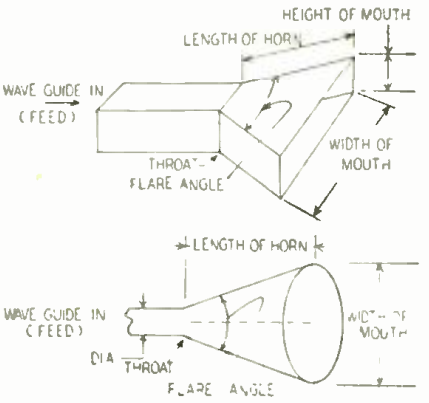


Fig. 2—Sectoral and conical horn radiators.

Field pattern will favor directivity in the plane perpendicular to the width of the horn mouth. Thus the horn shown

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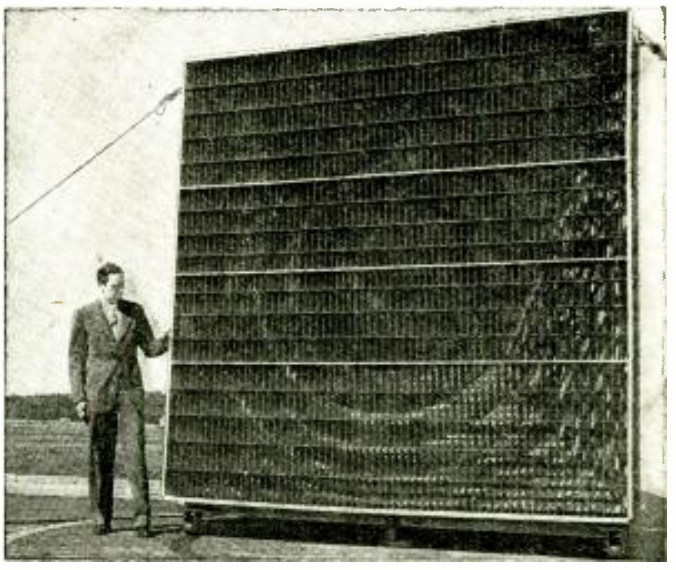

Photo Bell Telephone Laboratories

Photo A—A large stepped metallic lens used for 5-centimeter waves.

Transatlantic News

From our European Correspondent, Major Ralph Hallows

IT is just fifty years since Professor J. J. Thomson, lecturing on the cathode-ray tube, identified the constituents of the beam as minute charged particles, far smaller and lighter than the atom, which had been regarded until then as the smallest body in existence. Thomson called the particles "corpuscles." The name *electrons* was not adopted for them until some time later, though it was first coined in 1874 by Professor Johnstone Stoney of Dublin. In that year Stoney suggested that an electric current consisted of a stream of negatively charged particles, which he called *electrons;* but not till after Thomson's work on the cathode-ray tube could his conjecture be proved. The universal adoption of the name *electron* is a fitting recognition of Johnstone Stoney's early work. We've advanced a long way since then, and, thanks to the development work done on atomic energy, we have revised our ideas about the nucleus of the atom during the past few years. Not many physicists now agree with the statement still to be found in textbooks, that the nucleus contains individual electrons. The coming idea seems to be that it consists of neutrons and protons only. But just what forces bind the nucleus into a bundle so tight and so difficult to unfasten is still a mystery. There is still in some of our atomic laboratories a feeling that neither the neutron nor the proton may be found eventually to be simple bodies. The proton, representing unit positive charge, weighs 2,000 times as much as the electron, which represents unit negative charge. This scarcely seems a logical arrangement, and it is now being suggested that both neutron and proton may be conglomerations of about 1,000 positive particles, or positrons, and 1,000 negative particles, or electrons. If that is so, a neutron is composed of an equal number of positrons and electrons, the positive and negative charges neutralizing one another; the proton would contain, say, 1,001 positrons to 1,000 electrons and so have unit positive charge. Should this idea have any foundation in fact, the breaking up of the neutron might furnish a vastly richer store of atomic energy than nuclear fission.

Measuring instruments

Among the most interesting of the radio and electrical appliances at the recent exhibition of the Physical Society in London were the high-grade multirange devices for measuring current and voltage. I was much struck by one with 40 a.c. and d.c. ranges, which is certified to have sub-standard accuracy on d.c., and on a.c. to work within one-half of the tolerances permitted by the British Bureau of Standards for first-grade instruments. Another instrument gives a full-scale deflection for 50 microamperes of direct current, the resistance of the voltage scales being 20,000 ohms per volt. Voltage up to 2,500 can be measured. On a.c. the resistance is 1,000 ohms per volt and there is again a range up to 2,500 volts.

I wonder whether American radio servicemen know the joys of using the wobbulator? The "wob" was originally a radar device, designed to vary the recurrence frequency of the transmitter and so to enable two radar stations to operate on the same carrier frequency without giving rise to mutual interference. Now a wobbulator attachment has been designed for the service oscillator and a very interesting type was shown at the exhibition. Attached to the oscillator, it acts as a frequency modulator and enables response curves to be obtained on the oscilloscope when receiver alignment is in progress.

Radio for the deaf

One of the neatest radios that I have seen among those especially designed to enable the deaf to listen to broadcast programs has just made its appearance here. It may be described as a family model, being just as useful to those of the household who have normal hearing as to the deaf person. With the switch in position 1 the set works as an ordinary radio, operating its loudspeaker. Position 2 cuts in a pair of headphones in parallel with the speaker and provided with separate tone and volume controls. In position 3 the loudspeaker is out and only the headphones are working. Put the switch to position 4 and the apparatus becomes a deaf-aid, enabling the person with defective hearing to take part in general conversation. This is done by bringing into action a crystal microphone with a wide angle of response, connected through the a.f. circuits to the headphones. In addition the headphone circuit is provided with an output limiter to protect the deaf listener from the shock of sudden loud noises. You can imagine what a blessing this instrument may be to those who suffer from one of the greatest of all handicaps, defective hearing.

Mountain antenna mast

One of the most remarkable antenna systems ever erected is in a lonely spot in southwest England. The British Post Office has for many years been operating at Rugby a high-powered 16-kc transmitter, used for world-wide contacts. During the war it was decided to erect a reserve transmitter of the same power output rating at a place less liable to enemy bombing. Actually the work was done only just in time, for the Rugby station was eventually put out of action. The biggest problem in building the new station was antenna masts. Only 3 600-foot self-supporting towers were available and there was no possibility of making others, since every ounce of steel was required for other purposes. As the antenna current was to be some 400 amperes with a peak voltage of 220,000, a large antenna capacitance was necessary. It did not seem possible to produce the required 15,000-20,000 μμf with any arrangement using only the 3 towers as supports. Then the idea illustrated in the drawing was thought of. As a result of intensive high-speed reconnaissance a hill 1,200 feet high with a very steep forward slope was found in a suitable locality in the West Country. The

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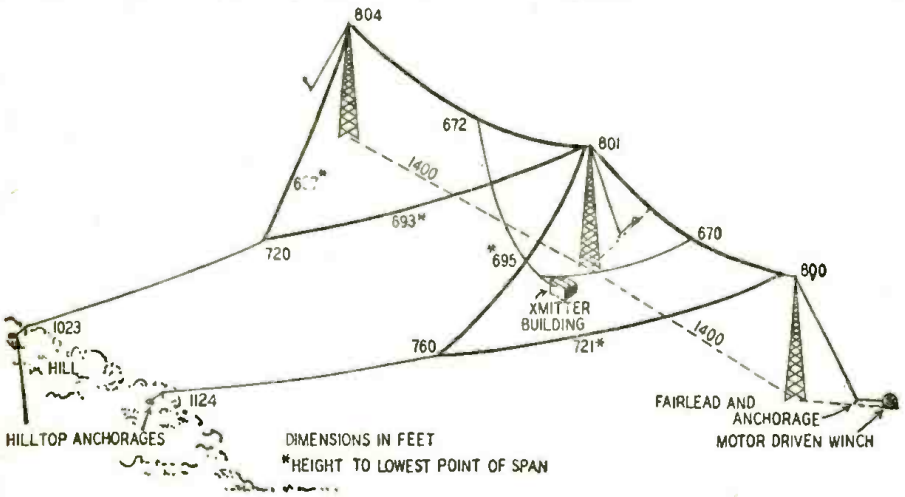

How a mountain is used as a mast to support one end of a giant broadcasting antenna.

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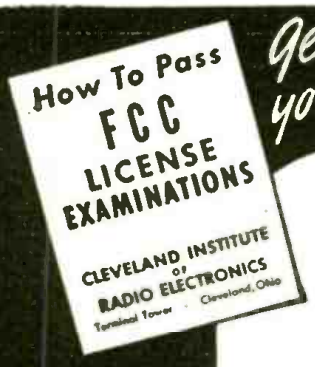

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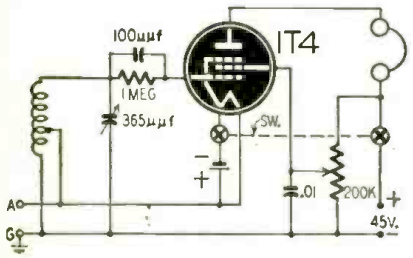
RADIO-ELECTRONIC CIRCUITS

SIMPLE ONE-TUBER

This writer has found the following simple receiver to give good service. One of the important features of the circuit is that the A-battery—an ordinary flashlight cell—is mounted in an insulated holder and operated at a slight r.f. potential above ground. This makes it possible to use an electron-coupled oscillator circuit with a filament tube. The coil is made from a standard tuned r.f. antenna coil with its windings con-



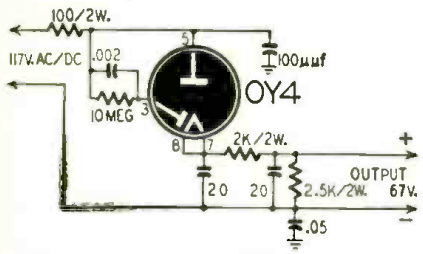
nected in series and the positive side of the filament connected to the junction of the coils. Regeneration is controlled by varying the voltage on the screen grid.

The entire unit, with the exception of the B-battery, is mounted in a metal box 3 x 4 x 5 inches. For good service, a ground is as important as the antenna. In an efficient installation north of Boston, Mass., stations were received from as far away as West Virginia.

GRAY TREMBLY,
Somerville, Mass.

HEATERLESS B SUPPLY

The low-voltage power supply shown here was designed to replace the B-battery in a camera-type portable radio. It can be built into a unit compact enough to fit into the space formerly occupied by the battery it replaces. No provisions were made for a filament supply, since it was not wished to rewire the filament circuit of the set. If the set has series filaments, a suitable dropping resistor may be used to make the set all-electric.

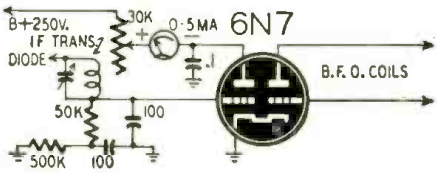


The supply uses an OY4 cold-cathode half-wave rectifier. The starter anode is connected to the main anode through a 10-megohm resistor, by-passed with a 0.002-μf condenser. The No. 7 and No. 8 pins must be connected together at the socket.

W. G. MURRAY,
Ottawa, Canada

ADDING AN "S" METER

I have found the accompanying simple circuit useful for adding an "S" meter to a receiver. In this case a 6N7 was used as a meter amplifier and beat-



frequency oscillator. The grid of the amplifier section is connected to the high side of the detector load resistor, and its plate voltage is obtained through a variable, 30,000-ohm, wire-wound resistor. If a variable resistor is not available, an adjustable resistor may be used, since the setting is seldom varied once it has been adjusted.

D. STAPLETON,
Grahamstown, South Africa

RADIO-CRAFT will present a new feature on this page next month. Watch for it! If readers approve, it may be adopted as a permanent feature of the magazine.

USEFUL POWER SUPPLY

The variable output power supply described here has many applications in the radio service shop, but I find it most useful for re-forming wet and dry electrolytic condensers.

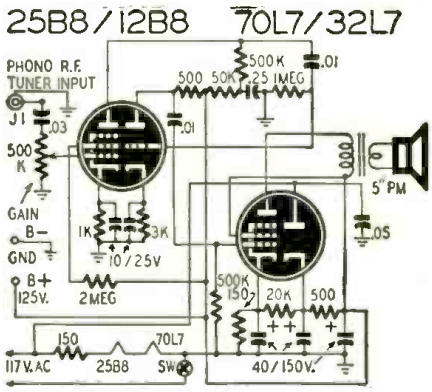
To re-form condensers, connect the condenser to the terminals of the supply—be sure to observe polarity—and adjust the input voltage so that the d.c. output is equal to the working voltage of the condenser. As the condenser reforms, its internal resistance increases, necessitating an adjustment in the input voltage. Most condensers will be completely re-formed in from 10 to 15 minutes, depending on the amount of leakage present. In cases where excessive current is drawn it will be necessary to apply about one-half working voltage and increase it in small steps every 10 or 15 minutes until working voltage is reached.

All new electrolytic condensers, or those that have not been used for several months, should be re-formed before installation to prevent overloading the rectifier tube.

W. M. FINLEY, JR.
Norfolk, Ark.

TEST AMPLIFIER

Here is a circuit of a handy little test amplifier and "B" supply that can be built very compactly on a small chassis. It can be used to test phono pickups, r.f. tuners and microphones.



The signal from the device under test is picked up by a probe that is plugged into J1. This signal is applied to the grid of the pentode section of the 12B8 amplifier. The triode section of this tube functions as a detector for r.f. voltages. The output of this detector is amplified by the pentode section of the 32L7 and coupled to the speaker.

The power supply consists of a half-wave rectifier utilizing the diode section of the 32L7. The output of the supply is connected to pin jacks on the panel so the voltage may be used to power small devices.

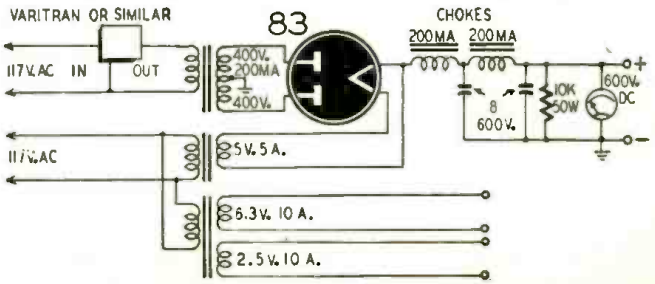
F. U. DILLION
Hollywood, Calif.

CODE OSCILLATOR

A midget receiver can be made into an efficient code oscillator with only one additional part and two circuit connections.

The part required is a paper condenser of .002 μf at 400 volts d.c. working. One lead of the condenser is connected to the plate of the output tube and the other is brought to a tip jack. A lead is then brought out from the

(Continued on page 61)



TECHNOTES

. . . . PHILCO CHANGERS

Failure of the reject mechanism on Model 35-1285 and 35-1289 record changers is due to failure of the slug in the reject solenoid to return to the correct position. This slug should be lengthened by unscrewing its two parts that are held together by threads.

A. W. POWELL,
Rolfe, Iowa

. . . . SILVERTONE 7036

Flickering dial lamps and fading are usually caused by loose rivets holding the sockets to the chassis. One side of the heater circuit returns to the chassis through one of the rivets on each socket. To remedy this complaint, solder a wire from each socket return directly to the chassis.

STANLEY A. KUBIT,
La Porte, Ind.

RADIO-CRAFT wants *Technotes* describing common troubles of well-known receivers or telling how rare or difficult problems were solved. A six-month subscription will be awarded for each unillustrated and a one-year subscription for each illustrated *Technote* published.

. . . . CHEVROLET 985695

These sets have been found to develop severe noise when the car is accelerating or being driven over a rough road. In three out of four cases, I have traced the trouble to a broken lead on the tone-control condenser. This condenser is located below the tone control switch and is connected from one side of the latter to the grounded side of the volume control. The break often occurs on the switch side of the condenser and is not easily detected during a visual inspection.

JOHN W. FINDARLE,
Modesto, Calif.

. . . . SIGNAL TRACER

I constructed the signal tracer described on page 756 of the August, 1946, issue of RADIO-CRAFT. It would motorboat when the gain control was advanced. I checked the circuit and found that a 50,000-ohm resistor and a 0.05-μf condenser were used as a decoupling filter in the plate circuit of the 6Q7-GT. I replaced the condenser with a 10-μf, 450-volt electrolytic and my troubles were over. Now the tracer works perfectly at all volume levels.

G. ABLEMAN,
Edgerton, Wis.

. . . . HUM REDUCTION

Recently I came across a set with a severe case of hum that was eventually traced to the converter stage. This stage used a thin wafer socket. The hum was caused by high-resistance leakage between the filament and grid prongs. Replacement with a socket of the molded bakelite type cleared up the trouble completely.

J. DUBOVY,
Bronx, N. Y.

. . . . SILVERTONE 4500

When hum in these sets is not traceable to condensers or tubes, try replacing the volume control. This control may open at one end while still controlling the volume effectively.

STANLEY A. KUBIT,
La Porte, Ind.

. . . . 1LE3 IN PORTABLES

The 1LE3, used in many portables, can be replaced with a 1LH4 if the No. 4 (diode plate) pin is clipped off. When this tube is substituted for the 1LE3 in oscillator circuits, it will often provide clearer reception and lower noise level.

JOSEPH H. BLACHE,
Denver, Colo.

. . . . WEAK R.F. AND I.F. AMPLIFIERS

Failure of i.f. and r.f. amplifiers is sometimes due to moisture in the coil assemblies. I remove the coils and dip them, padders up, in a can of hot paraffin. The heat dries out the padders, and the paraffin drives the moisture out of the coils and seals them against the effects of dampness. It is wise to tie the coils to the forms to prevent them from slipping while in the paraffin.

W. E. HARDIN,
Paintsville, Ky.

. . . . ALIGNING KINK

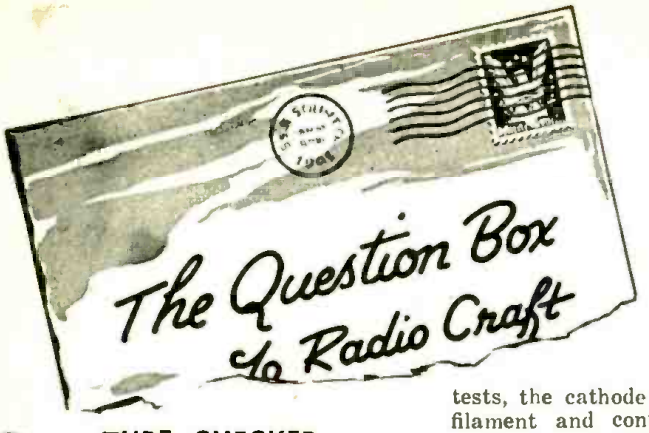
It has been my experience, when aligning sets with two or more i.f. stages, that the set will approach ideal alignment only to break into oscillation as the stages are brought into resonance. I find that by applying degeneration to each stage this problem is eliminated. Disconnect the cathode by-pass condensers of the i.f. stages and remove the grid returns for these stages from the a.v.c. line or ground and connect them to the cathodes, leaving the condensers disconnected. This reduces the sensitivity enough to check oscillation. Upon completing the alignment, replace the original connections and the sensitivity of the set will return to normal.

J. W. ESSEX,
Halifax, N. S.

. . . . SHORTED CONDENSERS

We all know how disconcerting it is to have to repair a radio in which the tuning condenser is warped and the plates scraping. Sometimes only one outside plate may be shorting and we can see this and correct it. At other times the plate may be in the center and not visible. My method is to disconnect the r.f. coil leads—only one is really necessary—and place a single dry cell across the section of the condenser which is causing the trouble. As the shaft is rotated, a distinct spark will be seen at the point or points of contact. Any sort of tool then may be used to straighten the plate. There are times when a spark may not be seen, but on careful examination after the condenser is out of mesh, a dark blemish on the rotor plate may be seen. This, of course, is caused by the current flowing across the plates when contact is made, with the plates so closely meshed that no spark results.

GEORGE E. HUFF,
Miami, Oklahoma



Question Box queries will be answered by mail and those of general interest will be printed in the magazine. A fee of 50c will be charged for simple questions requiring no schematics. Write for estimate on questions that may require diagrams or considerable research.

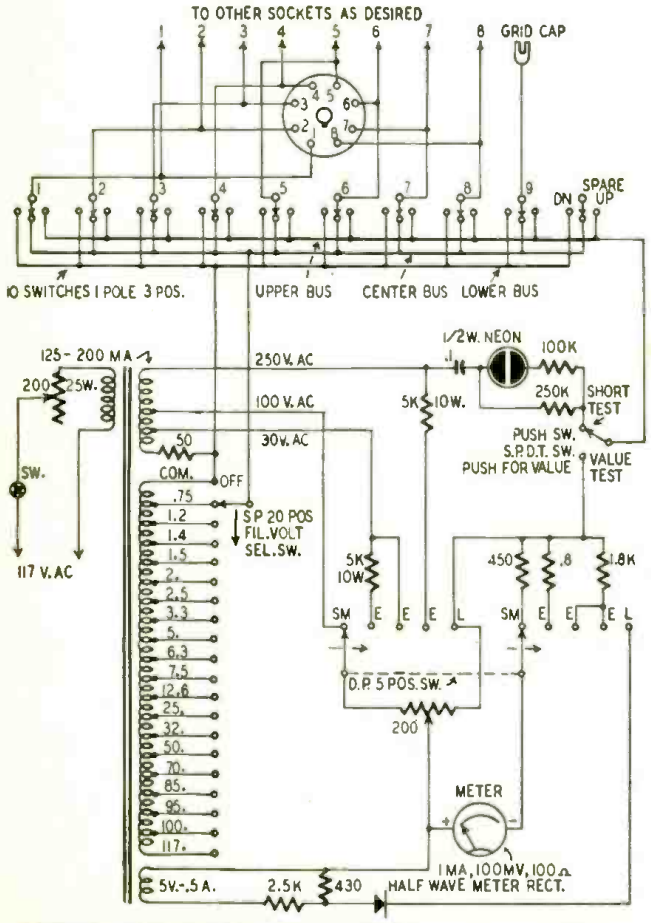
TUBE CHECKER

I would like to have a diagram of a tube checker that can be used for making mutual conductance as well as emission tests.—E.W.Z., Harrisburg, Penna.

A. The tester diagrammed provides a means of making mutual conductance and emission tests and will also check a tube for interelectrode shorts. It may be calibrated by testing tubes of known quality and recording the meter reading and control settings.

Emission tests are made by connecting the cathode and one side of the heater or filament to the lower bus and all other elements to the upper bus. The remaining heater connection is made to the center bus. The total current is taken as an indication of tube quality.

When making mutual conductance tests, the cathode and one side of the filament and control and suppressor grids are connected to the lower bus, the other heater to the center bus and plate and screen leads to the upper bus.



SM: TRANSCONDUCTANCE; E: EMISSION; L: LINE CHECK

SERVO AMPLIFIER

I have purchased a surplus servo amplifier using a pair of 7C5's, a 7F7, and a 7Y4. The power transformer is designed to work from a 110-volt 400 cycle power source. Can this be used? I would like to have a diagram showing how these parts may be used in a single-ended phono amplifier. Is it possible to use the 7F7 as a mixer so that I can use a T-17 surplus microphone, at the same time that I am playing the phonograph.—F.E., New Britain, Conn.

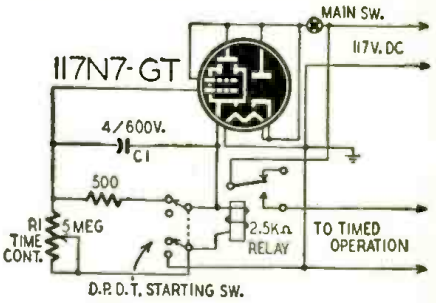
A. The diagram is shown. The T-17 is a single-button carbon microphone fitted with a special type plug. A Mallory type JK-33-A 3-circuit mike jack should be used on the amplifier panel. Voltage for the mike is obtained by tapping the 7C5 cathode resistor. The microphone transformer should be designed for coupling a 200-ohm single-button microphone to a grid.

The 400-cycle power transformer cannot be used on 60-cycle a.c. and should be replaced by a standard unit supplying 285 to 300 volts each side of center tap at 75 to 100 ma. Use either a speaker field or filter choke.

ELECTRONIC TIMER

I would like to have a circuit of an electronic timer with a timing range of from about 1/10 of a second to 2 seconds. The device is to operate from a 117-volt d.c. line.—J.I.C., New York, N. Y.

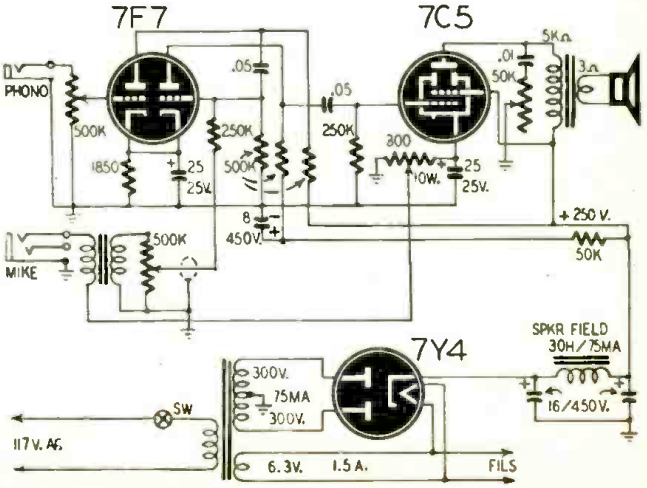
A. An electronic timer is shown. The duration of the time cycle is determined by the setting of the 5-megohm time control and the capacity of its associated condenser. The values shown should give the required range. The 4-μf



condenser C1 should be a low-leakage paper or oil-filled type. The condenser and time control should be mounted on low-loss material so that the time constant will not be disturbed. The 2,500-ohm relay should be equipped with s.p.d.t. contacts, with either circuit-opening or circuit-closing action.

If other time ranges are desired, an increase in the value of R1 or C1 will increase the length of the cycle.

A 50L6 or 25L6 may be used if a filament-dropping resistor is used.



CUSTOMERS AND LAUGHS

Customers are the queerest of animals, yet without them we not only would starve but would miss out on a good many hearty laughs in life. Some of the incidents which have been most annoying to me, yet amusing, follow:

A very irate customer appeared at the wicket one day asking for the manager of the department. Upon my appearance he let forth with a strange language which made even my hardened ears burn. From what I could piece together of the conversation I gathered that the d - - - tubes that he bought from us were no d - - - good. They would work not too badly on 1¼ volts (for which they were intended) and on 4 volts they played a little louder, but when 6 volts was applied they didn't play at all, and he wanted a refund.

Another customer placed an order for a new grid leak for his radio. He knew that the old one in his set was defective as there was no sign of moisture under it!

A new car radio was brought back under warranty service for repairs. Upon inspection the top of the vibrator can was found to be cut off. When the customer was questioned he admitted performing the operation with a can opener "to find out what was rattling inside."

Then there is the customer who complains, "My radio hasn't played a day since it was repaired" (especially if you are trying to collect an overdue account). Upon investigation you find all the trimmers of the i.f. transformers tightened down and half the parts missing.

Servicemen are turning gray overnight trying to convince the public that the trouble is not necessarily in the speaker because it won't speak.

The following telegrams add to the fun. "Rush two 6V67-GT tubes." Upon advice of "No such tube," reply came, "Rush instead two 6K67-GT tubes." The second time advice went "No such tube," and the reply came, "Rush instead two 6K6-GT tubes." These were forwarded, to be received back the next day with the following notation, "These do not fit set, please rush two 6JK7-GT tubes." Just about then some customer has to wander in and demand a condenser for a 1937 Victor radio.

Put everything together and radio servicing is still a great trade despite the customers.—*Raymond E. Wice*

INEXPENSIVE CHASSIS

To make a cheap, neat-looking chassis, take a cigar box and drill into it the necessary holes for mounting sockets and other parts. Cover the box with masking tape and punch through the previously cut holes. Excess tape is pulled through the holes and fastened to the underside of the box. Black cloth bookbinders' tape also makes a neat-looking job.

ALLEN J. SCHWARTZ,
Albany, N. Y.

New Radio-Electronic Patents

By I. QUEEN

VOLTAGE REGULATION

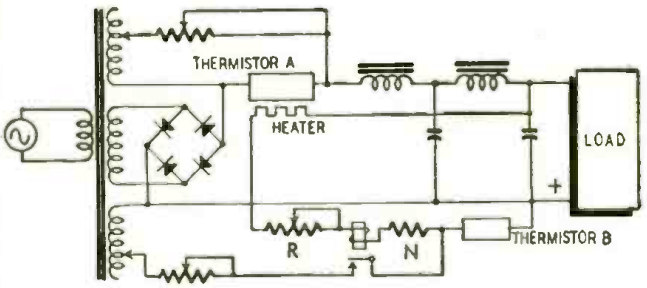
James A. Potter, Rutherford, N. J.
(assigned to Bell Telephone Labs., Inc.)
Patent No. 2,413,033

A thermistor is a simple element whose resistance changes with a change in temperature. This makes it useful as a voltage or current regulator or control unit. In this power supply two thermistors are used to maintain an output voltage which is constant within very narrow limits.

Thermistor A is connected across an auxiliary secondary winding to maintain its operating temperature near the optimum value. If the load increases more current flows through A and there is a tendency for the voltage of the system to drop. The increased current reduces the resistance and voltage drop across the thermistor. Therefore the output change is compensated for.

Thermistor B is connected in a circuit which shunts the output. If the load increases as before, the voltage across the shunt circuit tends to drop off and therefore the thermistor voltage will be lower. This corresponds to an increase in current through B and through the heater H of A. The indirect heating of A adds to its control effectiveness and practically eliminates any change in the output voltage.

Although it is desirable that the shunt circuit current undergo a large change for small load changes, this effect in itself constitutes a change in the effective load of the power supply, which in itself would oppose good regulation.



The problem is solved by including a variable resistor R and a negative-coefficient resistor N (such as silicon carbide) in the shunt circuit. When the variable resistor is correctly adjusted, the sum of voltages across the components will be practically constant over the operating current range.

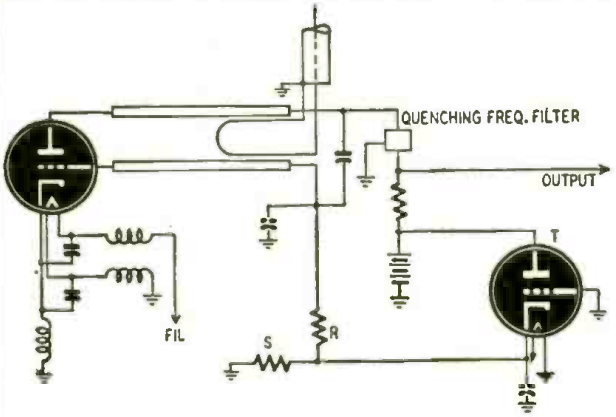
IMPROVED SUPERREGENERATOR

Joseph A. Worcester, Jr.
(assigned to General Electric Co.)
Patent No. 2,410,768

The superregenerative circuit is deservedly popular for reception of u.h.f. signals. It combines broad tuning (for wide-band modulation), extremely high sensitivity and simple design. Positive feedback is used to provide enormous amplification but sustained oscillations are prevented by periodically damping or interrupting the circuit. In its simplest form, the superregenerator includes its own means of quenching or interruption. Feeble oscillations are rapidly built up until the grid becomes sufficiently positive to attract an appreciable quantity of electrons. This charge is trapped because of a very large grid resistor. The negative bias then cuts off plate current until the charge can leak off.

Because of the inoperative periods required for a charge to leak off, the modulation frequency which can be received without distortion is much lower than the interruption rate. In the self-quenching circuit the latter ordinarily may be made as high as 50 kc by proper design of grid constants R and C. Therefore the circuit will distort FM or video signals, which have very wide modulation bands. This new circuit permits increasing the upper quenching limit to approximately 1 mc, thus making it suitable for television and other wide-band signals.

Few additional components are necessary. The oscillator grid leak is composed of two resistors, R and S. Resistor S is about 1 megohm and is much larger than resistor R. The grid cathode of tube T is connected across S. Ordinarily the tube is nonoperative because of the cathode bias produced at S, but when oscillator grid current flows through this resistor the tube conducts. This short-circuits S and permits a rapid discharge of the electron accumulation on the oscillator grid.



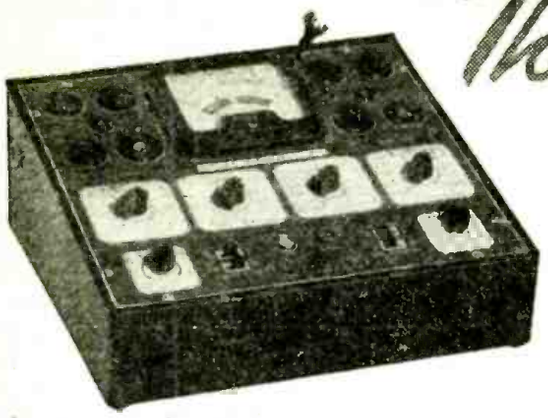
SUB-HARMONIC GENERATOR

Stuart W. Seeley
(Assigned to Radio Corp. of America)
Patent No. 2,403,559

It is often inconvenient to generate a desired frequency directly, so either multiplication or division must be resorted to. Frequency multiplication is a natural process since any distorted wave contains harmonics and it is only necessary to pick out the required one. Special circuits may be designed to generate sub-harmonics.

In this circuit sub-harmonics are generated by suppressing portions of the fundamental wave. For example, the third sub-harmonic is obtained by suppressing two cycles and permitting only

(Continued on page 73)



The New Model 60-T TUBE and SET TESTER

A COMPLETE TUBE TESTER

Tests all tubes including the new post-war miniature loctals such as the 12AT6, 12AU6, 35W4, 50B5, 117Z3, etc. • Tests by the well-established emission method for tube quality, directly read on the scale of the meter • Tests shorts and leakages up to 3 Megohms in all tubes • Tests leakages and shorts of any one element against all elements in all tubes • Tests both plates in rectifiers • Tests individual sections such as diodes, triodes, pentodes, etc., in multi-purpose tubes.

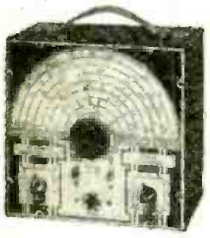
Model 60-T operates on 90-120 Volts 60 Cycles A.C. Housed in sloping leatherette covered cabinet. Comes complete with test leads, tube charts and detailed operating instructions.

A COMPLETE MULTI-METER

- 6 D.C. Voltage Ranges: 0 to 7.5/15/75/150/750/1,500 Volts
- 6 A.C. Voltage Ranges: 0 to 15/30/150/300/1,500/3,000 Volts
- 4 D.C. Current Ranges: 0 to 1.5/15/150 Ma. 0 to 1.5 Amps.
- Low Resistance Range: 0 to 2,000 Ohms (1st division is 1/10th of an ohm.)
- 2 Medium Resistance Ranges: 0 to 20,000/200,000 Ohms
- High Resistance Range: 0 to 20 Megohms
- 3 Decibel Ranges: −10 to +38 +10 to +38 +30 to +58 D.B.

$49.85 NET PRICE

EXTRA: WE CAN NOW SUPPLY THE MODEL 60 HOUSED IN A BEAUTIFUL HAND-RUBBED OAK CABINET. COMPLETE WITH PORTABLE COVER MAKING IT SUITABLE FOR EITHER BENCH OR OUTSIDE USE. ONLY $2.75 ADDITIONAL. SPECIFY MODEL 60-C.



THE NEW MODEL B-45

SIGNAL GENERATOR

Complete, ready to operate **$27.75**

Self-modulated — provides a highly stable signal. RF frequencies from 150 Kc. to 12.5 Mc. on Fundamentals and from 11 Mc. to 50 Mc. on Harmonics. Modulation is accomplished by grid-blocking action — equally effective for alignment of amplitude and frequency modulation as well as for television receivers. Self-contained batteries. All calibrations are etched on the front panel, permitting DIRECT READING.

Model B-45 uses a beautifully processed dualtone front panel. Comes housed in a heavy-gauge crystalline steel cabinet complete with shielded test lead, self-contained batteries and instructions.

THE NEW MODEL 670

SUPER METER $28.40

A Combination VOLT-OHM-MILLIAMMETER plus CAPACITY REACTANCE, INDUCTANCE and DECIBEL MEASUREMENTS

D.C. VOLTS: 0 to 7.5/15/75/150/750/1500/7500.
A.C. VOLTS: 0 to 15/30/150/300/1500/3000 Volts.
OUTPUT VOLTS: 0 to 15/30/150/300/1500/3000.
D.C. CURRENT: 0 to 1.5/15/150 Ma.; 0 to 1.5 Amps.
RESISTANCE: 0 to 500/100,000 ohms 0 to 10 Megohms.
CAPACITY: .001 to .2 Mfd., .1 to 4 Mfd. (Quality test for electrolytics).
REACTANCE: 700 to 27,000 Ohms; 13,000 Ohms to 3 Megohms.
INDUCTANCE: 1.75 to 70 Henries; 35 to 8,000 Henries.
DECIBELS: −10 to +18, +10 to +38, +30 to +58.
The Model 670 comes housed in a rugged, crackle-finished steel cabinet complete with test leads and operating instructions. Size 5½" x 7½" x 3".

THE NEW MODEL CA-11

SIGNAL TRACER $18.75

Simple to operate . . . because signal intensity readings are indicated *directly on the meter!*

★ SIMPLE TO OPERATE — only 1 connecting cable — NO TUNING CONTROLS.

★ HIGHLY SENSITIVE — uses an improved Vacuum Tube Voltmeter circuit.

★ Tube and resistor-capacity network are built into the Detector Probe.

★ COMPLETELY PORTABLE — weighs 5 lbs. and measures 5" x 6" x 7".

★ Comparative Signal Intensity readings are indicated directly on the meter as the Detector Probe is moved to follow the Signal from Antenna to Speaker.

★ Provision is made for insertion of phones.

The Model CA-11 comes housed in a beautiful hand-rubbed wooden cabinet. Complete with Probe, test leads and instructions.

THE NEW MODEL 450

TUBE TESTER $39.50

Speedy operation — assured by newly designed rotary selector switch which replaces the usual snap, toggle, or lever action switches.

SPECIFICATIONS

• Tests all tubes up to 117 volts. • Tests shorts and leakages up to 3 Megohms in all tubes. • Tests both plates in rectifiers. • New type line voltage adjuster. • Tests individual sections such as diodes, triodes, pentodes, etc., in multi-purpose tubes. • Noise Test-detects microphonic tubes or noise due to faulty elements and loose internal connections. • Uses a 4½" square rugged meter. • Works on 90 to 125 volts 60 cycles A.C.

EXTRA SERVICE—May be used as an extremely sensitive condenser Leakage Checker. A relaxation type oscillator incorporated in this model will detect leakages even when the frequency is one per minute.

OUR POLICY

We do not advertise any unit which is not available for immediate shipment from stock. • Less flowery adjectives, more detailed specifications. • All units are sold subject to one year guarantee except when components are damaged through misuse. • We do not solicit orders for any unit that does not meet our requirements for accuracy and honest value. Any item purchased from us is sold with the understanding that it may be returned for full refund after a 10 day trial.

20% DEPOSIT REQUIRED ON ALL C.O.D. ORDERS

GENERAL ELECTRONIC DISTRIBUTING CO.

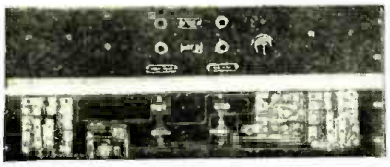
DEPT. RC-7, 98 PARK PLACE, NEW YORK 7, N. Y.

NEW RADIO·ELECTRONIC DEVICES

TRANSMITTER CONTROL

Ward Leonard Electric Co.
Chicago, Ill.

Offered in kit form or completely assembled and wired, this new transmitter control panel is completely interwired and mounts directly on a standard relay



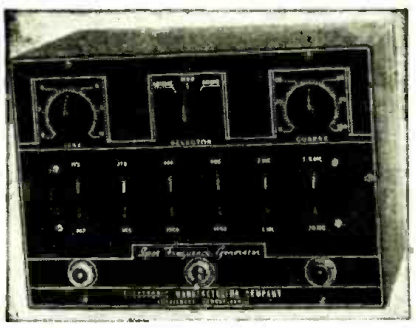
rack. It gives full automatic protection against damage to tubes, transformers, and other gear from overloads and power failures and provides finger-tip control of filament and plate supply. ON filament push button starts transmitter; ON plate button puts station on the air; OFF plate button interrupts transmission; and OFF filament button shuts down station.

The Ward panel includes 1 filament relay; double-pole, 15-ampere contacts; 1 plate relay, double-pole, 15-ampere contacts; 1 time-delay relay; one overload relay (250 or 500 ma); 2 push buttons each for filament and plate supply. Panel size is 3½ by 19 inches, with 1¾-inch maximum depth behind panel, furnished in gray or black crackle finish. For 115 volts, 60-cycle a.c.—RADIO-CRAFT

SIGNAL GENERATOR

Electronic Mfg. Co.
Harrisburg, Penna.

The Model 200 spot frequency generator features 12 preset frequencies chosen to cover adequately the most commonly used receiver test channels. Six switches are provided, with only a flip of a switch necessary to select or change to any desired frequency. Spot frequencies are: 175, 262, 370, 455, 456, 600, 1000 and 1400 kc and 2, 5, 7.5, and 20 mc. A single output jack makes it unnecessary to switch test leads.



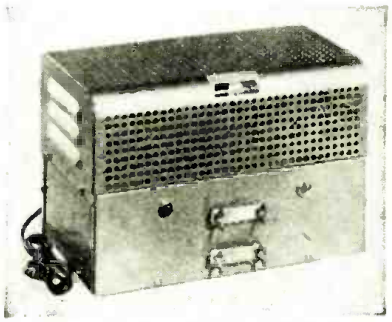
Stability is maintained by an electron-coupled circuit and low leakage is assured through use of double shielding. The generator attenuates to less than 1 microvolt. It is a.c. operated. The panel is of acid-etched aluminum with a steel case.—RADIO-CRAFT

LOW-VOLTAGE D.C. SUPPLY

Electro Products Laboratories, Inc.
Chicago, Ill.

The Electro Model A power supply is designed for use by servicemen on automobile, aircraft, and marine radio equipment operating from 6- or 12-volt d.c. supplies. It consists of two 6-volt, 7.5-ampere filtered d.c. power sources which can be placed in series to deliver 12 volts at 7.5 amperes or in parallel for 6-volt, 15-ampere operation.

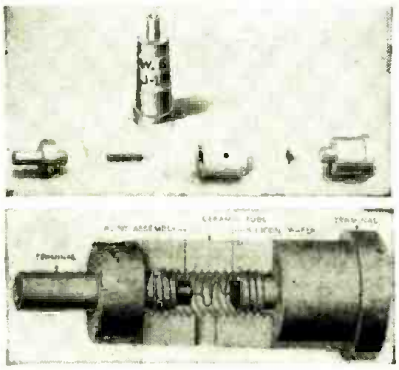
The unit operates from 105-, 115-, or 125-volt, 60-cycle power sources and weighs 31 pounds.—RADIO-CRAFT



CRYSTAL RECTIFIERS

Western Electric Co., Inc.
New York, N. Y.

Silicon crystal rectifiers used in radio links, telephone apparatus, portable test equipment, and other electronic circuits where space, power consumption, per-



formance, and economy are paramount, are announced by the Western Electric Company. These include the 1N21B, 1N23A, 1N23B, 1N25, 1N26, 1N28, and 1N31. They are applicable as frequency converters, low-level detectors, and instrument rectifiers. In the unshielded type, a ceramic insulator separates the point and crystal wafer; in the shielded type, a metal shield encloses the structure. These types employ point contact with the rectifying element.—RADIO-CRAFT

R.F. PROBE

Radio Frequency Laboratories, Inc.
Boonton, N. J.

This new r.f. probe is designed for functional testing of high frequency circuits. When the probe is subjected to an r.f. field, a proportionate r.f. current is induced, rectified and indicated on a Weston 506 meter. Useful for detection of standing waves, shielding power leaks, r.f. choke efficiency, and circuit tracing for r.f. in all radio frequency equipment and associated components, without affecting operation of the circuit. AM, FM and television transmitters up to 1500 megacycles, electronic heating and soldering equipment, antennas and transmission lines, and other r.f. units can be checked throughout. It is valuable in seeking out causes of lowered operating efficiency, damage or interference to neighboring equipment and components, spurious radiation and escape of r.f. into power lines.

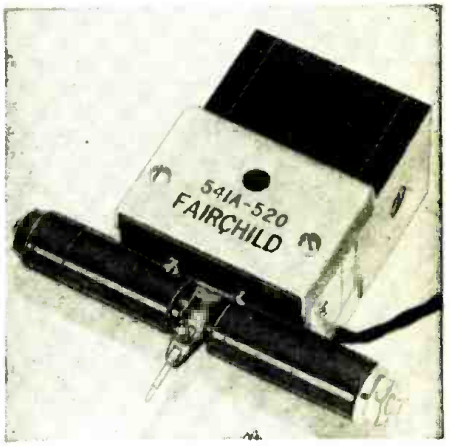
Probe element is ¾ inch in diameter, 5 inches long—RADIO-CRAFT

MAGNETIC CUTTING HEAD

Fairchild Camera and Instrument Corp.
New York, N. Y.

A new magnetic recording head complete with equalizer and standard mounting plate and suitable for any current model sound recorder is Fairchild's Unit 541A. It is designed to meet the standards of FM and AM broadcast and professional recording.

The unit has a guaranteed frequency response of ±2 db over a range of 30 to 8,000 cycles with less than 1 percent distortion at 400 cycles. Its construction provides for adjusting and maintaining the alignment of the armature without



disassembling the cutting head. The cutting head has a 500-ohm impedance and requires a 0.6-watt + 20-db power level. A stylus ⅝ inch long and 0.062 inch in diameter is used.—RADIO-CRAFT

ALUMINUM VOICE COILS

General Electric Co.
Schenectady, N. Y.

The new GE loud-speakers have aluminum instead of paper bases for voice coils. It is claimed that this type of form can handle greater power, has longer life, is unaffected by temperature or humidity, and will not warp or crack. The aluminum-based voice coil also can be produced with better control on gaps.—RADIO-CRAFT

IMPROVING THE MIDGET RADIO

MANY midget t.r.f. and superheterodyne radios make excellent tuners for record players and phono amplifiers. By using a large well-baffled speaker and possibly a tone-compensated amplifier, high-grade console radio performance may be enjoyed.

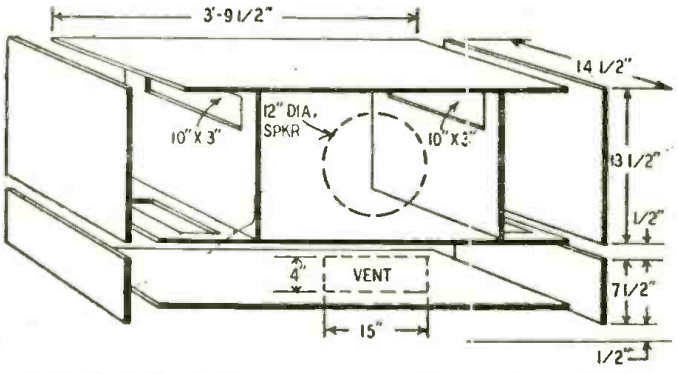
When converting the radio to a tuner, all power supply and amplifier components are removed. In most cases, the filament and high-voltage circuits of the r.f. section consume comparatively little current and all voltages may be taken from the amplifier power supply with little danger of overloading. If the radio is an a.c.-d.c. model, the filaments are connected in series. In this case, they should be connected in parallel and the leads brought to the filament prongs of the output tube socket, into which the power-supply plug from the amplifier is to be plugged. The cathode terminal of this socket is grounded to the chassis and is the tie-in point for the B-minus and common ground between the two units. The B-plus line in the tuner is connected to the plate terminal as the high-voltage input terminal.

A 0.05-μf, 600-volt blocking condenser is connected between the output of the detector (or the plate of the first audio stage if a diode-triode is used) and the grid terminal of the output socket. This terminal is connected to the input terminals of the amplifier. *(It is advisable to shield the input lead since it may pick up hum from filament leads and other sources.)*

The midget receiver itself may be improved by installing it in a baffle cabinet to extend its audio frequency range.

The bass response of many speakers can be improved through proper baffling. Where a suitable speaker enclosure or console cabinet is unavailable, efficient baffles can be built into bookcases, cabinets, radiator enclosures, and other household furnishings. A drawing of a suitable baffle is shown in the figure. (The front and back panels are not shown in the drawing, dotted lines show vent and speaker openings.) For improved appearance the unit is designed to be enclosed within some sort of case or cabinet. The case shown is assembled from celotex panels either ½ or ¾ inch thick. Screws and glue are used on the assembly. The inside ports are 3 by 10 inches. Dimensions of the vent vary directly as the speaker size. Values given are for a 12-inch speaker.

This baffle, of course, will not help an overloaded speaker, a frequent source of distortion in midgets. If volume is kept down, however, it will produce a startling improvement in the quality of any midget. In such a case the midget is mounted right inside the baffle (after removing the cabinet). Speaker and vent holes are scaled down to proper size and, where desirable, all dimensions can be decreased, with some sacrifice of low-note response.—*John Kwietinskas*



A simple baffle of this type improves midget quality startlingly.

WORLD-WIDE STATION LIST

Edited by ELMER R. FULLER

SO far this summer conditions have not been very good, and very little real dx has been heard. Dx on the ham bands has been very good a few times, and several overseas countries have come in. Hams have been heard from Guatemala, Mexico, Puerto Rico, Canada, Swan Island, Guantanamo Bay, Barbados, Antigua, Haiti, Costa Rica, Newfoundland, Alaska, Panama, Bermuda, Jamaica, Canal Zone, Nicaragua, Montserrat (small island in the B. W. I.), British Honduras, Labrador, Honduras, Greenland, Martinique, Grenada, Dominican Republic, Bahama Islands, Saint Lucia, Virgin Islands, Trinidad, Brazil, Argentina, Ecuador Uruguay, Colombia, Peru, Curacao, British Guiana, Chili, Venezuela, Paraguay, Dutch Guiana, Isle of Man, Wales, England, Scotland, Germany, Northern Ireland, Italy, France, Belgium, Luxembourg, Switzerland, Denmark, Norway, Irish Free State, Malta, Netherlands, Greece, Spain, Gilbraltar, Portugal, Finland, Azores, Rumania, Sweden, Iceland, Sardinia Czechoslovakia, Austria, U. S. S. R., Algeria, Belgian Congo, Union of South Africa, Tangiers, Egypt, Northern Rhodesia, French Morocco, French West Africa, Liberia, Southern Rhodesia, Tanganyika, Mozambique, Gold Coast, Basutoland, Libya. Kenya, Hawaii, Guam, Australia. New Zealand, Canton Islands, Papua, Tasmania, Marshall Islands, Philippine Islands, Netherland Indies, Wake Island, Palmyra Island, New Hebrides, Fiji Islands, Iwo Jima, China, Okinawa, Japan, India, Burma, Hedjaz, Iran, Aden, Iraq, Ceylon. This is certainly a good list of catches for 10- and 20-meter ham bands, and were all received on the East Coast. Many of these are easy to get, but several are considered very good dx, and some of them only have one, or very few hams located there. Send in a report on any of these hard-to-get ones you hear. We have more details on them, but space does not permit us to publish the full dope, so drop me a line on the ones you are interested in.

A station in the Soviet zone of Berlin is being operated on 6.070 megacycles, and is being heard often on the East coast. As yet we have not received a report of any call letters being used. Schedule is also unknown to us. OTC in Leopoldville, Belgian Congo, is being heard on 17.770 megacycles from 0500 to 0830 hours, and on 9.745 megacycles from 1030 to 2300 hours, EST. Programs in English are heard from 1030 to 1200 hours, 1530 to 1645 hours, and 2100 to 2300 hours. The last is beamed toward the United States, while the other two are directed to the Great Britain area.

In another month or so, the dx season will begin, and some better reports are expected. So make the best of it, and some very good dx can be picked up even in the warm summer months.

All schedules are Eastern Standard Time.

Location	Station	Freq.	Schedule
UNITED STATES			
Washington, D. C.	WWV	2.500	U.S. Bureau of Standards.
	WWV	5.000	1900 to 0700 (440-cycle note only)
	WWV	5.000	0700 to 1900 (440 and 4,000-cycle note)
	WWV	10.000	Continuously
		15.000	
		20.000	
		25.000	
		30.000	
		35.000	

All WWV frequencies broadcast time and musical pitch. All broadcast a 400-cycle note, and an additional 4,000 cycle note is sent on 5, 10, 15, 20 and 25 mc. Time is announced in code at 5-minute intervals, and voice announcements are made every half hour.

Location	Station	Freq.	Schedule
U.S.S.R.			
	Kiev	11.720	
	Komsomolsk	9.560	0100 to 0530; 1100 to 1400; 1545 to 1650; 1700 to 1830
	Moscow	5.810	schedule unknown

RADIO TERM ILLUSTRATED

Suggested by: Michael Kosinski, Brooklyn, N. Y.

"A good multiplier"

Location	Station	Freq.	Schedule
Moscow		6.030	schedule unknown
Moscow		9.680	1600 to 1745; 2315 to 2345
Moscow		7.300	1300 to 1800; 1815 to 2100
Moscow		9.180	0000 to 0100; 0530 to 0815; 1100 to 1130; 1500 to 1700
Moscow		9.650	1100 to 1220; 2200 to 2235
Moscow		9.710	2300 to 0730
Moscow		9.860	2200 to 0200; 0830 to 0930; 1000 to 1200
Moscow		11.630	1930 to 0300; 0600 to 0800; 0830 to 1300
Moscow		11.780	0900 to 1000; 2000 to 2130; 2200 to 0100
Moscow		11.830	2200 to 0600; 0730 to 0845; 1100 to 1600
Moscow		11.880	0720 to 1900; 2200 to 0600
Moscow		12.080	0800 to 1100
Moscow		15.320	0000 to 0500; 0530 to 0800; 0830 to 1100; 2200 to 2300
Moscow		15.340	2200 to 0800; 1000 to 1100
Moscow		15.230	0530 to 0830; 0915 to 0930; 1030 to 1330; 2200 to 2400
URUGUAY			
Montevideo	CXA6	9.620	1530 to 2100
Montevideo	CXA19	11.830	0600 to 2200
Montevideo	CXA10	11.900	1830 to 2115
VATICAN CITY			
HVJ		5.970	0900 to 0930; 1000 to 1100; 1300 to 1330
HVJ		9.680	1200 to 1330
HVJ		11.730	0015 to 0025; 0830 to 0900; 1100 to 1145
HVJ		15.120	0830 to 0930; 1100 to 1145
HVJ		17.440	0715 to 9845
VENEZUELA			
Barquismeto	YV3RS	3.490	1630 to 2130
Barquismeto	YV6RC	3.510	1800 to 2130
Barquismeto	YV3RN	4.990	0630 to 2230
Ciudad Bolivar	YV6RD	6.200	1700 to 2315
Caracas	YV5RY	3.380	0930 to 2230
Caracas	YV5RW	3.400	0530 to 2230
Caracas	YV5RX	3.500	0930 to 1100; 1530 to 2230
Caracas	YV5RM	4.970	0530 to 2230
Caracas	YV5RS	3.330	0530 to 2230
Caracas		4.920	0600 to 2230
Coro	YV1RY	4.770	1600 to 2130
Maracaibo	YV1RT	3.370	1730 to 2230
Maracaibo	YV1RU	3.110	1900 to 2130
Maracaibo	YV1RV	4.750	1730 to 2130
Maracaibo	YV1RL	4.810	0530 to 2230
Maracay	YV4RK	3.390	1800 to 2230
Merida	YV2RC	3.120	1800 to 2130
Puerto Caballo	YV4RQ	3.180	1700 to 2130
Sao Christobal	YV2RN	4.830	1100 to 2130
Trujillo	YV1RD	3.310	1700 to 2130
Valencia	YV4RP	3.460	1730 to 2130
Valencia	YV4RP	4.780	1630 to 2130
Volera	YV1RZ	4.840	1630 to 2115
YUGOSLAVIA			
Belgrade		6.150	1130 to 1800
Belgrade		9.420	0000 to 0230; 0630 to 0815; 1110 to 1125
Belgrade		9.360	

U.H.F. TUBE SET

EVER since starting radio building, I have been greatly interested in t.r.f. circuits, because their tonal quality is far superior to that of any superhet. The circuit's simplicity makes it easy for even a beginner to assemble, and the small number of parts required keeps the cost well down.

My ambition has always been to build as small and compact a t.r.f. receiver as possible and still keep the circuit's qualities. The result is shown here.

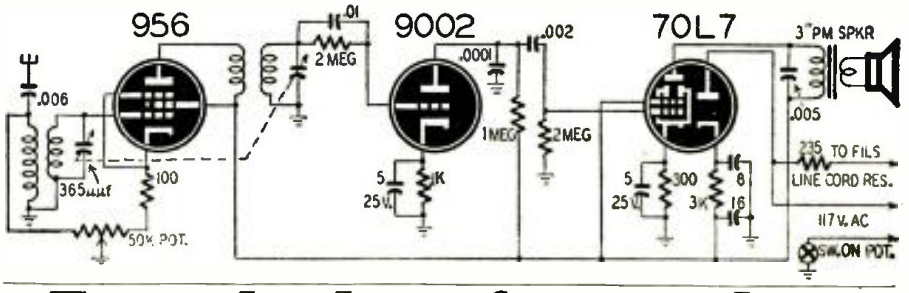
The circuit in itself is straightforward, and as the parts are not too crowded, it is really no trick to assemble it. By using a 70L7 in the output stage, a full 2 watts is had on a 3-inch speaker—more than enough volume to fill a large room.

The u.h.f. 956 acorn is used as r.f. amplifier and the 9002 as triode detector. By using variable iron-core transformers, the gain per stage is increased.

The 8 x 4 x 4-inch cabinet is painted white; the knobs, the handle, and the speaker cloth are red.

Performance of this set is so good that I practically have to chain it in place for fear visitors will take it away with them.

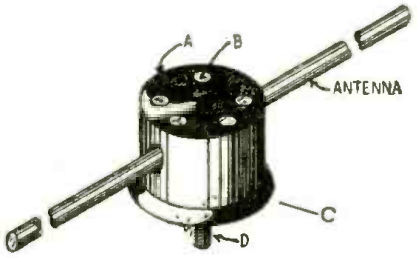
—Gerard Boult



TRY THIS ONE

ANTENNA MOUNTING

While looking for an insulator for a 4-section automobile antenna, I found that the base of the antenna was too large to pass through the insulators that I had in stock.



A—bakelite tube base; B—solder in prong holes; C—rubber grommet; D—center screw.

I took the base of a 4-prong tube, cut the prongs as close to the base as possible, and sealed the holes with solder for a smooth finish. I then drilled a ¼-inch hole through the base. The rubber grommets from available insulators fit perfectly over the open end of the base to complete a satisfactory installation.

EMERSON PAYNE,
Quebec, Canada

HANDY "V" TOOL

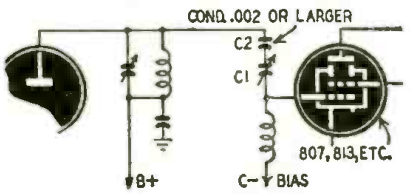
By putting two V notches (V for versatility) in a small screw driver, as shown, its usefulness to the serviceman can be increased. The notch in the side is used as a hook to pull dial cable through openings and the notch in the end may be used to push cable through holes in the dial assembly.

This end is also used to twist wire in the form of eyelets or loops for connecting wire to screw terminals. It is also very handy when installing resistors and condensers, since the leads can be formed and held in position to facilitate the soldering operation.

JOSEPH S. NAPORA,
Dayton, Ohio

PROTECTING BEAM TUBES

In many transmitters, variable-capacity coupling between the oscillator or driver stage and the control grids of 807's, 813's, and other beam tubes controls the grid excitation. Unless the condenser is widely spaced it may arc or short and place a positive voltage on



the driven grid. This may ruin the driven tube and burn out r.f. chokes.

To prevent such accidents, connect a 0.002μf, or larger, mica condenser C2 in series with the variable condenser C1. The blocking condenser should have a working voltage of about 2,000 volts for low-power rigs. A small 100μμf trimmer can be used safely for C1.

SAMUEL H. BEVERAGE,
Melrose, Mass.

NOVEL PANEL FINISH

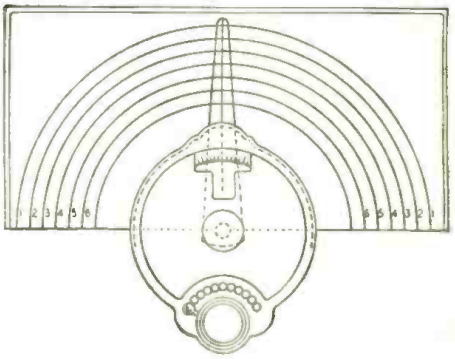
A novel crackle finish for radio panels and cabinets may be made by painting the surface with a slow-drying enamel. When the surface is partially dry, apply a coat of quick-drying enamel. The top coat of enamel will tend to shrink and wrinkle the more pliable bottom coat and produce a realistic wrinkle-finish surface.

RONALD G. BERLYN,
Ballerat, Australia

EXPANDED DIAL

National Type B vernier dials can be altered to provide increased scale length and higher reset accuracy.

The inside rim of the dial housing is cut down about ⅛ inch as indicated by the dotted lines on the drawing. A pointer is cut from thin clear plastic or celluloid and the edges finished with very fine sandpaper or pumice and water. A fine line is scribed down the center of the underside and filled with black India ink. A hole, large enough to clear the condenser shaft, is drilled in the lower end and the pointer glued to the underside of the dial.



The outer dial scale may be made by drawing concentric semicircles on white Bristol board and covering with a thin sheet of celluloid.

AUGUSTINE MAYER,
Tiffin, Ohio

REMOVING RIVETS

While attempting to drill out some rivets that were used to hold an audio transformer to a chassis, the rivet began to turn with the drill, making further drilling impossible. Other parts located on the chassis prevented the use of a chisel. I flooded solder between the rivet and chassis. This held it tight while the job was completed with a drill.

GEORGE F. CUTRESS,
Welland, Ont.

Nationally FAMOUS GIFTS FREE

OLSON

MARLIN Rifle .22 Caliber

REMINGTON Electric Shaver

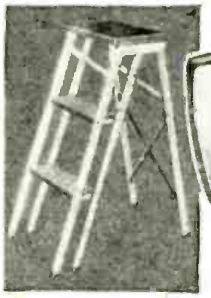
EASTMAN KODAKS

MASTER METAL Radio Tool Box

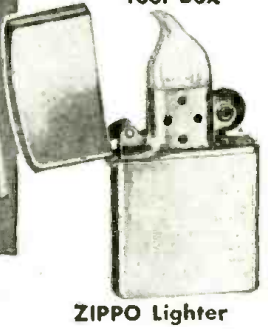
DORMEYER Electric Mixer

HEALTH-O-METER Scale

DELTA SPOTLIGHT

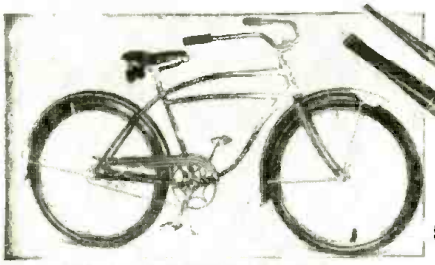


STEP-STOOL Folding Ladder



ZIPPO Lighter

ELGIN Compact



ARNOLD SCHWINN BICYCLE

PIPES by KAYWOODIE

RADIO-STEEL Scooter

NOBODY BUT OLSON GIVES YOU A BREAK LIKE THIS!

CALLING all radio servicemen and amateurs! You get an honest-to-gosh share in the profits when you buy your Radio Parts from OLSON. We sell the dependable quality Parts you need, and we GIVE AWAY these fine premiums and many others listed in our new Catalog. No strings to these gifts; you don't pay a cent of cash for them. You get them with OLSON POINTERS. Read on!

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WESTINGHOUSE Electric Roaster

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OLSON RADIO WAREHOUSE, Inc.

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ADDRESS _____

CITY _____ STATE _____

PARASITIC OSCILLATIONS

(Continued from page 37)

and plate leads should be redressed, noting the effect of each change.

It may seem that high-frequency parasitics are difficult to eliminate because the resonating elements are not even known, but they often can be tuned out or damped by using very small trap circuits. Three types of high-frequency parasitic suppressors are shown in Fig. 3, as T1 and T2. A small coil (about 10 turns on a ¼-inch form) tuned by a mica trimmer (10-100 μμf) often eliminates the undesired oscillation, with little effect on the desired frequency. A resistor of about 10 ohms may be substituted for the condenser as shown in the grid lead. These tuned or untuned circuits are connected in series with plate or grid leads as required. Another effective method is the use of 50-ohm resistors in the screen lead. In this case, the screen by-pass condenser C should be connected at the junction of the suppressor-resistor and the normal screen-dripping resistors.

Many amateurs have difficulty in operating beam power tubes such as the 813 and 4-125A without neutralization, although manufacturer's specifications often state that it can be done. The high-power sensitivity of these tubes makes operation critical under some conditions. Whenever they are operated straight through, the input and output circuits *must* be well isolated. One of the best ways to do this is to mount the grid-tuning elements below the chassis and the plate elements above it. It is also helpful to submount the socket for these tubes so that the internal shield is level with the chassis.

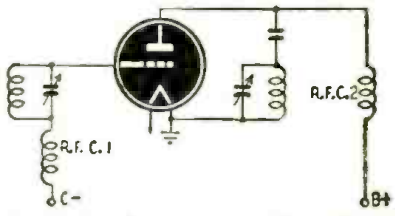

Fig. 2—Low-frequency parasitic oscillator.

When using receiving-type and low-power beam tubes, metal shields should be used between the grid and plate coils and the tubes enclosed in suitable shields.

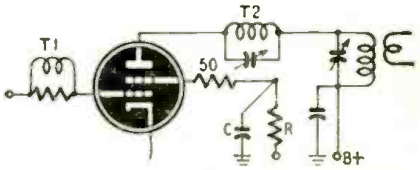

Fig. 3—Three types of parasitic suppressors.

In all transmitting circuits, the grid, plate, filament, and screen grid by-pass condensers should be connected to the socket terminals and returned to a common ground point through the *shortest possible leads.—I. Queen, W2OUX.*

SIGNAL GENERATOR

THIS handy little signal generator covers from 70 to 1500 kc in three ranges, using fundamental frequencies. A large portion of the short-wave bands may be covered by using the harmonics of the broadcast band.

The circuit (Fig. 1) uses a 1G4-G in a series-fed Hartley circuit. Power is supplied by a 22½-volt B-battery and a 1½-volt flashlight cell. Modulated r.f. is available at the output terminal by adjusting a variable grid leak on the front panel. When the resistance is increased, a grid-blocking action is set up, modulating the r.f. at frequencies from 500 to 1,000 cycles, depending upon the resistance in the circuit.

A dial, with 3 unmarked scales, is required. This may be one of several available on the market or may be made by the experimenter by cementing a celluloid pointer to a knob and using a sheet of white bristol board for the dial.

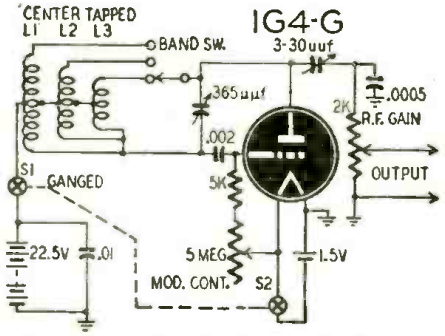

Fig. 1—Portable signal generator circuit.

The 3 coils are "jumble-wound" on a ½-inch dowel 4½ inches long, as shown in Fig. 2. L1 covers band A and tunes from 75 to 220 kc. It consists of 1,100 turns of No. 34 wire. L2, 200 to 500 kc. has 450 turns of No. 32 wire. L3 covers the broadcast band (band C) from 500 to 1500 kc. This coil has 175 turns of No. 26 wire. All coils are center tapped. The coils are mounted on the chassis and are provided with a shield that fastens to the chassis. The wire used was double-silk-covered. Single-cotton-covered wire may be used, and enamelled wire two sizes larger will also be satisfactory.

Band switching from the front panel is provided by a 3-position rotary switch. Plate and filament voltages are controlled by a double-pole single-throw switch. A 2,000-ohm potentiometer gives variable control of the r.f. output.

An accurately calibrated broadcast receiver, with tuning indicator, is used to calibrate the oscillator. The receiver and oscillator are allowed to preheat for about ½ hour. The r.f. output is adjusted for maximum and a.f. modulation is applied. The output of the signal generator is loosely coupled to the an-

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A SMALL RECORDING STUDIO

(Continued from page 36)

follow the groove. Too deep a groove can cause distortion as it may place too severe a load on the recording head.

Trying to record at the same level at 33 1/3 r.p.m. as at 78 r.p.m. will cause distortion, particularly near the center of the disc.

An asymmetrical cutting or playback needle will cause serious distortion. The playback pickup cannot track such a groove accurately. A recording needle which is not symmetrical should be resharpened. The recording assembly should be checked to ascertain what caused it to wear unevenly.

A recording stylus which is not perpendicular to the plane of the record, that is tilted sidewise, will cause an asymmetrical groove which is impossible for a pickup to track. The playback arm itself must be mounted so that its needle is held perpendicular to the record and the arm bearings do not bind.

5. *Variations in chip.* These variations may be caused by difference in hardness of the blank coating. A dry, hard blank will produce a chip which feels dry and looks dull. The chip should lie straight and be shiny. If it breaks into small pieces, it is an indication of too light a cut, a dull needle, or improper needle angle. The depth of cut has a profound effect on the chip. Experience is the best teacher of the appearance of ideal chip, and measurement with a micrometer is considered good practice.

6. *Intermittent noise.* If you are recording from a radio receiver in the winter or in a dry place, it is essential to ground all the metal parts of the recorder, as a static charge is built up on the disc, which causes a loud crackling sound in the radio. There are several liquid products which when lightly brushed on the disc surface prevent accumulation of a static charge. Of course, loose connections and microphonic tubes also can cause intermittent noises, which under some conditions might be mistaken for those caused by atmospheric electricity.

7. *Wow.* Wow is the result of a variation of speed within one revolution of the turntable. It can he caused by intermittent slippage of the drive system, by glazed rubber drive wheels or worn drive surfaces, uneven loads on the drive motor caused by binding in the lead screw mechanism. If the record does not lie flat, the depth of cut will vary during one revolution, which might vary the turntable speed. Some recorder's depth of cut and cutting angle are dependent upon the thickness of the blank. When different thickness blanks are used, the recorder should be properly adjusted. It is wise to check the turntable and then the lead screw with a spirit level. They should be absolutely parallel.

There are also many more obvious causes of poor recordings, but a generous application of common sense is usually adequate to cope with them.

Notes on Servicing

By HARRY A. NICKERSON

IN repairing a radio set, first try to learn the symptoms: Does it work at all; faintly? Does it hum badly? Is it intermittent; noisy? Does it whistle? Did the condition come on gradually? Etc. Ask the owner, but be careful not to take his replies too literally.

Test the tubes for emission and shorts or leakage. If possible, test the set *in operation* with tubes known to be OK.

While a.c. sets may have a bleeder resistance or a speaker field across B- and high B-plus, it is well to test the resistance between B-minus and high B-plus. If it is less than a few megohms, learn the cause before operating the set with a good rectifier tube in the socket. Note that some a.c.-d.c. sets (in compliance with Underwriters' regulations) have a small fixed condenser between B-minus and chassis, so chassis is not B-minus. The cathode of a cathode type rectifier, such as 25Z5, and the filament or tap on filament winding of the 80 type rectifier will be high B-plus.

Be sure that all tubes are in the correct sockets. The 12SA7, 12SK7, 50L6, and 35Z5 have filaments at 2 and 7 prongs; but 12SQ7 has filaments 7 and 8. In small portables it may be worth while to check the resistance of each tube at the socket terminals, remembering that some may be 0.05-ampere filaments with a shunt resistor to operate with 0.15-ampere tubes in series. Resistance of a hot filament varies considerably from that of cold filament.

Of course the most used formula in radio repair is Ohm's law. Put this in a circle, as shown on this page. Cover over any letter and you have the value of that letter. Thus, cover I and you have $\frac{E}{R}$.

Line cords and ballasts

A resistance line cord may be used to replace a ballast tube, and vice versa, as a general rule. If the resistance is not known, divide the voltage drop by the current in amperes passing through the string of tubes. Suppose we have 25L6, 25Z6, 6K7, 6J7 in series. The tube voltages will be 25, 25, 6.3, and 6.3, total 62.6 volts, which subtracted from line voltage of 117 leaves voltage drop of 54.4 volts. Dividing 54 by 0.3 (amp) filament current the resistance wanted will be 180 ohms.

Pilot light calculations

If we wish to insert a pilot light in such a set, we connect it in parallel with a portion of the 180-ohm resistance, or in practice, use a lower-resistance line cord, such as 160 ohms, and insert a 20-ohm resistor in the circuit, for a 6-8 volt lamp, 0.25 ampere filament. Again using Ohm's law, we find the value of the resistor by dividing the voltage of the dial light, 6, by 0.25, with the answer 24 ohms. The higher value we make the resistor, the brighter will burn the light, because with the two in shunt, more current will pass through the lamp when the resistance of the resistor is increased. (See Fig. 1.)

Calculating resistor wattage

The resistor in the above example must be able to carry the current that passes through it without overheating. So we calculate the required wattage, still using Ohm's law:

$W = \frac{E^2}{R}$ or watts $= \frac{6v^2}{20\omega}$; or about 2 watts.

In practice the wattage is figured at 2 or more times the calculated value, so we would use a 5-watt resistor.

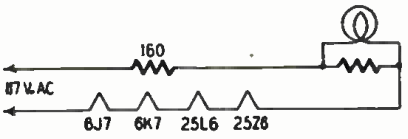

Fig. 1—The pilot lamp and its resistor.

Set manufacturers got together and decided on a color code for wires, resistors, condensers, etc. In most modern sets, you can look at a wire and say, "this is high B-plus," "this is filament," etc.

Filament is yellow; B-plus, red; grid, green; black, negative; plate, blue. On power transformers, the primary is black, and high-voltage secondary is red. The primary will have a low resistance of say 5 ohms, while the high-voltage secondary may have total of as much as 200 ohms resistance. Other low-voltage secondary windings will have very low resistance. Secondary winding carrying the most amperage will be heavier wire.

Modern color codes

If you have not memorized the color code, you can perhaps use this help: Starting with zero, the initials of the colors are BBROY for the first five: black 0, brown 1, red 2, orange 3, yellow 4. A 1-megohm (1,000,000-ohm) resistor has green as the "third significant figure." Thus, the central spot or third circular line figured will be green. Also, beginning with red, you will notice that the colors red, orange, yellow, green, blue, violet are those in the so-called color spectrum.

While many service men are still getting along with a voltohmmeter, a pair of pliers, soldering iron, and screw driver, particularly the latter, together with a lot of experience, it is probably impossible to repair all sets without considerable equipment, which includes a complete set of service manuals or

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COLOR TELEVISION

(Continued from page 25)

nel width of 32.5 mc, which is impractical under present frequency allocations and limitations of receiver design.

Network transmission of sequential color images were successfully carried out by CBS engineers. One test was made over the A. T. & T. Co. co-axial cable loop between New York City and Washington, D. C., and return (total length 453 miles). Tests were made at 40 and finally 48 color frames per second. The video frequency cutoff in this instance was 2.9 mc. A check by one of the visiting engineers who saw the demonstration revealed that there was a considerable loss in resolution due to the cutoff at 2.9 mc, which was to be expected.

A later test was conducted with the sequential system over the microwave radio link operated by the Bell Telephone Laboratories between New York City and Murray Hill, N. J.

The simultaneous color system developed and demonstrated by RCA has the advantage that all three basic colors are continuously transmitted. Some experts claim that this is very essential, when it comes to the pickup and transmission of fast-moving objects such as a football, tennis ball, etc., during the course of a game. Color breakup is liable to occur in such cases, with the result that an object moving across the screen rapidly may be seen in several different colors, owing to the fact that the necessary color is not transmitted at certain critical instants with the sequential system.

The first theater-size color television

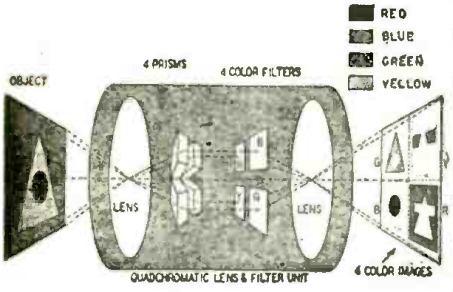

Fig. 4—The Sleeper color analyzer unit.

was recently demonstrated by RCA in Philadelphia, using the simultaneous system. The image was 7½ by 10 feet. The number of lines used was 525. Many who witnessed the demonstration said that the color television images compared favorably with regular technicolor movies. As 525 lines compares with 16-mm home movies, about 1,000 lines are ordinarily required to give the same resolution as 35-mm movies. As explained previously, the addition of color gives an apparent definition considerably above the 525 lines used.

With the simultaneous system of color transmission black-and-white receivers can pick up monochrome images by the simple addition of a converter. (Present B & W receivers cannot pick up the high frequencies allotted to color television.) This is not possible with mechanical sequential systems. Another feature is that electronic color television receivers can pick up black-and-white programs. A broadcaster with the RCA electronic system can operate a B&W and also a color transmitter, using the signals from the color camera to operate both transmitters.

Other features of the electronic simultaneous system are: freedom from flicker; greater picture brightness; no color breakup; less band width than that required for the sequential system; greater flexibility for network operation; compatibility with present commercial television, to the extent of interchangeability and consequent avoidance of obsolescence of one by the other.

AN ELECTRONIC PHOTOMETER

(Continued from page 27)

the negative potential generated by the phototube. The instrument is adjusted for an arbitrary balance at zero illumination after the preliminary warming up. It is then ready to read directly on the voltmeter circuit accurate relative values of light falling on the phototube. Referring to Fig. 4, the operation of the instrument is:

Light falling on the phototube surface releases electrons which in passing through the high resistance R1 generate a high negative potential on the grid of V1, upsetting the balance of the circuit and causing the indicating meter to drop to zero. If now a bucking positive potential is applied to resistor R1 from potentiometer R2 the circuit balance will be restored when this potential equals that generated by the electron flow from the phototube.

Cathode coupling is employed between V1 and V2, as the current flow through V1 is so small—measured only in microamperes—that few experimenters have an instrument sensitive enough to detect the flow with any degree of accuracy. This small current, however, can generate an appreciable voltage across cathode resistor R3. This voltage is communicated to the grid of V2. The latter tube, being of reasonably high mutual conductance, and operating with reasonable impedance in its grid circuit, is capable of passing a plate current well within the capability of a low-range milliammeter. The author employed a meter with a full scale of 1.5 ma. It gave a convenient balancing point at the 1-ma mark, and provided a sensitive voltmeter when operated with suitable series resistors.

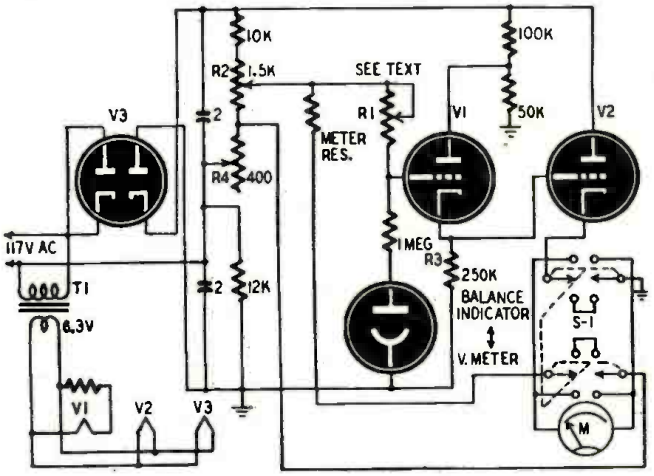

Fig. 4—This 4-tube meter is more reliable than the optical type.

The tubes used were Australian types, but the meter is uncritical as to tubes, and any medium-mu triode may be used. It would be convenient to use a double-triode type (one of those with separate cathodes).

The mechanical construction may take any form suited to the components available. Mount the phototube in a shielded head. If suitable light-gauge metal tubing is not available, a coating of tinfoil pasted inside a bakelite tube will provide combined electrostatic and mechanical shielding for the tube. Provide a small cover for the phototube window so that it may be balanced under dark conditions against any leakage or thermal currents which form a threshold conductance.

To eliminate series resistance and the use of tubes with high-voltage heaters, the heaters should be operated from a small transformer with an output voltage of 6.3 volts. The approximate 150 volts positive and negative may be obtained from any full-wave twin-heater rectifier as used in voltage-doubling circuits, or two old receiver tubes may be operated—as in the author's equipment—to give the same result. Always operate V1 with as low a heater power as possible, preferably about 4 volts. It functions with a very high grid resistance, and any appreciable heating of the grid may produce sufficient emission to make the circuit unstable. As the function of this tube is to pass only a few microamperes of current, this low voltage will be found ample.

Having connected the circuit as indicated, cut off all light falling on the phototube, and adjust the balancing resistor R4, keeping R2 at zero position, until V1 and R3 balance to produce just enough negative potential on the grid of V2 to allow V2 to pass 1 milliampere (or any value suitable for a balance of the circuit). Note this current, and preferably mark it on the instrument scale for all future resettings. Now if light is allowed to fall on the phototube the negative potential produced on the grid of V1 will reduce the voltage across R3 and apply a negative voltage to the grid of V2, dropping its plate current to zero. Adjustment of R2 now will produce an opposing voltage on R1 which will restore circuit balance as indicated by the meter in the cathode of V2 returning to its balanced condition. Now throw the change-over switch and read the value of the voltage tapped off from potentiometer R2.

Once having calibrated the phototube photometer against a photograph exposure meter at some convenient illumination level it now will be possible to read proportionately the light value falling on the enlarger easel.

Selection of the phototube is of importance. For photographic purposes a phototube sensitive to blue light should be employed. The author's phototube is

a type 929, which in addition to having a good blue sensitivity has a light sensitivity well above that of the 926, 922, or wide-band and red-sensitive tubes. This tube has a sensitivity of approximately 45 microamperes per lumen, which with its light window area of 0.6 square inch or 1/240 square foot corresponds to 45/240 microampere per candlepower or approximately 0.2 microampere. From this it is possible to estimate with a considerable degree of accuracy the voltage which can be generated on the grid of V1 if the load resistance is known. With a value of 20 megohms of well-insulated resistors, the theoretical voltage generated at the grid of V1 will be 4 volts per candlepower, and as it is possible to read the voltmeter accurately to as low as 0.1 volt, light as low as 1/50 candlepower may be read with ease. By fitting a small rotary resistor with taps to the *low impedance* end of R1, it is possible to vary the sensitivity of the instrument for reading of high-candlepower light. If the positive balancing potential from R2 is limited to 10 volts, range 1 with 20 megohms will cover 0.02 to 1 candlepower. Range 2 with a load resistance of 2 megohms covers 0.2 to 10 candlepower and range 3 with R1 reduced to 200,000 ohms covers from 2 to 100 candlepower.

The mechanical layout may well be left to individual tastes. The author has found that a very convenient arrangement is to mount the indicating meter potentiometer resistor R2 and zero-setting resistor R4 plus change-over switch S-1 in a small box to be mounted near the enlarging easel. This leaves the location of the rectifiers and indicating tubes merely a matter for personal convenience. Connect the meter panel to the rectifier and tube section by a flexible cable. Screening is not necessary for this cable, as high-impedance circuits do not pass through it. The cable to the phototube, however, should be well screened, as this portion is very sensitive to electrostatic fields, the consequences of which are highly noticeable when attempting to operate with an unscreened cable.

Mount the phototube with a small handle similar to the construction described for the elementary oil-spot instrument. It then can be moved across the field of the enlarger easel for rapid analysis of the projected image.

This instrument may be used for other than photographic purposes. If a relay is connected in the plate circuit of V2 it is possible to make an equipment for switching on lights at predetermined light levels. The application will be evident to experimentally-minded photographers.

CODE OSCILLATOR

(Continued from page 42)

high side of the volume control and connected to another tip jack. When the key is plugged into the jacks and depressed, part of the audio voltage is fed back through the volume control, causing a "howl." This howl or audio note can be varied in pitch by adjusting the volume control which, in this case, acts as a tone control. This makes an efficient and handy code practice oscillator.

RALPH BLOOM,
Brooklyn, N. Y.

STEREOPHONIC SOUND

A high-fidelity stereophonic sound system developed in Germany during the war creates the illusion of natural auditory depth, a recent Department of Commerce report states. The system uses three communication channels for reproducing sound as well as for recording it. The recorder handles a frequency range of 23 to 10,000 cycles and a dynamic range of 60 decibels with a harmonic distortion of less than 3 percent. Film noise is 70 decibels below the greatest amplitude.

DYNAMIC PICKUP

(Continued from page 30)

arm (measured from needle point to pivot) gives an error of two degrees maximum with two perfect points of tracking.

The pickup is now ready for use, but a coupling transformer is necessary as the impedance of the coil is not more than a few ohms. A microphone transformer of a ratio of 50-1 to 150-1 is excellent for this purpose. For preliminary testing, an ordinary output transformer will do very well, but for best results in service, a transformer designed for this type of work should be used. Watch for hum with a high-ratio transformer. Shielding may be needed.

The grid of the first tube should be fed through a condenser of 0.01 μf.

Compensation circuits

One further point should be mentioned. This pickup is truthful. Records are not. They have a bass cut and often a treble boost, running each way from a crossover point, usually in the octave above middle C on the piano. This method of recording is necessary for technical reasons which need not be discussed now. But it follows that a tone correction, or equalization, network is necessary. The circuit published in RADIO-CRAFT, November, 1945, in an article by George Bertsche is excellent. Using full, or nearly full, bass boost and 60 to 75 percent top cut, the results are as near to perfection as may be found.

In conclusion, here are a few practical notes. Do not use too strong a magnet, or its attraction to the turntable will increase the apparent weight of the pickup. If this effect is troublesome, pad the turntable top with composition board or, better if you can get it, ¼-inch sponge rubber. Best practice is to mount the pickup floating on sponge rubber.

Finally, this type of pickup is not critical, and the design given may be adapted or changed in any way desired. The only essential thing is a light coil of few turns vibrating across a magnetic field. A pickup using this principle, but using rubber damping (which in this writer's view is undesirable) can be made by cutting a celluloid form barrel-shaped from an old toothbrush handle, making it ⅜-inch long, ⅛-inch in the middle, and 1/16-inch at the ends. A groove can be nicked around its length, 20 turns of wire set in, and the needle jammed in a tapered hole in the middle. The coil is mounted by slipping small pieces of rubber tube over the ends and clipping these in the pole pieces. Results are good, too, and what could be simpler? The device outlined, however, while taking more time and care to work out, will be more durable and give trouble-free reproduction indefinitely.

Those who are addicted to that horrible sound known as "a lovely mellow tone" will not be interested in this pickup. But any who want accuracy and brilliance; who seek the nearest thing to perfect reproduction may find it worth constructing.

to tune this set. The capacity is varied by pressing one metal plate against another, a thin sheet of insulation separating the two. The closer the plates, the higher the capacity and vice versa.

The receiver is housed in an attractive plastic cabinet; the aerial, ground, and headphone terminals are mounted on the back. When the tuning knob is turned, the indicator moves across the scale observed in the window in the front panel. This sets needs a long aerial (100 to 150 feet) and a good ground connection, preferably to a water pipe. Signal strength depends largely on the grade of headphones used.

The tuning condenser varies the capacity across the secondary coil of the aerial coupler. The loosely coupled aerial coil helps to improve the selectivity. No batteries are required and no detector adjustment is necessary. Reception thus becomes a very simple matter. The crystal detector is a 1N36 type. A small fixed condenser is connected across the headphones.

A wristband radio

The Da-Myco "Dick Tracy" wristband receiver illustrated in Photo C is one of the latest and smallest *tunable* sets to make its appearance. The photo shows the compact manner in which the designers have built this receiver, which even has a tuning inductor stowed away inside the phone shell. The headphone has a steel magnet and a single coil of about 1,000-ohm resistance, mounted on a soft iron core at its center.

The tuning coil has an iron slug in-

THE CRYSTAL RADIO MAKES A COME-BACK

(Continued from page 33)

side it to boost its inductance. A slider moves across the turns, over a path cleared of insulation in one of the simplest and oldest of tuning circuits. The tuning lever moves in a slot cut through the side of the case and terminates in an

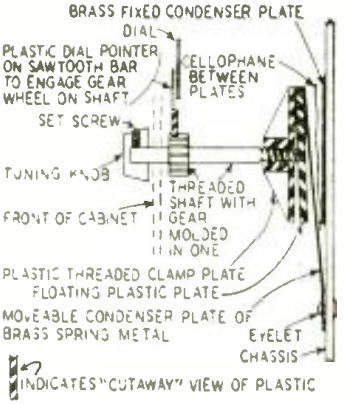

Fig. 3—Detail of Revell radio condenser.

insulated knob. The fixed crystal detector is of the radar type (germanium) and requires no adjustment. A fixed condenser is shunted across the headphone.

Two jacks are provided in the side of the case to which aerial and ground connections are made. As with all other crystal sets, the longer and higher the aerial, the better the reception. For apartments and other locations where it is not convenient to erect an antenna, a compact aerial device is available which plugs into any convenient outlet. This device consists of a condenser and a suitable inductor which bridges the set across the lighting circuit; no current passes through the set.

References:

Diode vs Crystal Detector, RADIO-CRAFT, Jan., 1946.

Loud Crystal Reception, J. D. Amorose, RADIO-CRAFT, Sept., 1945.

Interflex (Short-Wave) Crystal and 1 tube set, RADIO-CRAFT, Oct., 1943.

Crystal Signal Tracer, RADIO-CRAFT, May, 1944.

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Link-Coupled Crystal Set (with superior selective tuning), Carroll Utermahlen, RADIO-CRAFT, Sept., 1946.

A Modern Crystal Set, W. J. E. Spain, RADIO-CRAFT, Dec., 1946.

Germanium Crystal Probe, R. E. Altomare, RADIO-CRAFT, Sept., 1946.

V.F.O. EXCITER OR TRANSMITTER

(Continued from page 32)

previously discussed. To calibrate the other half, switch in C3 and set the variable tuning condenser at maximum capacitance. Then vary C3 until 3.5 mc just comes in. In this particular case it occurred when C3 was a parallel combination of two ceramic condensers, a 100- and a 20-μμf unit. Otherwise it would have been convenient to use a 100-μμf fixed and a midget variable of about 35 μμf, adjusting the latter as required. The calibration was made to coincide with the band and it is impossible to operate out-of-limits unintentionally.

The plate circuit of the e.c.o. is tuned to the second harmonic, 80 meters. This provides good isolation and permits a higher stability to be attained. The coil is wound with 42 turns of No. 22 wire on a ¾-inch polystyrene form screwed to the chassis. A 100-μμf trimmer condenser sets the band and tuning is done with a 50-μμf midget variable. Actually this is a non-critical control unless frequency is changed over a wide range.

Voltage to the 6SJ7 is regulated by a VR-150 and a VR-90 in series. The large screen resistor drops the voltage to about 100. This tube does not become too hot to touch even after operating a long time.

Oscillator output is capacitance-coupled to the buffer 6F6. This tube runs Class A for good isolation. An r.f. choke is used in the plate circuit in series with the dropping resistor. This runs quite hot but is well within its ratings. The plate voltage is about 300 and the screen runs at about 150. If more output is desired this tube may be run at higher ratings.

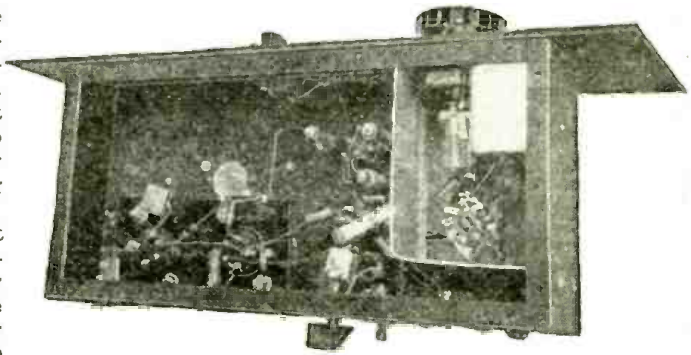
The 2-tube exciter may be either capacitance- or link-coupled to a multiband exciter. If the leads are long it is better to use link coupling, using an untuned transformer instead of the r.f. choke in the 6F6 plate circuit.

After experimenting with various push-pull, push-push, and single-ended final stages, the latter was finally adopted, but two tubes in parallel are used for more output. Such a circuit can feed a single wire effectively and is easier to excite than push-pull. The latter does, of course, give slightly higher output for a given input. Type 807 tubes are used in the final. These are now available at less cost than most receiving tubes. They require little excitation, give ample output, and are well shielded. With 400 volts on the plates, output is about 50 watts.

High-frequency parasitics were very much in evidence at the start, as is to be expected from sensitive beam power tubes. They were completely eliminated by connecting 50-ohm resistors in each screen lead and small chokes in each plate lead. The chokes are made of 15 turns of No. 28 wire wound on (and in parallel with) 10-ohm resistors. Shielding between grid and plate circuits is ample, due to sinking the tubes below the chassis level. The tubes were found to run cooler without shields.

A pi network couples the final to the antenna and gives very satisfactory results. This system permits continuous variation of power input to the final by adjusting the coupling. The tank coils are wound to resonate in the desired band when both coupling and tuning condensers are at half capacitance. This is done because an increase in capacitance of one condenser must be accompanied by a decrease in the other to maintain resonance.

The coils L for the pi network are space-wound with No. 22 wire on plug-in forms 1½-inch in diameter. There are 38, 15, and 6 turns on the 80-, 40-, and 20-meter coils respectively. Length of winding is 2 inches on the 80, 1½ on the 40 and 1 inch on the 20-meter coil.



L2 and its air and mica condenser are well seen at bottom center.

Choice of either oscillator or final keying is provided. Many amateurs prefer the latter in order that there will be no interruption of the oscillations. Oscillator keying permits break-in operation; however it is necessary to add a C-battery or cathode bias resistor to prevent overloading the 807's.

Calibration and operation

After the rig has been completed and tried out is a good time to calibrate the main dial. A 100-kc crystal oscillator (preferably checked with WWV) is good for this purpose. To obtain 100-kc markers through the 80-meter band, crystal harmonics are caused to beat against the v.f.o. signal while listening in on a shortwave receiver. If the latter is tuned to 80 meters, zero beats will be heard at 3.5, 3.6, 3.7 . . . kc and these points can be marked on the v.f.o. dial. To obtain 50-kc markers, the receiver is tuned to 40 meters. Here again zero beats are heard every 100 kc, that is, 7.0, 7.1, 7.2 . . . , but the first corresponds to 3.5 on the v.f.o. dial and the second is a harmonic of 3.55. In other

words, the calibration now may be made every 50 kc on the 80-meter v.f.o. dial. Tuning to 20 meters will provide a marker every 25 kc, and so on. Since the scale is practically linear, it is hardly necessary to go so far.

To put the rig on the air on 80, the main dial is set to the desired frequency with the switch under the dial thrown to the proper range. The buffer condenser is used as a trimmer and is adjusted for minimum reading on the panel meter. This indicates maximum excitation to the final grids. The antenna-coupling condenser is then set to

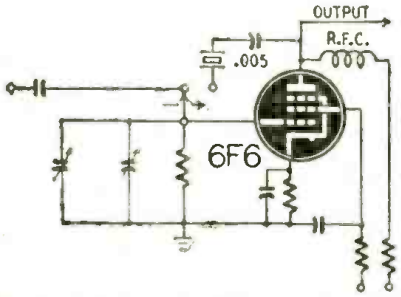


Fig. 2—Switching circuit for crystal control.

the desired loading and the tuning condenser is resonated for lowest meter indication. If this reading represents an input lower than that desired, the coupling condenser is set at a lower capacitance and the other condenser retuned.

The panel meter is calibrated directly in watts input rather than conventional plate milliamperes. A resistance-wire shunt is adjusted across the 1-ma meter until full-scale deflection corresponds to 250 milliamperes. Since the plate voltage is 400, full deflection is equal to 100 watts input. Furthermore, power is proportional to current when the voltage is kept constant, therefore the meter reads in watts. For example, if the indication is 0.47 it is equal to 47 watts. With the circuit constants shown in the schematic, maximum (off-tune) input is about 100 watts. Don't permit the transmitter to remain in this condition for long.

Most QSO's will probably be on a single frequency for both reception and transmission. If the final is keyed, this means that the local transmitter oscillator will interfere with the incoming signal. The toggle switch under the v.f.o. dial can be used to good advantage here. After transmitting on the desired frequency, throw the switch; this changes the frequency but does not interfere with oscillations. However, don't forget to throw the switch back again before transmitting!

A flexible frequency is handled more conveniently when a rig contains few controls and circuits, and in any case the signal stability is best at low harmonics of the oscillator frequency. For these reasons and to keep the rig from becoming too complicated, no doublers have been incorporated. However, the buffer circuit resembled a Pierce crystal oscillator so much that we couldn't resist trying it out that way (without affecting the v.f.o. channel in any way, of course). All that has to be added for alternative crystal control is a socket, blocking condenser, toggle switch, and, of course, a crystal (Fig. 2). This makes it possible to change to crystal control at will and easily.

The output of the single-tube oscillator is slightly greater than that of the 2-tube v.f.o. unit. There is ample excitation to the final stage so that it can operate as either a doubler or a tripler with suitable crystals. We can even get a few watts of 28-mc ouput from a 7-mc crystal!

denser are installed in the screen-grid circuit.

The 7E7 tube functions as the second detector, a.v.c., and also, when required, as the first audio amplifier in a reflex circuit. It is with the hookup and action of this tube in the circuit that we are primarily concerned.

In the conventional connecting of a diode detector in a radio circuit the high potential of the signal input transformer L4 is connected to one or both of the diode plates. This results in comparatively low gain and poor selectivity for the preceding stage of carrier amplification, because the secondary of its transformer is loaded with the audio signal. These two serious disadvantages of the diode detector are eliminated in the circuit of this set. At the same time the distinct advantages of the diode as a linear detector are retained.

A simple inductance L5 is placed in the cathode circuit of the diode tube, causing cathode voltage to vary with the r.f. signal, and detecting it on the cathode's negative half-cycles.

The value of L5 should substantially equal the carrier frequency. In other words, it should be large enough to resonate at about 456 kc when the i.f. is 456 kc. A spare i.f. winding with several turns added will work, or a 2.5-mh radio-frequency choke coil will serve the purpose. The larger the coil, the greater will be the gain and selectivity, but this must not be carried so far as to produce instability in the set.

L5 is not tuned, and should be shielded if necessary, though if carefully placed, shielding should not be required.

The secondary of L4 is connected directly to the control grid of the 7E7 tube. To prevent loading of this transformer, and thus to secure the greatest gain and increased selectivity, the control grid must be maintained at a negative voltage greater than the incoming peak signal voltage. The negative bias and a.v.c provided in this circuit are intended to do just that.

A novel and thoroughly practical feature of this set is the use of two variable controls, R5 and R12. Each should be a minimum of 1 megohm, and for better results may be 2 megohms. Each has a tone tap, preferably near the center but not less than 500,000 ohms up from the bottom.

The schematic clearly indicates the proper connection of the two controls in the circuit.

R5 in the detector circuit is used to cut the reflex feature in or out at will. When the center arm *b* is moved to a position between *c* and *a* the audio signal is reflected to the control grid (No. 6) of the 7E7 tube, thus making it the first audio amplifier. When this is done the center arm *b* of R12 is moved to a position near *a* of that resistor. Now not only volume but tone as well can be controlled. For example, if the volume is lowered by adjusting R5 and correspondingly increased by adjusting R12, the resulting tone will have a lower pitch, and much noise will be eliminated.

Suggested by R. C. Redmond, Auburn, N. Y.

"What kind of current does it use—A.C. or B.C.?"

The first audio amplifier is eliminated when the center tap *b* of R5 is moved to a position between *c* and *d*. In this case the center tap of R12 is moved to *c* for the loudest signal, and thereafter the volume is controlled by R5.

For local stations by day and most stations by day or night the first audio amplifier will not be required for reasonable volume and the best in radio reception. The additional audio amplification will be undesirable also when headphones are used.

When properly constructed and accurately tuned the selectivity of this set is high. In this connection it must be remembered that the i.f. transformers cannot be tuned with a metal screw driver. A nonmetallic driver such as can be made from a plastic toothbrush handle is a necessity.

When building this or any other radio set it is well to keep in mind certain fundamentals: Use good parts. Keep grid and plate leads as far apart as practicable; also keep these leads as short and direct as is convenient. Keep all leads close to the chassis or other grounded parts. Make sound soldered connections.

When more than 100 volts is available for the B-supply it may be necessary to install appropriate resistors and by-pass condensers in the screen-grid circuits.

Use good components and careful work in the audio end.

No amount of effort applied on the detector circuit will compensate for poor performance of the power amplifier or for a poor, mismatched or low-quality speaker.

In connection with the foregoing, and especially for beginners, attention is directed to the excellent article "Problems Underlying Construction of a 5-Tube Superhet," by H. A. Nickerson, page 24, RADIO-CRAFT for October, 1946.

A set that fails to bring in far distant stations regularly should not be condemned for that reason alone. The fault may be due to atmospheric conditions which may continue for several days at a time. So do not be in too big a hurry to tear up your set.

NOTES ON SERVICING

(Continued from page 58)

circuit diagrams of the set under repair, the voltohmmeter, a signal generator, tube tester, and the usual tools. A beginner can locate trouble in a set more quickly with a signal tracer (usually a t.r.f. set which has a probe to connect at any point on set under test and gives an audible signal or operates an "electric eye") than the experienced service man with his voltohmmeter and great experience.

Too many servicemen stop with replacement of a defective part. An effort should be made to put the set in top-notch condition by adjustment of trimmer condensers and alignment of superheterodynes, cleaning out variable condensers, the addition of baffles, lubrication of bearings, addition of shields and condensers when needed.

250-WATT FM-AM TRANSMITTER

(Continued from page 31)

age developed across the capacitor, or from the 6SJ7 grid to ground, lags the oscillator tank voltage by approximately 90 degrees. This lagging grid voltage causes the 6SJ7 plate current to lag the oscillator tank voltage by approximately 90 degrees and the tube then appears as an inductance in parallel with the oscillator tank circuit. The frequency of the oscillator will then be increased.

When the transconductance of the 6SJ7 modulator tube is varied by an a.f. signal voltage applied to its control grid from the 6SN7-GT speech amplifier, the magnitude of the inductive reactance across the 6F6-G oscillator tank is caused to vary at an audio rate. This change in inductive reactance causes the oscillator frequency to be varied at a rate which depends upon the frequency of the audio modulating voltage and by an amount which depends upon the amplitude of that voltage. If the circuit values as given here are carefully followed, the 6SJ7 reactance-tube modulator will give linear frequency modulation and produce many times as much deviation as will be required for amateur communication.

Frequency-stabilizing circuit

The 6H6, the 6SA7 and the 6C5 tubes function as a correction or stabilizer circuit to stiffen the frequency shift or deviation of the 6F6-G oscillator during the periods of frequency modulation. It will be noticed that the grid of the 6SJ7 reactance-tube modulator is returned to ground through two isolating resistors of 500,000 ohms each and the two 100,000-ohm resistors connected across the 6H6 discriminator output circuit.

As mentioned above, any change in the transconductance of the 6SJ7 will cause an effective change in the oscillator tank circuit inductance and thereby cause a corresponding change in the oscillator frequency. The *direction* of the frequency shift, i.e., to a higher or lower frequency, will depend upon the *polarity* of the voltage applied to the control grid of the 6SJ7 tube; the *amount* of the frequency shift, i.e., the number of kilocycles, will depend upon the *magnitude* of the 6SJ7 control-grid voltage.

The r.f. voltage at the grid of the TB-35 will be fed, in proper proportion, to the signal grid of the 6SA7. The 6C5 tube is connected as a Pierce crystal oscillator, operating on a frequency of 7,000 kc; its output is fed to the injection grid of the 6SA7. When operating on FM, the 6F6-G grid circuit will be tuned to a frequency of approximately 3,675 kc and the 6F6-G plate and TB-35 grid circuit will be tuned to *twice* 3,675 kc, or 7,350 kc. Now, if we mix the 7,350 kc and 7,000 kc signals we shall obtain in the plate circuit of the 6SA7, a beat frequency or "i.f." equal to the difference between 7,350 kc and 7,000 kc or 350 kc.

The 6SA7 output signal voltage is fed to the primary of a tuned discriminator transformer adjusted for resonance at 350 kc. Connected to the secondary winding of the discriminator transformer is a 6H6 diode rectifier tube. Two load resistors of 100,000 ohms each are connected in series across the cathodes of the 6H6 tube with the d.c. return from

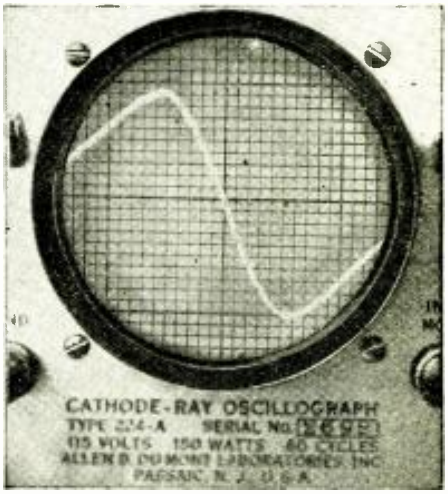


Fig. 2—Discriminator curve. Center is 350 kc; left: higher frequencies, right: lower.

the discriminator transformer secondary winding center tap connected to the resistors.

As long as the i.f. signal remains exactly 350 kc, equal signal voltages will be applied to the two diode plates, equal currents will be drawn and equal and *opposite* voltage drops will appear across the two load resistors. Under these conditions, the net d.c. voltage output from the discriminator will be zero, and consequently no change in the transconductance of the 6SJ7 will take place. Thus, the frequency of the 6F6-G oscillator will not be affected. However, under conditions of frequency modulation, the 6F6-G oscillator frequency will vary with changes in the amplitude and frequency of the modulating signal. Thus, the i.f. or beat frequency will swing higher or lower than 350 kc. As shown in Fig. 2, for frequencies higher than 350 kc, the discriminator output will produce a d.c. voltage having a *positive* polarity with respect to ground; for frequencies lower than 350 kc, a d.c. voltage having a *negative* polarity with respect to ground will be produced. The d.c. voltage output *vs.* frequency deviation will be linear up to approximately 50 kc each side of center frequency, depending upon the design of the discriminator transformer. This d.c. discriminator voltage is fed back to the control grid of the 6SJ7 reactance tube, causing its transconductance to change so as to correct the variation.

This change in transconductance is reflected into the 6F6G oscillator tank circuit as a change in inductance and the frequency of this circuit is caused to vary correspondingly.

The second half of this article, describing the oscillator and r.f. sections of the transmitter, will appear in an early issue.

TELEVISION OVER A LIGHT BEAM

(Continued from page 22)

the telecast program is handled by a second light beam comprising the modulated light source at the transmitter, and the second photomultiplier cell at the receiver, feeding into the conventional loudspeaker system. In the demonstrations the two light beams have been paralleled and directed so close together as to overlap in part at the receiving end, yet each is directed by adjustable lenses and mirrors on to the respective photocells. The video and audio reproductions are on a par with the usual electrically transmitted television systems.

The equipment demonstrated so far has established that photovision is practical. Further research promises to make the system applicable to many fields of transmission of images and sound. For example, in the transmission of color television, four color channels may be independently modulated with the red, green, blue, and sound signals for a composite simultaneous color television system. Use of filters at the receiving point allows selective separation of all these signals even though transmitted from a single sending point.

Very recently DuMont engineers have been testing photovision equipment over distances of several miles. Even with the not yet perfected equipment now available, promising results are being obtained with both video and audio signals. The remarkable thing is that the seemingly dull light spot on the transmitting cathode-ray tube can be picked up several miles away by suitable focusing means, despite bright sunshine, and translated into pictures and sound. Fog is still an obstacle, but further developments may largely circumvent such interference. Full-scale commercial equipment has been evolved in the past few weeks and will soon be demonstrated as photovision steps out of the laboratory and makes its bid for a place in everyday telecasting.

The highly directional characteristics of this communications medium, coupled with its privacy feature and reliability, makes it very probable that it will find many uses in other fields of communication, such as broadcast relaying and point-to-point telegraph or telephone.

SIMPLE PANEL MARKER

If you have one of the small rubber-stamp sets, it can be used to do a neat job of lettering on the panels of receivers, transmitters, and other apparatus. After setting up the type, attach a piece of cotton cloth to a block of wood and coat it with a light coat of white lead. Press the stamp lightly on the surface of the cloth to pick up a thin coating on the raised surfaces of the type. The stamp is then pressed lightly against the surface of the panel to transfer the lettering. Light pressure should be used to prevent smearing the letters.

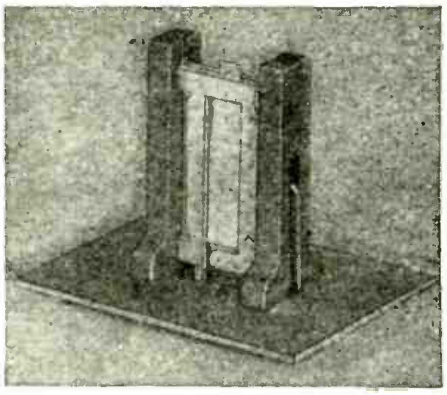
EDWIN BOHR,
Chattanooga, Tenn.

New magnetic material, *supermalloy*, developed by Bell Telephone Labs, has 200 times the permeability of silicon steel.

GUILLOTINE FOR FM

TO radiomen, the "guillotine" means a new variable-inductance tuner developed by General Electric. For use in FM reception, it is considered more efficient than condenser-coil tuning.

The tuner consists of 2 identical silver-plated brass frames which, when connected at their open ends, form a 2-turn inductance. The inductance of the 2 turns is varied by insertion of a silver-plated brass "guillotine" blade between the turns. The effect of the blade is to reduce the inductance of each turn and also the mutual inductance between the turns. The tuning curve is adjusted by cutting slots in the blade which provide an easy and permanent means of tracking the oscillator and r.f. circuits with each other. Both terminals of the tuner project through the receiver chassis, making very short leads possible and providing a rugged tie point for soldered connections.



Electrical design problems in the front end are greatly simplified by the use of the guillotine tuner because it is possible to localize each tank circuit within a small area and to keep r.f. chassis currents at a minimum. Microphonic troubles are almost completely absent. The blade of the guillotine being ungrounded, sliding contacts and pigtails are eliminated. All of these factors contribute toward an efficient electrical design providing high performance with little trouble from regeneration or alignment difficulty.

The tuner assembly is enclosed in a metal box for shielding and for mechanical protection and dust-proofing. Operation on the short-wave spread bands is comparable to communication receiver performance and ease of tuning. Short-wave microphonic howl, reports I. J. Kaar, manager of General Electric's Receiver Division at Bridgeport, Connecticut, is completely absent with the use of the guillotine tuner, making the full audio power output of the receiver usable on all bands. Loop reception is provided on all of the AM bands and power line pickup is used for local FM reception. Terminal connections are provided, of course, for AM antenna and FM dipole use wherever required.

It appears that this particular technique will be widely employed in the future for applications involving the upper frequencies, especially for FM and television.

"BRAND" VS. "ORPHAN" RADIOS

(Continued from page 17)

As long as the receiver plays it will sell to the unwary who cannot afford to pay for a standard brand. It stands to reason that such receivers will cost more in servicing and upkeep than reputable makes.

Unfortunately the public at large does not know all of this, because to most laymen a radio set is a radio set, no matter what its make, as long as they hear it play loudly in the store. Later on, when it refuses to perform and has to be serviced several times, the eventual brunt may fall on the unhappy serviceman who gets blamed for not being able to repair the set so that it will continue to work satisfactorily.

Sooner or later the public will come to understand that the set itself was no bargain. This completes the cycle where both the dealer and the serviceman get the ill will of the public. This situation is by no means new.

These same cycles have occurred repeatedly in this country and they will most likely continue for a long time under our free economy.

The radio industry itself is to be blamed in large part for this condition. **To the best of our memory, the radio industry has never made a concerted effort to educate the public that it is cheaper in the end to buy a higher-priced known brand set than to buy unknown fly-by-night makes—the so-called "orphan" radio sets—of which there are many today.**

If it is impossible—which it probably is at the present time—to reduce drastically the price of standard brands, it would seem the intelligent thing for the established radio set manufacturer to use newspaper and magazine publicity to prove to the public that it pays to buy well-known sets with a real guarantee behind them and educate the public to stay away from "distress" and "orphan" radios.

The serviceman, particularly, would be smart to refuse to service such unknown makes, which in the long run bring only grief and make no friends or money for him.

Once a serviceman tells a set owner that it is not good economics for him to invest service and repair money in a set that will only go to pieces later on, half the battle against this type of merchandise is won. The serviceman should furthermore protect himself *by stating in writing* to the set owner that such repairs should be considered only as of a temporary nature and that in all probability the particular receiver calls for endless and costly servicing. This protects the serviceman; it also demonstrates that the servicing industry is working in the public's interest rather than in its selfish own.

Such a strategy always commands the respect of the set owner because he is warned beforehand that he cannot expect orphan receivers to stand up and perform indefinitely.

Radar, already important at sea and in the air, is to be used on land as an aid to highway safety, Connecticut State Police report.

A microwave signal sent out by the radar is reflected by the moving car, Doeppler effect causing a wavelength shift which increases with the car's speed. This shift is translated into miles per hour on the instrument's meter.

NARROW-BAND FM FOR HAM RADIOS

(Continued from page 23)

The circuit of Fig. 6 can be readily switched from AM to FM. It requires a little more digging into the receiver than some of the other arrangements described. However, it makes it possible to employ the existing transformer in most communications receivers. A 3-pole, double-throw switch is required. Both ends of the secondary are switched from the AM diode detector to the crystal diodes as FM detector. The third switch connection switches the audio from the existing detector to the discriminator output. Remember that attaching the components of these circuits to your set will detune the i.f. transformer to which they are connected. Realignment of the last stage will be necessary! The circuit of Fig. 6 for a single-ended transformer can be applied as in Figs. 2, 3, and 4.

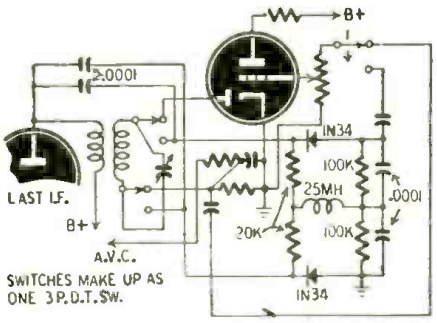


Fig. 6—Practical circuit for standard coils.

Possibly the best way to make the additions for narrow-band FM reception will be to build a separate unit as is shown in Fig. 2 and connect as in Fig. 3. In using these adapters it is also wise to remember that the wide-band position of the receiver selectivity control gives best reception.

Any one of the arrangements can have a center-frequency indicator as described above. The connection is the same as for Fig. 2. It is essentially a high-impedance d.c. voltmeter.

The discriminator can be aligned very simply. Connect the signal generator to the last i.f. stage and adjust frequency to the exact center of the intermediate frequency. With the discriminator secondary detuned somewhat, adjust the primary for maximum deflection of the center-frequency indicator. Then retune the secondary for zero (center indication of the meter). When properly tuned, signals an equal frequency difference away from the center frequency should give the same reading on opposite sides of the zero on the indicator.

A special adaption

A means of utilizing the double ended (push pull) detector of an RME 69 receiver is described in Fig. 8. A 3-pole double throw switch may be employed to switch the ends and center tap of the detector transformer. The materials required in addition to the switch are a 50-μμf mica capacitor, "E" connected from last i.f. plate (6D6) to the r.f. choke of the discriminator at "B." Also required are the crystal elements and the necessary resistors and condensers comprising the discriminator circuit. A center frequency indicator is shown in this arrangement. The audio output connection from the discriminator goes directly to the audio input because when the discriminator is unconnected to the transformer it will be inoperative.

Alignment of the RME 69 will be required after the adjustment. It is important that the switch be placed as close as possible to the detector i.f. transformer so that leads will be short. The center-frequency meter will have to be an external device. The parts indicated in the diagram by numbers as C15, R17, etc., are the designations of the RME 69 service schematic.

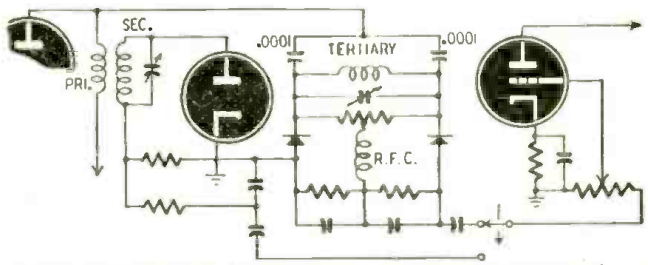


Fig. 7—If transformer has 3 windings, this circuit can be used.

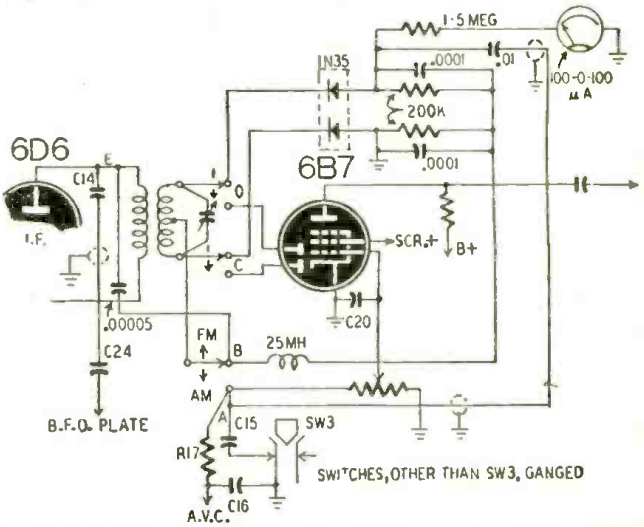


Fig. 8—How an RME 69 can be made an AM-narrow-band FM receiver.

ANTENNA PRINCIPLES

(Continued from page 39)

at A in Fig. 2 will be directive primarily in the horizontal plane of the field.

An important variation of the sectoral horn is known as the *pyramidal horn* which is flared in *two* dimensions. This horn can be designed to provide an equidirectional pattern in all planes of radiation.

A *conical horn* (B in Fig. 2) provides a similar equidirectional field pattern.

Width of the horn mouth should be at least ten times the wave length. The distance from throat to mouth should be approximately the same. The amount of flare, or flare angle, will have marked effect on the directivity pattern. But, again, exact design data is complicated because of the various modes involved.

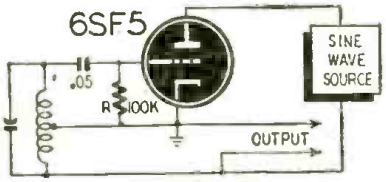
Conical horns are ordinarily used with circular wave guides. Field patterns are equally directional in both horizontal and vertical planes.

NEW RADIO ELECTRONIC PATENTS

(Continued from page 48)

every third cycle to be applied effectively.

The fundamental frequency is applied in the plate circuit of a triode (from "sine wave source" in the accompanying diagram). The triode is hooked up in a Hartley circuit tuned to the frequency of the desired sub-harmonic.



During the first positive half-cycle of source frequency the tube conducts and the circuit oscillates. A positive pulse is applied at the grid as a result, and current flows in the grid circuit. In flowing through the resistor, the current biases the grid past cut-off so that (temporarily) the fundamental frequency has no further effect. However, oscillations can continue in the resonant circuit at the sub-harmonic frequency to which it is tuned. Maximum output is obtained by adjusting R and C so that the grid charge leaks off just in time to permit every *n*th cycle of the fundamental frequency to start oscillations, where *n* is the order of the sub-harmonic desired. In the figure the source is tuned to 10,000 cycles and the Hartley circuit (and therefore the output) to 1,000 cycles. Constants given are correct to produce the desired effect.

TRANSATLANTIC NEWS

(Continued from page 40)

towers were modified to increase their height to 800 feet, and an antenna was erected consisting of 2 triatics with the apex of each supported by a cable anchored to the mountaintop.

The capacitance of the antenna proved to be about 24,000 μμf; its resistance and that of the ground total 0.5 ohm. The antenna tuning inductor has a diameter of 15 feet 9 inches. It is wound with litzendraht cable containing 6,561 strands and having an over-all diameter of 1.5 inches. At 16 kc the inductance of the coil is 5.36 millihenries and its h.f. resistance 0.14 ohm. The Q is thus 3,850!

FRENCH RADIO-MODEL AUTO

(Continued from page 21)

L3—Same as L2.

In the transmitter which sends waves of varying frequencies, condenser C is replaced by one of the condensers in the control box of Fig. 3.

The modulator furnishes an audio-frequency voltage capable of modulating the transmitter. Construction of this part is straightforward. Fig. 2 is a diagram of a simple unit for this purpose. In effect, the first tube, a 6C5, is an audio-frequency oscillator; the other tubes amplify its signal. The oscillator coil may be a Class B audio transformer with one center-tapped winding. The condenser across the tapped winding is selected to give a desirable note. The grid resistor may be varied also, for the same purpose. Any convenient power supply may be used.

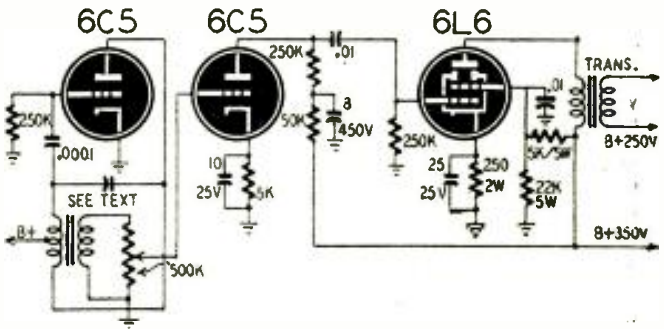


Fig. 2—Any convenient modulator circuit may be substituted for this.

The transmitting antenna may be of any type, but it is desirable that the maximum energy be radiated. An excellent solution is to use a horizontal half-wave antenna of the center-fed type. A quarter-wave vertical antenna, fed at the base, also might be used. In that case, it is necessary to add a capacitor in series with the antenna, tuning it so the maximum current is at the base, and therefore the maximum voltage is at the summit of the antenna.

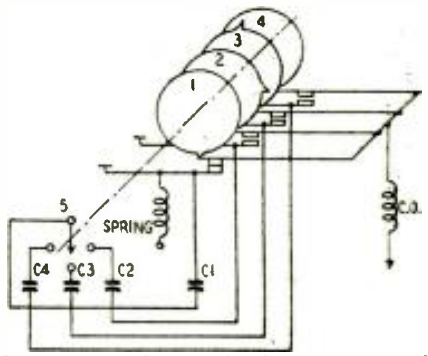


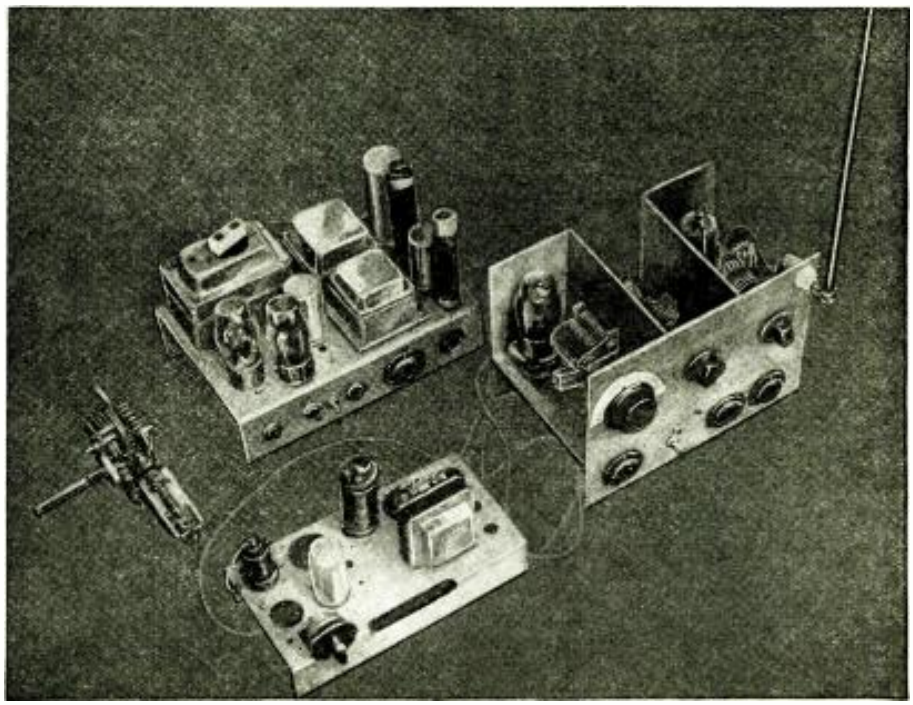
Fig. 3—Tuner for variable-frequency control.

Two possible procedures

In the first method, the control signals are produced by impulses; the modulated r.f. wave exists only during short intervals, several milliseconds, for example, the extent of the impulsion depending on the control signal. The key, or, better, automatic telephone dial apparatus, permits obtaining different durations of signal. The key or dial is then connected at X in Fig. 1.

In the second case, *control depends on* the wave length. For example, the "left" control is transmitted on a frequency F1, the "right" control on a frequency F2.

Fig. 3 shows the apparatus adopted



Fixed station includes control, modulator, power pack and the 2-tube transmitter proper.

by M. Roveyaz: four cams are fixed on a shaft connected to a motor, one for each control wave length. The motor shaft carries a selector S. Each key has a dog corresponding to one of the selector positions and placing a capacitor across the oscillating circuit CO. Whichever capacitor is switched in then will be condenser C of Fig. 1.

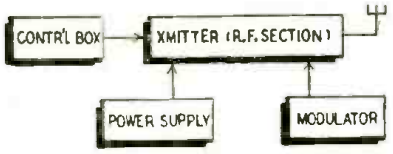


Fig. 4—Line-up of transmitting equipment.

The control apparatus includes several buttons, pressing one of which is sufficient to obtain a carrier frequency corresponding to the signal required to effect a desired maneuver.

Fig. 4 is a block diagram of the transmitter used when control is effected by changing frequency—the method with which we have experimented. However, both methods have been tried successfully by M. Roveyaz.

The receiving equipment

The automobile equipment consists first of an antenna to collect the waves emitted by the transmitter; the best solution seems to be a vertical aerial. This may be a demountable whip antenna, in the form of an aluminum or brass tube, fixed on the left side of the automobile.

The receiver preferably should be of the superregenerative type, permitting reception with a minimum of tubes. The detector may be a 955 or a 6J5.

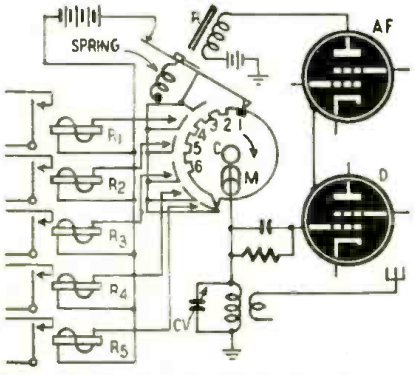


Fig. 5—Receiver, showing relay connections.

Fig. 5 gives a rough plan of the receiver used when the frequency-change control system was employed.

The cam C is coupled to the variable condenser CV by an insulating coupling, the ensemble being turned by the automobile's motor through a suitable step-down gear system.

If a carrier is transmitted, when the condenser CV in its rotation arrives at a point where its value is such as to tune its circuit to resonance to the transmitted frequency, a current is set up in the detector plate circuit. Amplified by a second tube, it causes relay R to close. This stops the cam as long as the signal continues. One of the relay circuits (R1, R2, R3, R4, R5) is closed and acts directly on its associated control.

If the frequency F1 stops, the relay opens and the cam again begins to turn with the variable condenser. A transmission on a new frequency F2 will again immobilize the cam in a new position corresponding to another "command."

Dry batteries supply power to the receiver's plates and the filaments are lit by small storage cells. In a full-sized automobile, the power could be supplied by the car's storage battery and a vibrator pack.

The mechanical controls are among the most delicate parts of the equipment. One arrangement, described in the magazine *QST* (March, 1940) includes a relay E supplied with current from a battery. Its armature A carries a hook at one end and a spring at the other. A cross which is coupled to a motor so that it turns on an axle when released . . . (see Fig. 6) has a spur on

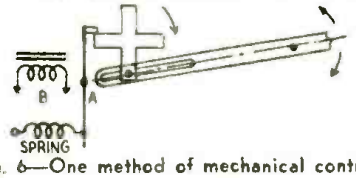


Fig. 6—One method of mechanical control.

one arm which passes through a slot in a lever which can control the steering gear or the speed governor of the model. The current passes through the electromagnet when its associated relay (R1, R2, etc., of the receiver) closes, the armature is drawn toward it, liberating the cross which moves a quarter turn, placing the lever in a new position.

This system moves by jerks. It is preferable to use electric motors, working on 2 to 4 volts and reversible by reversing their polarity; the shaft would be in the form of a worm gear into which a spur gear would fit. This would give a smoother control of direction.

Editor's note

The advanced radio experimenter will have little trouble with the construction of suitable sending and receiving equipment, which may differ widely in circuit design from that described here. However, he may find a recent American receiver circuit interesting. This, using the Raytheon RK-61 thyratron tube, is shown in Fig. 7. For greater ranges, it

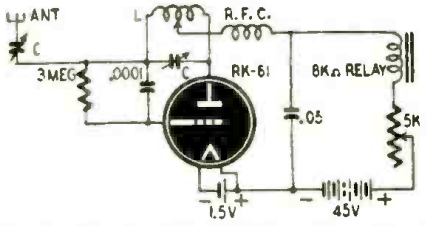


Fig. 7—Circuit of the RK-61 receiver. This is a sub-miniature thyratron, manufactured by Raytheon, operating with a B-battery of 45 volts; filament uses .05 ampere at 1.4 volts.

of course may be used as the second tube of a 2-tube receiver, and can be coupled readily to a superregenerative detector by any convenient means, such as an ordinary audio-frequency transformer.

Translated from the French, by special arrangement with the magazine La T.S.F. Pour Tous (Paris)

The FCC has revised and reissued its primer. The ABC of the FCC, a nine-page booklet first published in 1940. The object and nature of FCC regulation; origin and set-up of the Commission; broadcast-application procedures; assignment of call letters, and other FCC activities are explained in it in simple question-and-answer form.

Communications

PROGRESS DUE TO EXPERIMENTERS ONLY

Dear Editor:

Here is one who takes your predictions and opinions with several pounds of salt before digesting.

Where are all the great improvements, where are the innovations, the better radios, and the urge of the public to buy these millions of radios you predicted?

The great hoop-la of the coming radio has died down. The only advance in Radio has been in the price, as I thought all along and still think. And this is the only logical way to think. The radio makers are interested in profits, not radio, and will be to the end of time, and the 5 to 8 tube superheterodyne fits the bill very nicely.

The only dynamic work left in the Radio reception field is the regenerative circuit, which in my opinion has been but half explored. T.r.f.'s with regeneration—fixed crystals with regeneration—have hardly been touched.

In closing I can only say the public will get no advances in Radio unless it means more profit for the makers, like FM, etc.

All advances are known and shared only by a small coterie of home construction experimenters and hams. And even these are placed in a disadvantage. The manufacturers refuse to make ready-made coils of the plug-in or switch type any more. Anyone with the factory equipment capable of turning out coils for construction would clean up.

H. L. LUCE
(No Address)

(Few manufacturers would quarrel with Mr. Luce's assumption that they are in business for the money to be made in it; not because they are interested in radio. While there is a tremendous market for small, easily constructed receivers, it would be naive to expect the manufacturer to refrain from making them.

When the customer is sated with a.c.-d.c. "All-American" midgets and wants something better, the manufacturer will be able to make such sets more profitably than the cheaper ones, and there will be no shortage of them on the market. The same goes for plug-in coils. At present the coil companies are working at full capacity supplying coils for the new receivers. As soon as it becomes profitable to wind coils for the much smaller experimenters' market, they will become available.—*Editor)*

R-C HEADS 15-MAGAZINE LIST

Dear Editor:

Just received the December number, which is an anniversary number for me, as it is exactly ten years since I bought my first copy of RADIO-CRAFT *(Short Wave Craft* one year earlier). I should like to congratulate you on the amazingly high standard your magazine has maintained over the period. Yours is the only magazine I have purchased without a break for ten years (except of course the copies I lost to the U-boats). I now look forward to the next ten years!

One small brickbat. Could you hurry up the dispatch of RADIO-CRAFT? The December copy issued on the 25th of November arrived here on the 20th of February. Three months is a very long time. Most of the other magazines from the U.S.A. arrive the same month and are never more than a month late.

In closing I should like to state I subscribe to 15 British and U.S. radio magazines, with yours right at the top.

A. W. J. MARSH,
Isle of Wight, England

RADIO TERM ILLUSTRATED

Suggested by:
Wm. F. Wendt,
Napa, Calif.

Saturation Point

"THE CUSTOMER WILL GYP YOU . . ."

Dear Editor:

I have never written to any publication before, but now I feel I must. My reason: the disparaging remarks that are circulating among the layman in regard to the honesty of the radio serviceman. Personally I am not a serviceman; but I have spent six years of my life as a communications technician, two of those years as a radar technician for Uncle Sam's Army, and the rest in the fascinating hobby of amateur radio.

The other day I met an old friend and his buddies. My friend seemed peeved at something, and on inquiring I found that his dislike for the radio serviceman was immense. It seems that his Philco was in bad shape and he had taken the mess (and I do mean MESS) in for repairs at his local radio store. The estimate ran to about $21.60. He tried several other radio shops and received approximately the same estimate. Of course the radio serviceman was an out-and-out crook, screw-driver mechanic, and an assortment of other choice cuss words!

For your benefit I shall list the cause of his complaint:

One open r.f. coil (replaced)
Dial drive cable and pointer (installed)
One noisy volume control (replaced)
Multisection electrolytic filter unit (replaced)
One power tube (replaced)
Re-center voice coil and alignment.

After my examination of his Philco, I wondered why the serviceman had not doubled his previous price. I feel that the price of $21.60 was quite fair. How do you feel about it? How did that radio get into that condition?

I've traveled a great deal about this country and from my own experience in speaking with the average layman, they have the same thing to say about the radio serviceman—with a few exceptions. Can anything be done to remedy this appalling situation? It seems that the layman judges the service fraternity not only by the unscrupulous few that exist, but by his own wishful and uninformed thinking as to how much a repair should cost.

PHILIP SPAMINATO,
New York, N. Y.

"COILS, CORES, MAGNETS"

Dear Editor:

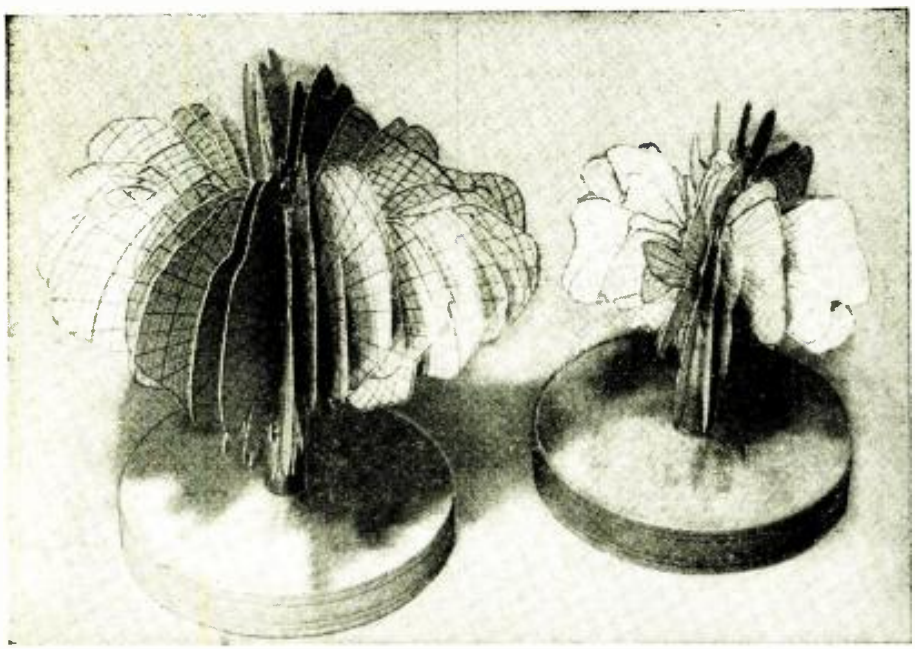
I certainly appreciate reading articles such as "Coils, Cores and Magnets" by H. W. Schendel (October and November RADIO-CRAFT). I found an error on page 42. The decimal point was accidentally omitted in calculating the effective cross-sectional area of metal in the core. (0.80 x 0.75 inch equals 0.6 sq. in.)

Permit me to suggest that you run more articles of the nature of that one. Materials on peaking transformers, saturable reactors (not swinging reactors) and devices or control circuits for electronic counting, sorting, regulating etc., would be especially welcome.

ROLAND R. RHODES,
Hackensack, N. J.

(An article by Jordan McQuay on electronic counting and sorting circuits will appear in an early issue.—*Editor)*

THREE DIMENSIONS ON PAPER



Ohio State University has solved the problem of designing non-confusing models of antenna field patterns by using a series of 2-dimensional cardboard figures. Each represents the vertical field pattern in a given direction. Energy radiated in each direction is indicated by the radial distance from the center of the cluster to the edge of the pattern elements.

INDEX TO ADVERTISERS

RADIO SCHOOL DIRECTORY
(See Page 80)

Television for the middle-sized community is possible, say Sarkes and Mary Tarzian, who May 8 were granted a construction permit for a television station at Bloomington, Indiana. To insure a large audience, Mr. Tarzian has announced that he will manufacture several thousand television receivers and sell them "at the lowest possible figure" to persons within range of the station.

BOOK REVIEWS

APPLIED PRACTICAL RADIO, by the Technical Staff, Coyne Electrical and Radio School. Published by Coyne Electrical School. Three volumes, 306, 299 and 403 pages, plus volume index in each volume and additional master index in first volume. Stiff cloth covers, 5½ x 8½ inches. Price per set, $9.75.

This text was written to fit the needs of a resident radio and television course, at the same time keeping in mind the person who wishes to increase his knowledge of radio but has not the time or money for a resident or correspondence course.

The first volume deals with elementary electrical theory. An effort has been made to approach the subject from a fresh viewpoint. A number of new analogies and illustrations appear, and where old ones are used, they are dressed up with new types of drawings which add to the interest.

The second book covers applied radio theory, beginning with alternating currents in radio, and continuing through resonance and tuning, r.f. coils and radio tubes to r.f. coupling and coupling circuits. The third volume is devoted to radio receivers, including chapters on antennas, amplifiers and loudspeakers, receiver controls, and receiver alignment.

Superheterodyne receivers, automobile radios, public address systems and shortwave reception receive a chapter apiece, while two chapters each are devoted to FM and FM receivers and to television and television receivers. A large number of breakdown drawings are used to explain the text.

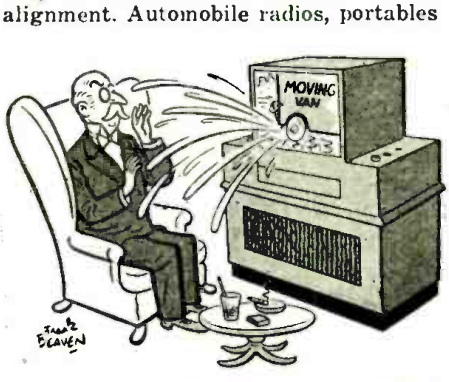
ELEMENTARY RADIO SERVICING, by William R. Wellman, chairman, radio department, Brooklyn High School for Specialty Trades. Published by D. Van Nostrand Co. Stiff cloth covers, 6½ x 9 inches, 260 pages. Price $3.00.

Writing for the man who has gained some knowledge of radio construction or has had vocational-school or wartime training, the author seeks to show how such knowledge may be applied to radio servicing.

The first chapter deals with the use of test equipment, then there are chapters on each section of standard receivers, on miscellaneous troubles and on alignment. Automobile radios, portables and FM receivers are given special treatment.

Job sheets and experiments are included with almost every chapter. These, together with lists of questions and references, fit the book for use as a text in an elementary service course as well as for independent study.

No radio theory or mathematics are included. The approach is purely practical, presupposing a knowledge of elementary radio theory and familiarity with schematics.



Suggested by Grego Banshuck

ELECTRONIC CAPACITORS, by Paul McKnight Deeley. Published by the Cornell-Dubilier Electric Corporation. Stiff cloth covers, 5½ x 8 inches, 276 pages. Price $3.00.

Written by a practical engineer, this book is much more "down-to-earth" than earlier works on the subject. Description of the fabrication of capacitors and the forming of anodic films is unusually thorough and complete.

In some directions, the author has been limited by the fact that various manufacturers have processes that may be termed trade secrets. While disclosure of such processes has been studiously avoided, the general subject has been covered completely. The "leaning over backwards" attitude shown in some books, which avoid printing even formulas for electrolytes, is absent.

The presentation is clear. According to the slip cover, the intention was to write "in simple language readily understood by students, repair and maintenance men and highly trained engineers alike," and that object was attained.

THE ELECTRONIC CONTROL HANDBOOK, by Ralph R. Batcher and William Moulic. Published by Caldwell-Clements, Inc. Flexible leatherette covers, 6 by 9¼ inches, 344 pages (including a 3-page bibliography and an index). Price $4.50.

This well-written work, with its large number of charts, diagrams and photographs, is presented in such a manner that it can truthfully be called a "handbook." It is a useful compilation of twenty-one chapters of essential data on electronic control systems and their possible applications in industry.

Its chapters are grouped into five sections according to the type of material under discussion. The presentation is such that all material can be readily understood by students, engineers and maintenance men with a basic knowledge of electronic principles.

The five sections; Basic Elements of Control, Conversion Elements, Electronic Modification Circuits, Activation Elements and Control Motors present, in a useful light, discussions of methods of converting pneumatic, hydraulic, mechanical, electrical and other physical changes into electronic voltage or current variations to actuate control mechanisms.—R.F.S.

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Founder

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Electrical Experimenter	1913
Radio News	1919
Science & Invention	1920
Radio-Craft	1929
Short-Wave Craft	1930
Wireless Association of America	1908

Some of the larger libraries in the country still have copies of ELECTRICAL EXPERIMENTER on file for interested readers.

From July, 1913, ELECTRICAL EXPERIMENTER:

How to Select and Operate Wireless Telegraph Sets, by *H. Winfield Secor.*

How to make Selenium Cells, by *Samuel Wein.*

A Handy Vacation Wireless Set, by *Henry Scott.*

Horizontal Antenna

New Wireless Instruments, by *H. Gernsback.*

New Ear Telephone

SIGNAL GENERATOR

(Continued from page 56)

tenna post of the receiver. The tuning condensers of both units are fully meshed. The receiver is tuned slowly across the band and the frequencies where the oscillator signal is heard are noted. The frequency difference between any two points on the receiver dial is equal to the oscillator frequency. This frequency is marked on the oscillator dial. The oscillator is shifted slightly and new repeat points noted.

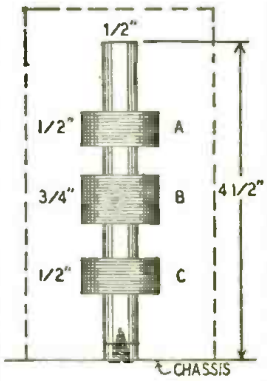

Fig. 2—Coil details.

This process is continued until band A is completely calibrated. The same method is used to calibrate band B. Band C is calibrated by reading the frequency directly from the dial of the receiver.—*Les Jones*

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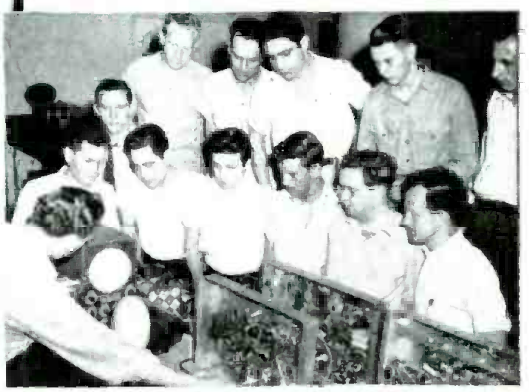

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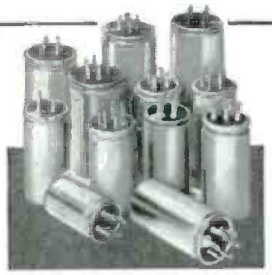

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